

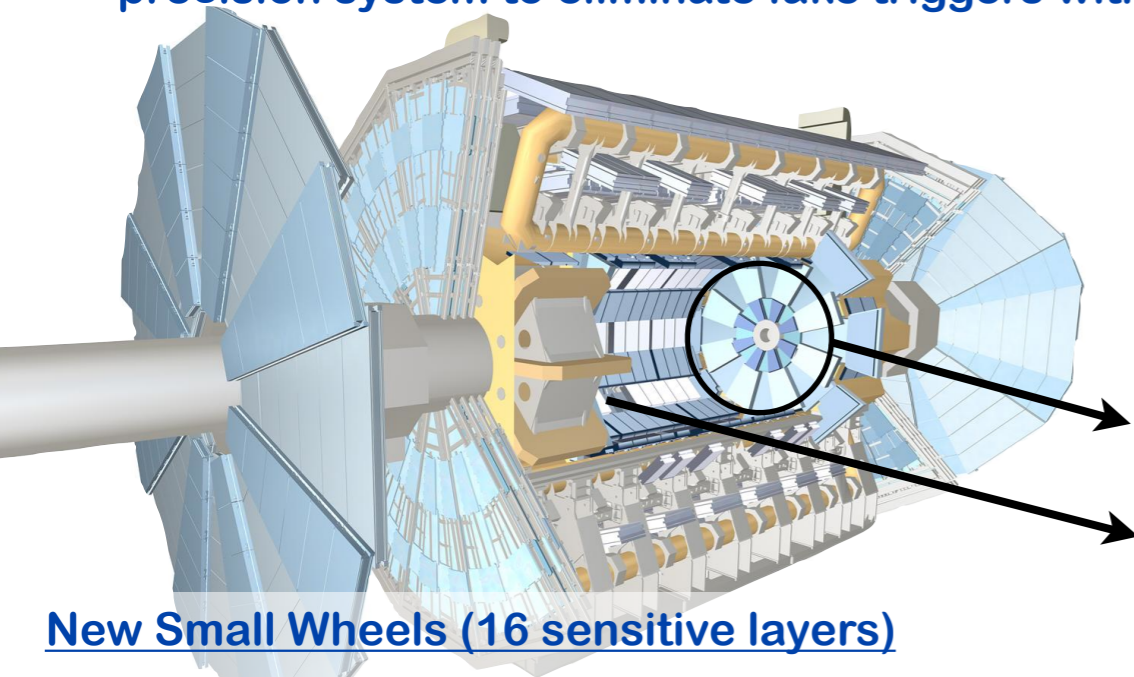
VMM ASIC

George Iakovidis

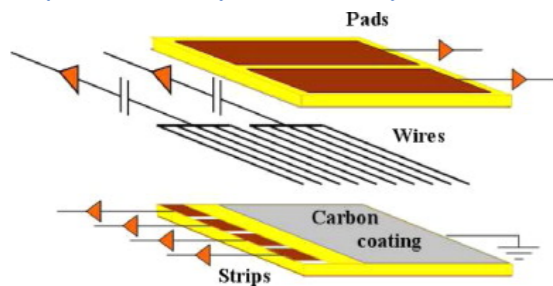
*Physics Department
Brookhaven National Laboratory*

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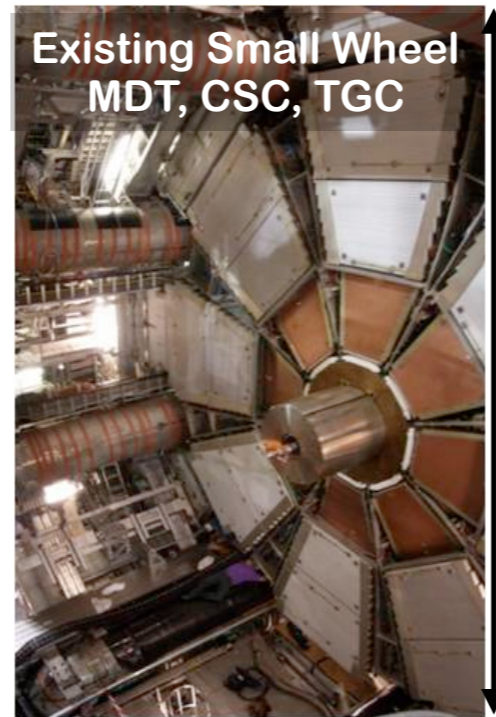
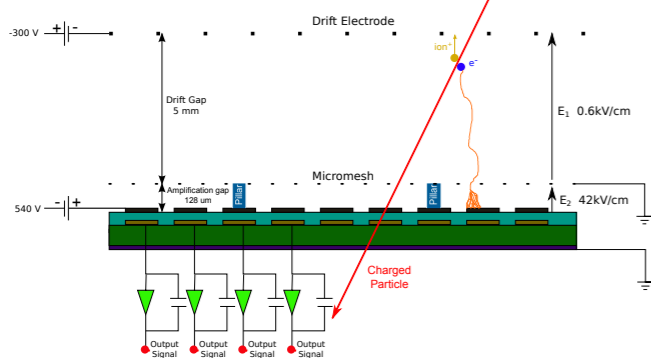
- The ATLAS upgrade is motivated primarily by the pile-up rate expected at high luminosity which will lead to an increased trigger rate. There is a need of replacing the innermost muon station with an efficient trigger and precision system to eliminate fake triggers without loss on physics acceptance.



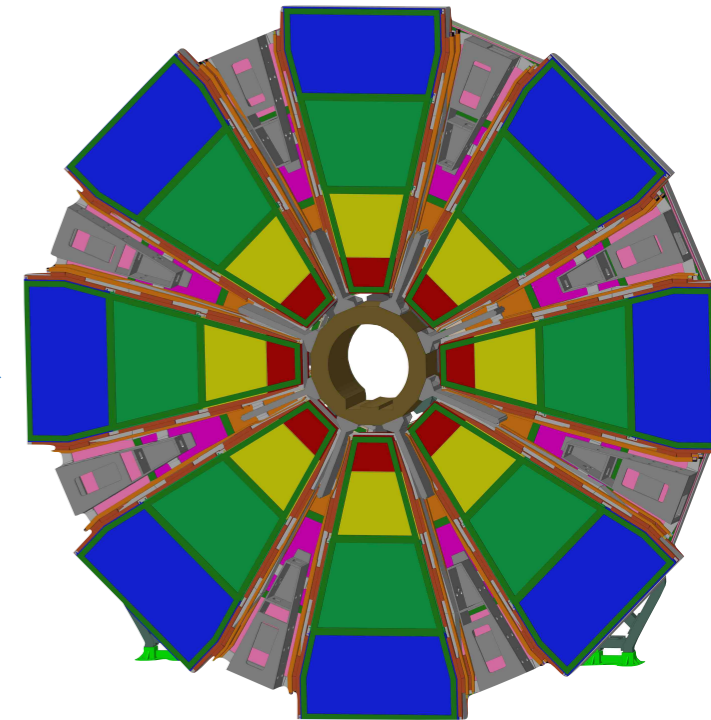
- sTGC (8x)
(small strip Thin Gap Chambers)



- Resistive strip Micromegas (8x)
(Micromesh Gaseous Structure)



NSW →



Front-end Electronics Requirements (need of custom ASIC)

- Another challenge for this Project - More than 2.4 million channels total (2.1M for Micromegas and 300k for sTGC)
- Operate with both charge polarities
- Sensing element capacitance 10-200pF (sTGC Pad up to 2nF)
- Charge measurements up to 2pC @ < 1fC RMS (6pC for sTGC pads)
- Time measurements ~ 100ns @ < 1ns RMS
- Multiple Trigger primitives, complex logic
- Low power, programmable
- Space requirements on the detector

New Small Wheel Technical Design Report: [Link](#)

Technical Design Report for the Phase-I Upgrade of the ATLAS TDAQ System: [Link](#)

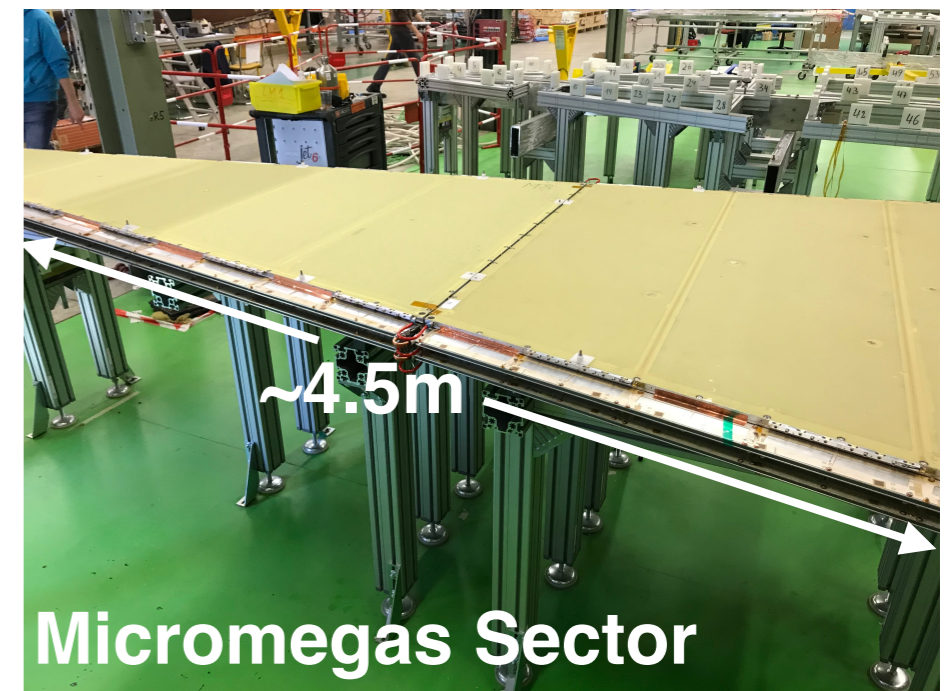
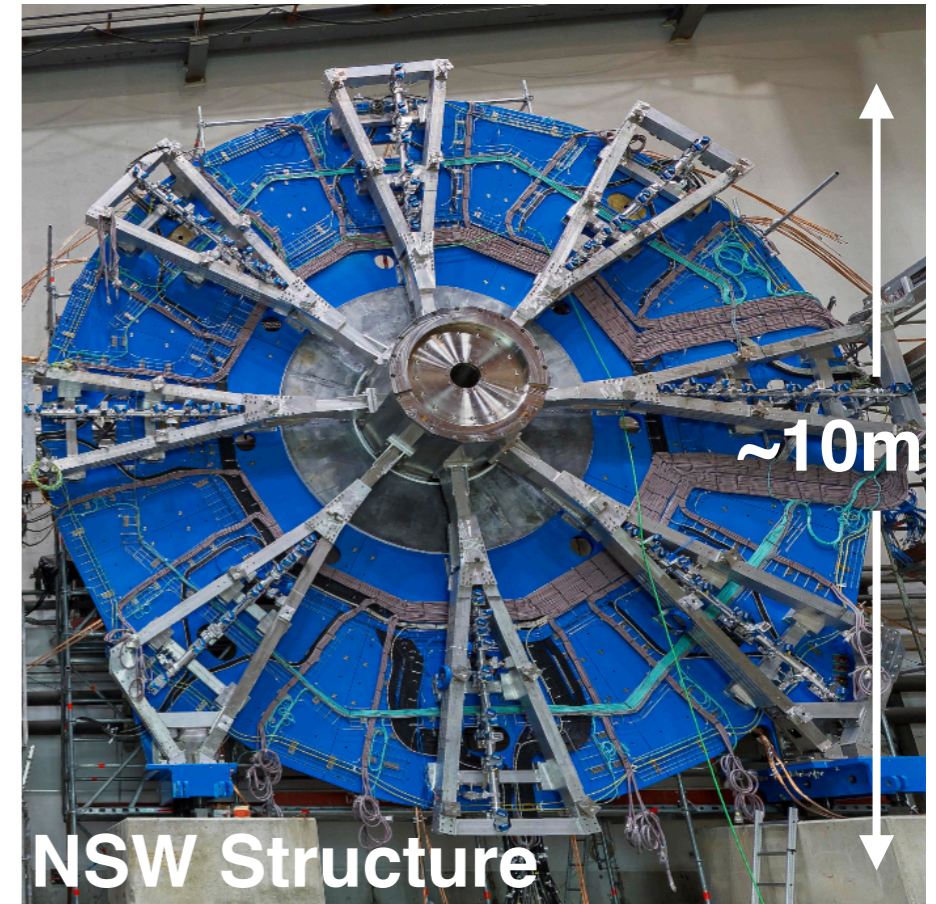
*The **VMM** was developed in the context of the **ATLAS**

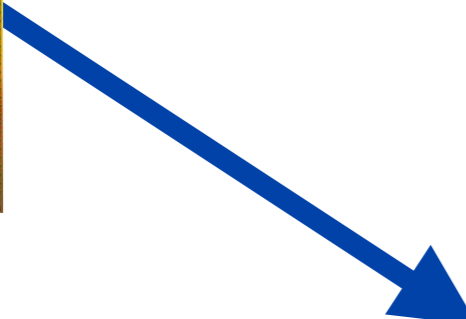
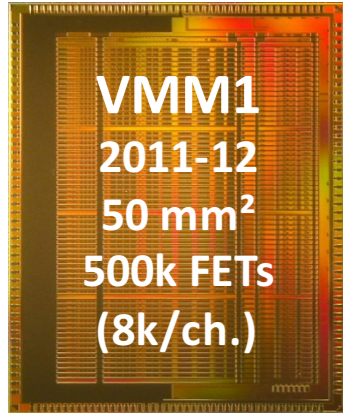
New Small Wheel Upgrade

- Largest MPGD development and densest Muon spectrometer upgrade to date
- ATLAS Muon spectrometer 1.2M
- NSW alone is 2.4 M channels

*In that talk:

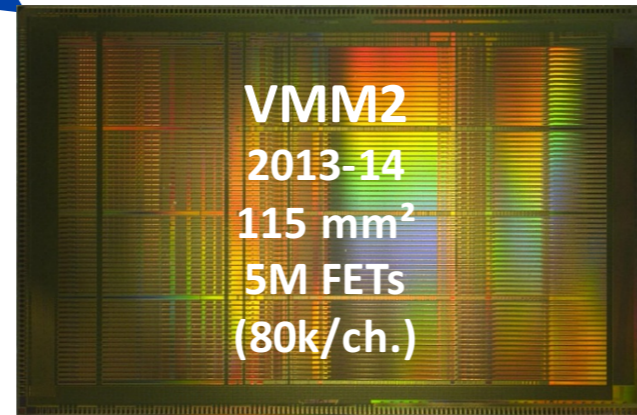
- **VMM evolution** and the **production version**
- Functionality, **architecture** and readout
- **Performance** highlights
 - Bench measurements
 - Resistive Micromegas test beams
- Integration **highlights** on high channel density Micromegas detectors
- **Production**, reticle layout of the **wafer**





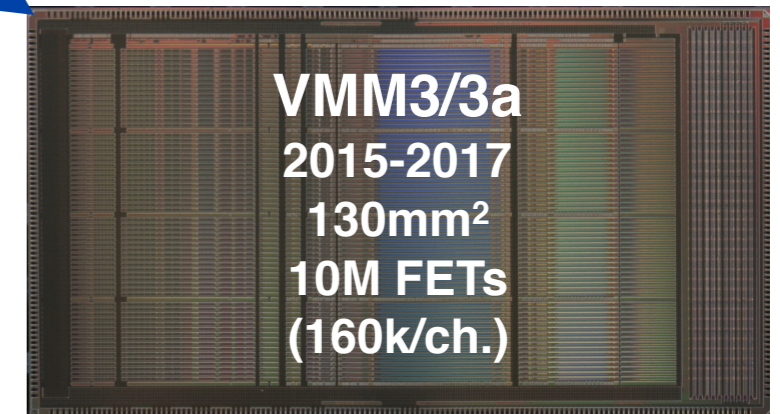
- ✓ Mixed-signal
- ✓ **Continuous** readout
- ✓ Current-output peak detector
- ✓ **Increased** range of gains
- ✓ **Three ADCs** per channel
- ✓ FIFOs, **serialised data with DDR**

- ✓ **Mixed-signal**
- ✓ 2-phase readout with external ADC
- ✓ **peak and timing** information
- ✓ neighbouring readout
- ✓ sub-hysteresis **discrimination**
- ✓ few timing outputs



- ✓ Serialised ART with DDR
- ✓ Additional timing modes
- ✓ **64 timing outputs**
- ✓ Additional functions and fixes

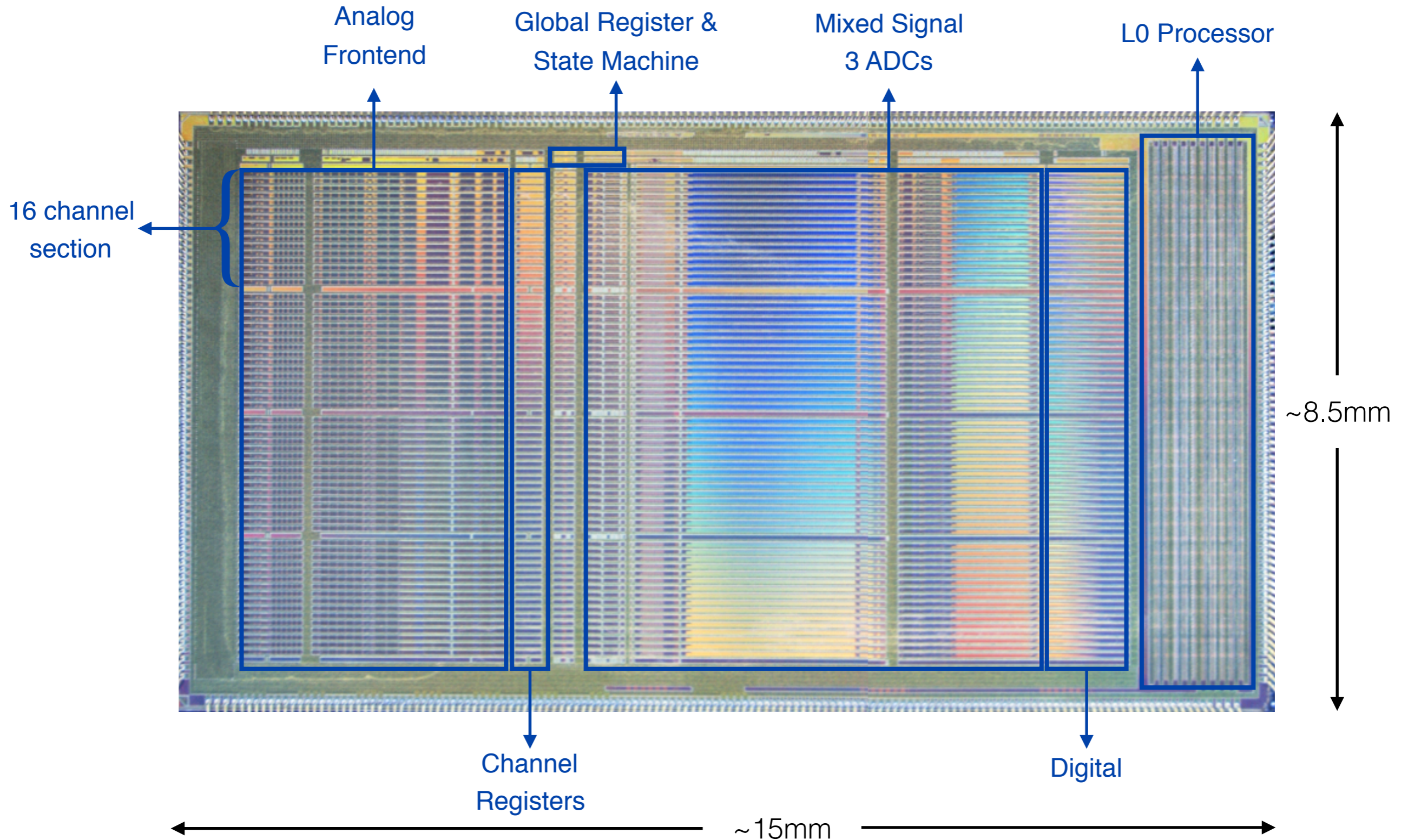
- ✓ **LVL0 pipeline** and buffering for ATLAS
- ✓ **SEU-tolerant logic**
- ✓ **Revised front-end** for high charge and capacitance (2nF, 50pC, fast recovery)
- ✓ SLVS signals
- ✓ Reset controls
- ✓ **Timing at threshold**
- ✓ Timing ramp optimisation
- ✓ Ion **tail suppressor** (fast recovery)
- ✓ Int. Pulser range extension
- ✓ ART synchronisation to BC clock
- ✓ additional functions and fixes
- ✓ **VMM3a fixed open bugs** from VMM3 and introduce some stability fixes on the ADCs and Front-end

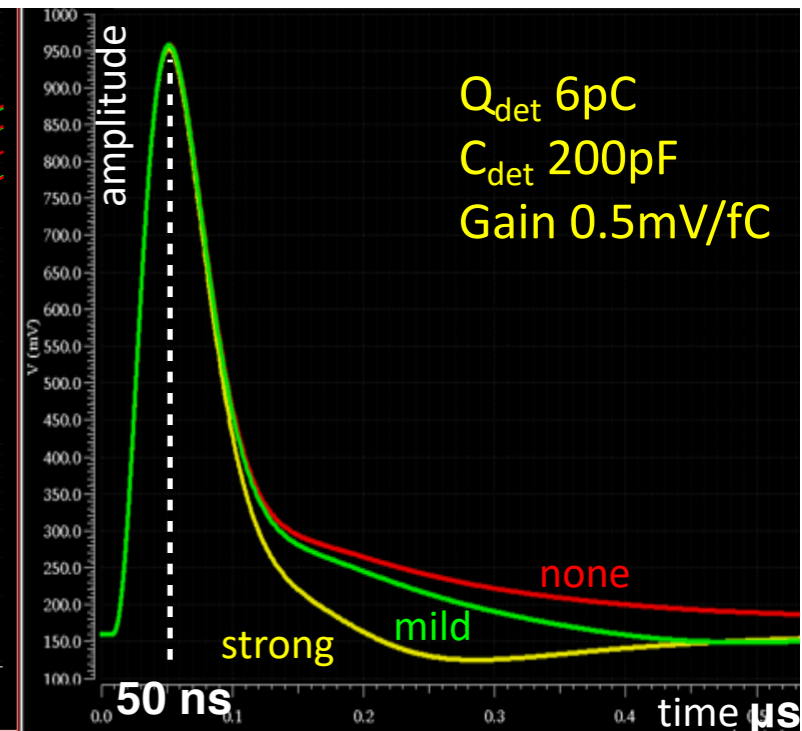
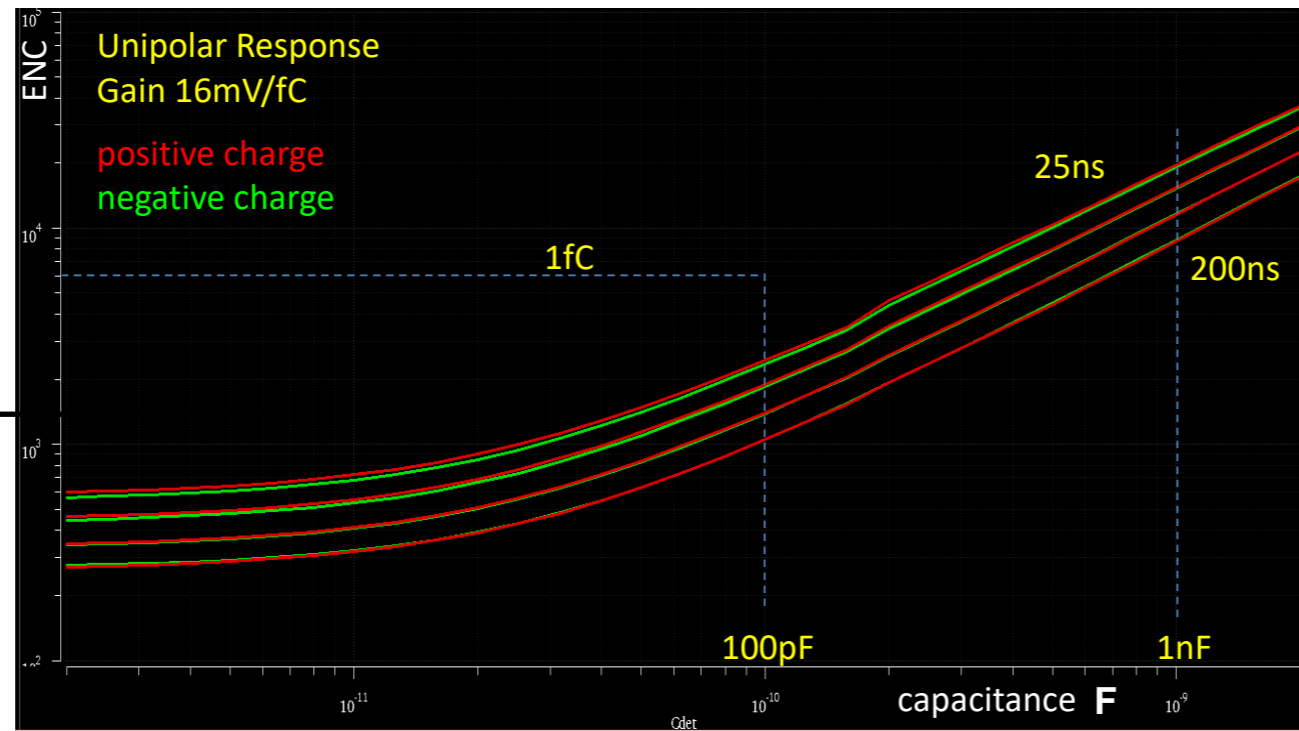
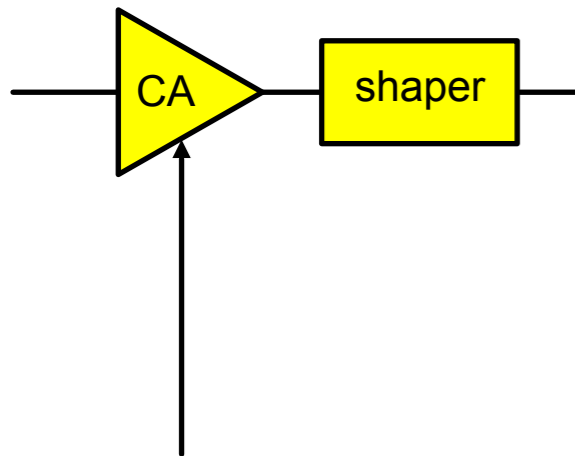


VMM3a - Production Version !

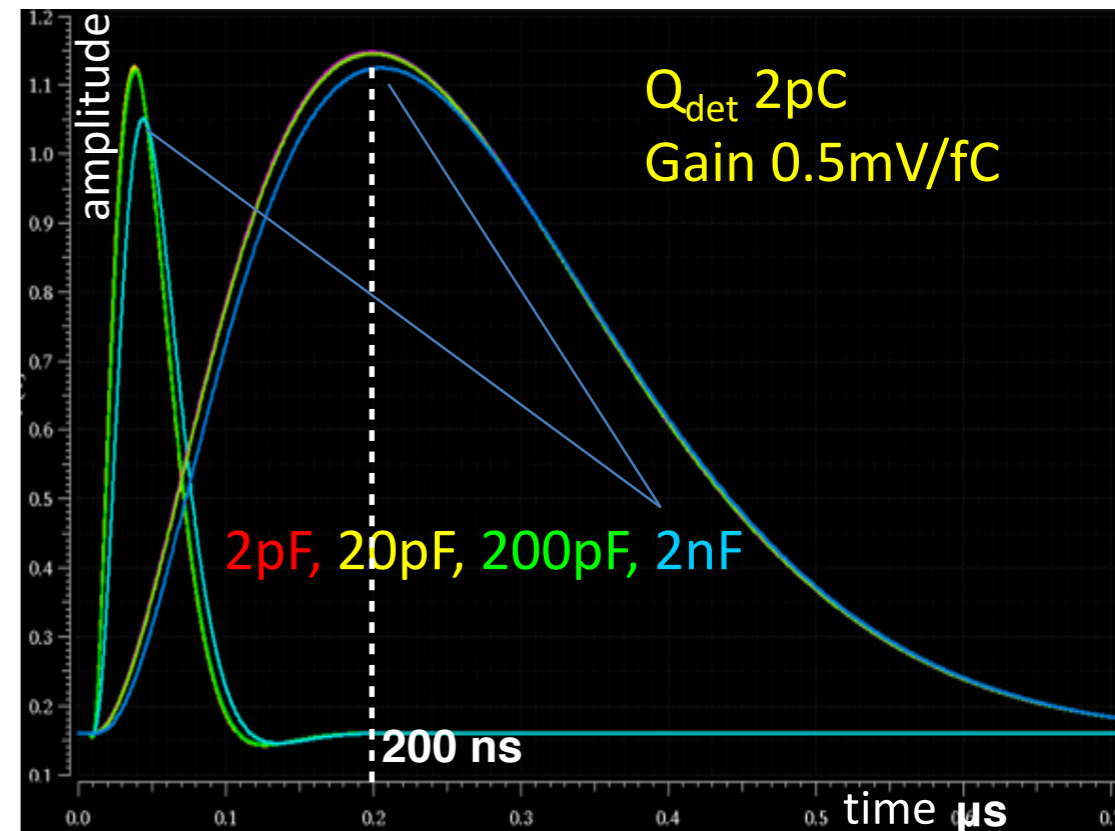
- ★ *The VMM was designed at BNL in collaboration with IFIN-HH*
- ★ *It is fabricated in the 130nm Global Foundries 8RF-DM process (former IBM 8RF-DM)*

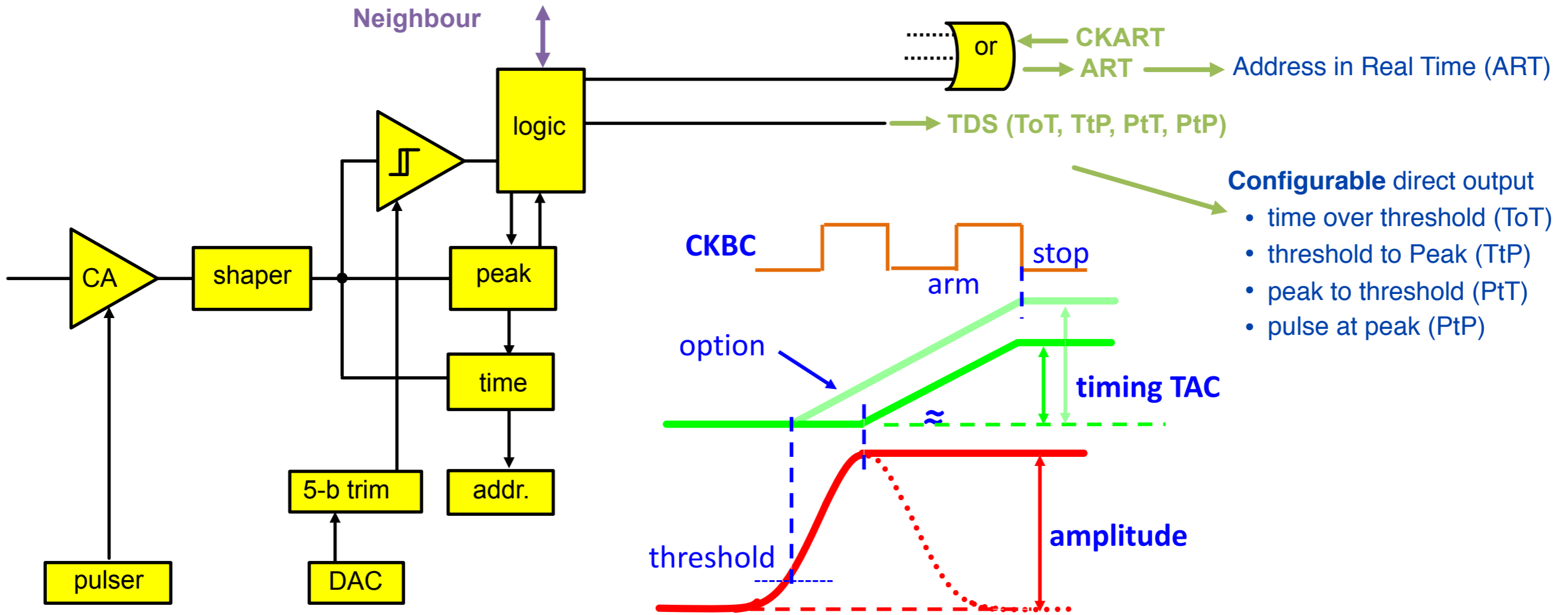
- The ASIC features **64 channels** that extend along the size of the die. At the end the L0 section (explained in later slides) is separated to isolate the noise from the digital activity





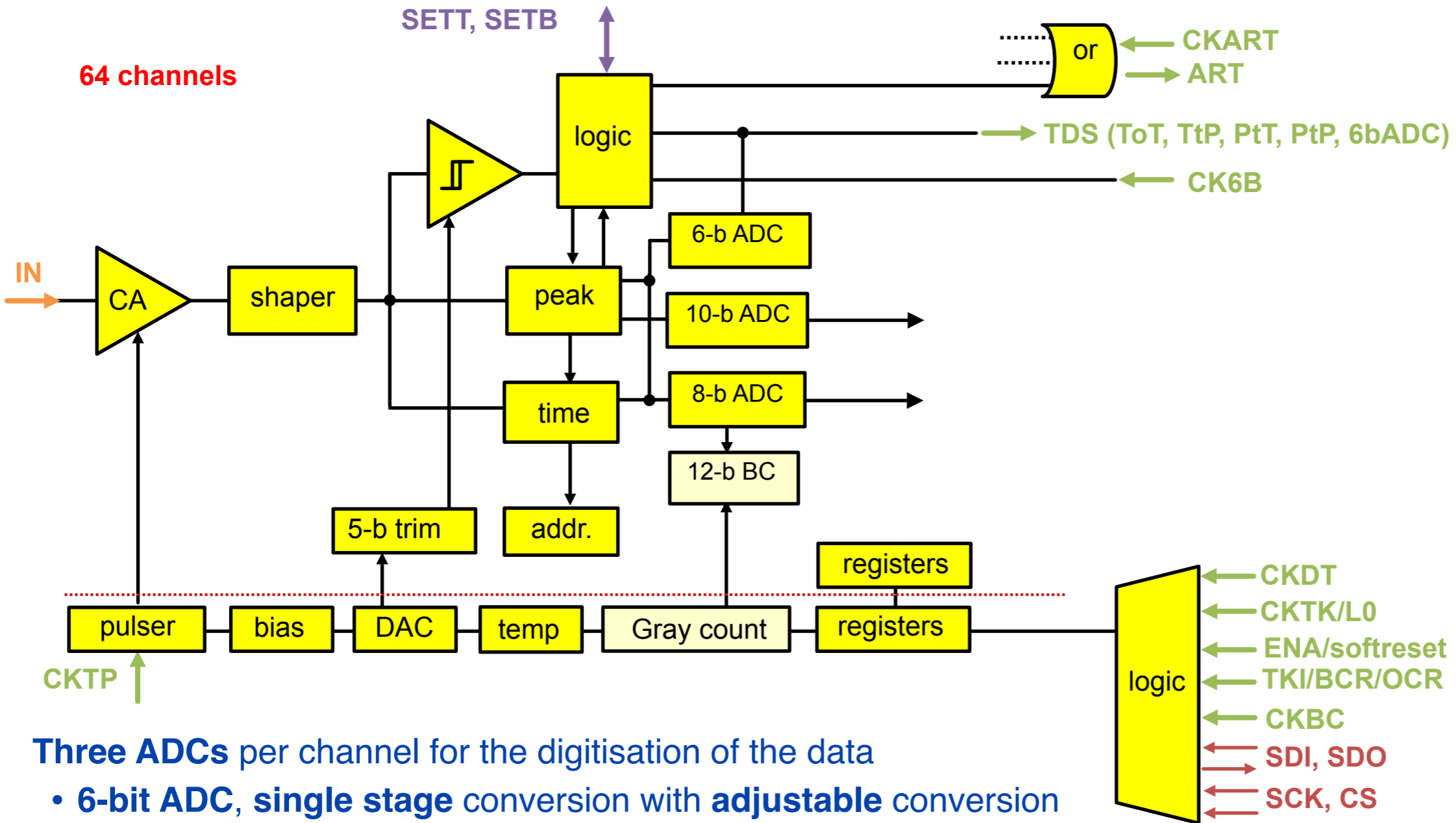
- Input transistor: **PMOS** 180 nm x 20 nm, 3 stage amplifier,
 - 2 stages used for **adjustable gain**: 0.5, 1, 3, 4.5, 6, 9, 12, 16 mV/fC
 - 1 for **adjustable charge polarity**: positive or negative
- Input **capacitance**: can operate from sub-pF to several nF
- Maximum **charge**: 2 pC in linear range, fast recovery from 50 pC
- Semi gaussian **shaper 3rd order**
 - **Configurable** ion tail **suppression**: none, mild or strong
 - **Adjustable peaking** time: 25, 50, 100, 200 ns
 - Leakage-adaptive, DDF shaper, **BGR-stabilised baseline**





- Configurable direct output**
- time over threshold (ToT)
 - threshold to Peak (TtP)
 - peak to threshold (PtT)
 - pulse at peak (PtP)

- Global 10-bit DAC for adjusting the **threshold - Discrimination** with sub-hysteresis (effective 2mV)
- Adjustable **5-bit discrimination** threshold **per channel** to adjust at ~mV level
- **Neighbour** logic to trigger sub-threshold channels with inter-chip communication
- Configurable **direct output** per channel and serial fast output of address as an OR of all channels
- **Peak detection**: measurement of peak **amplitude** and storage in analog memory
- **Time detection**: measurement of **peak/threshold** timing through a configurable time to amplitude converter (**TAC**: 60, 100, 350, 650 ns) and storage in analog memory
 - Clock working mode on **synchronous** machines but also as strobe for **asynchronous** operations

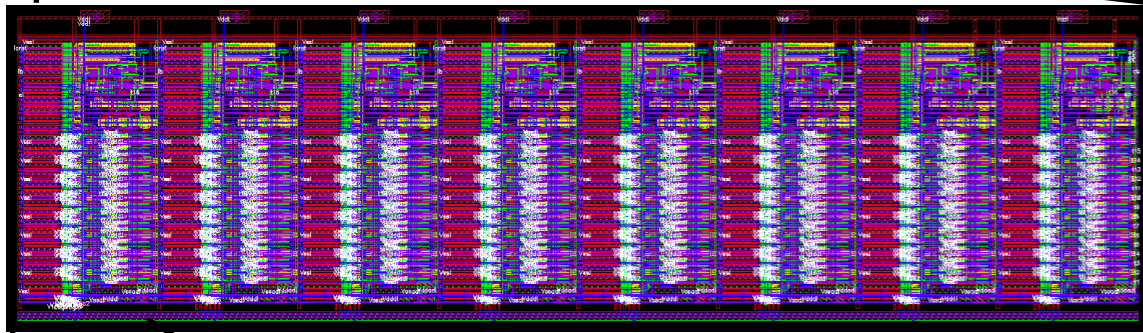


Three ADCs per channel for the digitisation of the data

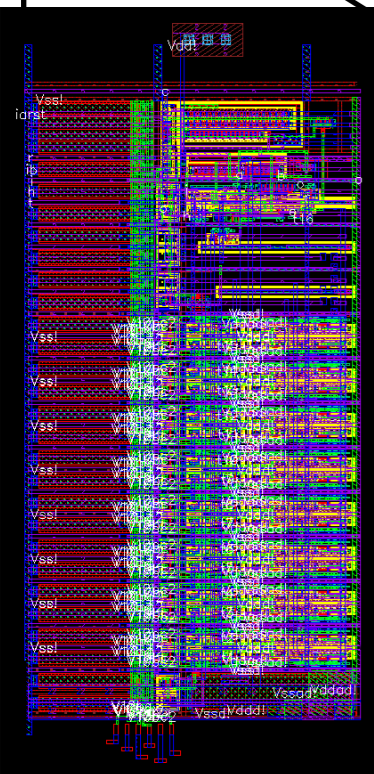
- **6-bit ADC, single stage conversion with adjustable conversion time and offset, completes within 25 ns from peak**
- **10-bit ADC, 200 ns adjustable conversion time/offset, for peak amplitude conversion**
- **20-bit timing information with 8-bit ADC, 100 ns conversion time + 12-bit Gray-code counter, BC clock**
- **2 step mode conversion for 10-bit & 8-bit ADCs** - First stage the comparison identifies one of the macro-cells and at the second stage the micro-cell is identified, possibility to jump through macro-cells

ADC Cells

8x macro-cells



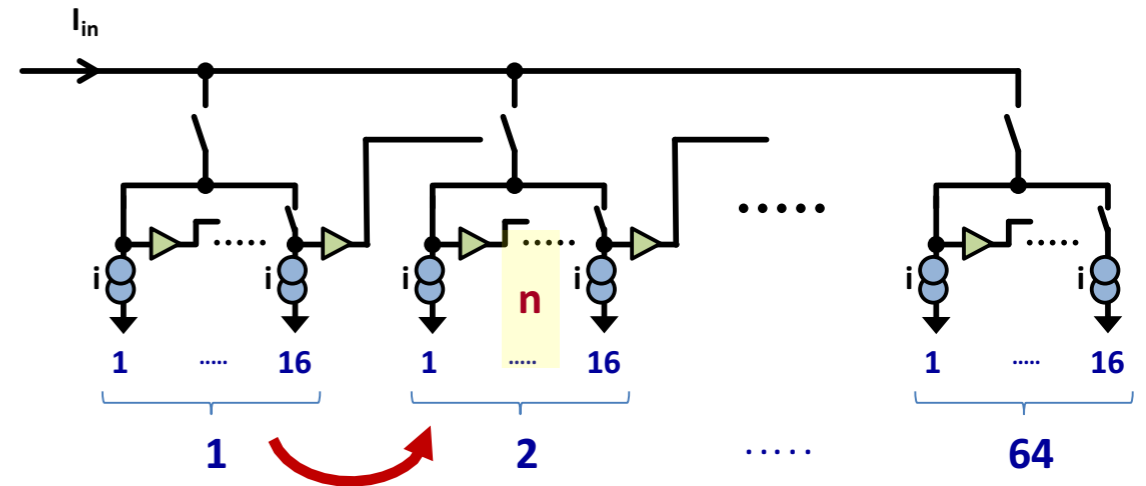
1 macro-cell



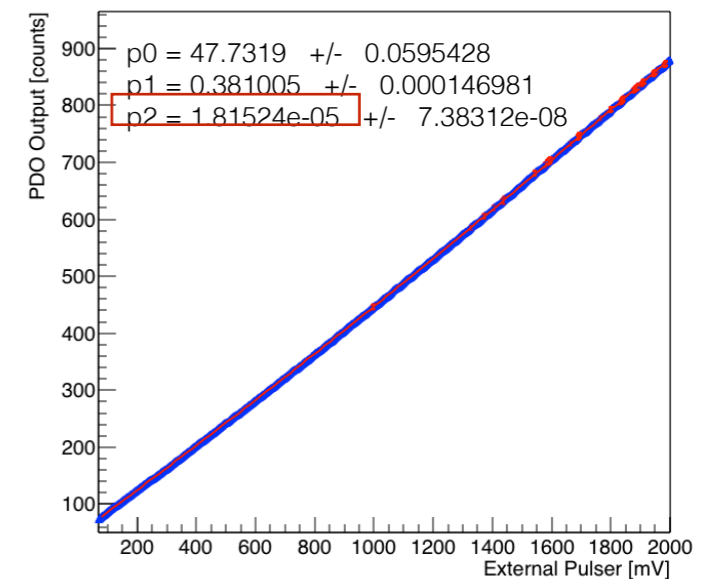
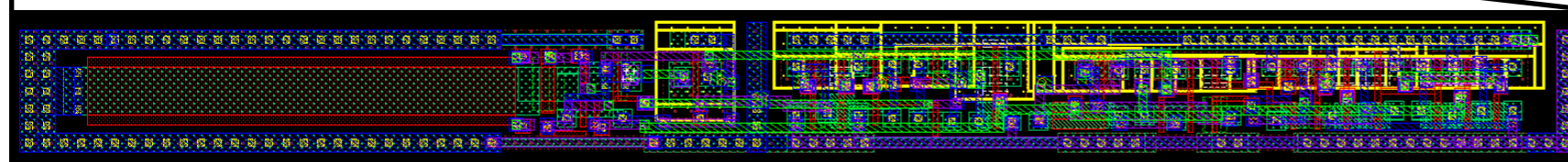
16x micro-cells

ADC Core (per channel)

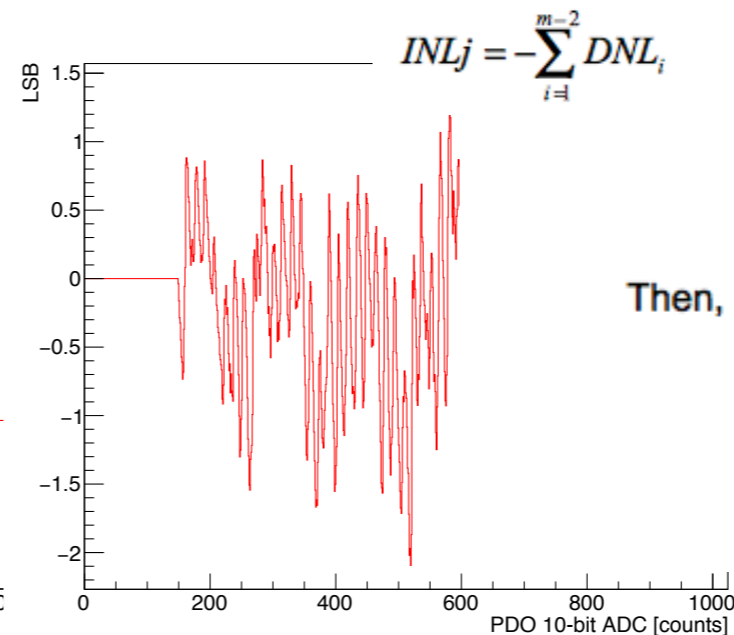
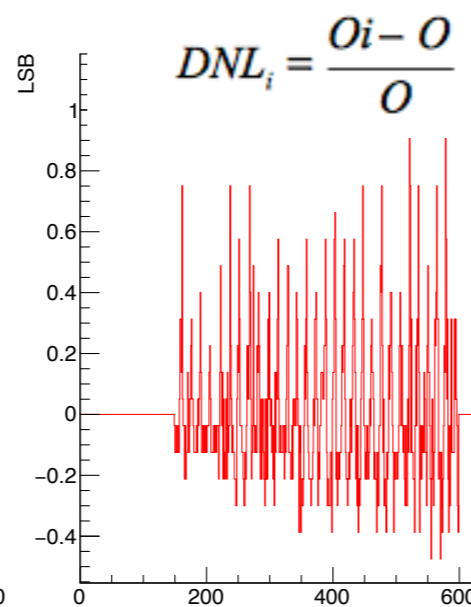
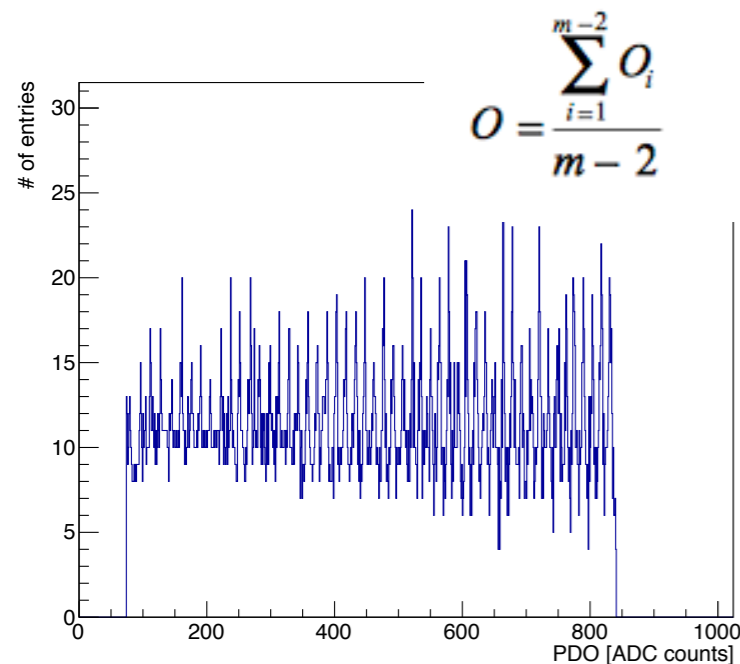
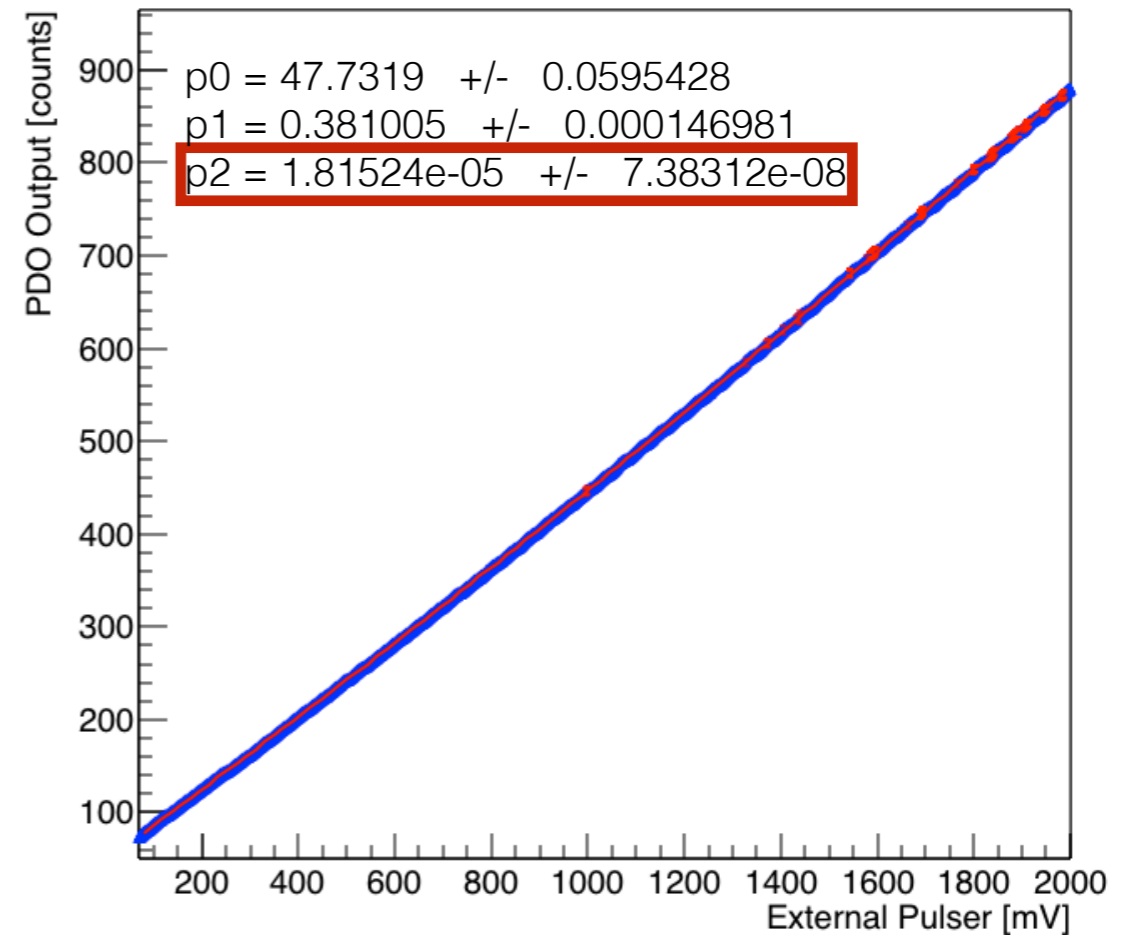
- 1024 current sources (similar to a digital thermometer)
- 64 macro-shells (6 upper bits xxxxxx0000), 16 micro-shells (4 lower bits 000000xxxx)
- **2 step mode** - First the comparison identifies one of the 64 macro-cells. Then on second step the lower 4 bits are identified (200ns conversion time + reset)
- **8 bit ADC** is built in the similar way (5+3)
- **6 bit ADC** is a **single stage conversion** similar to the 64 macro shells with fast digitisation (50ns)



1x micro-cell



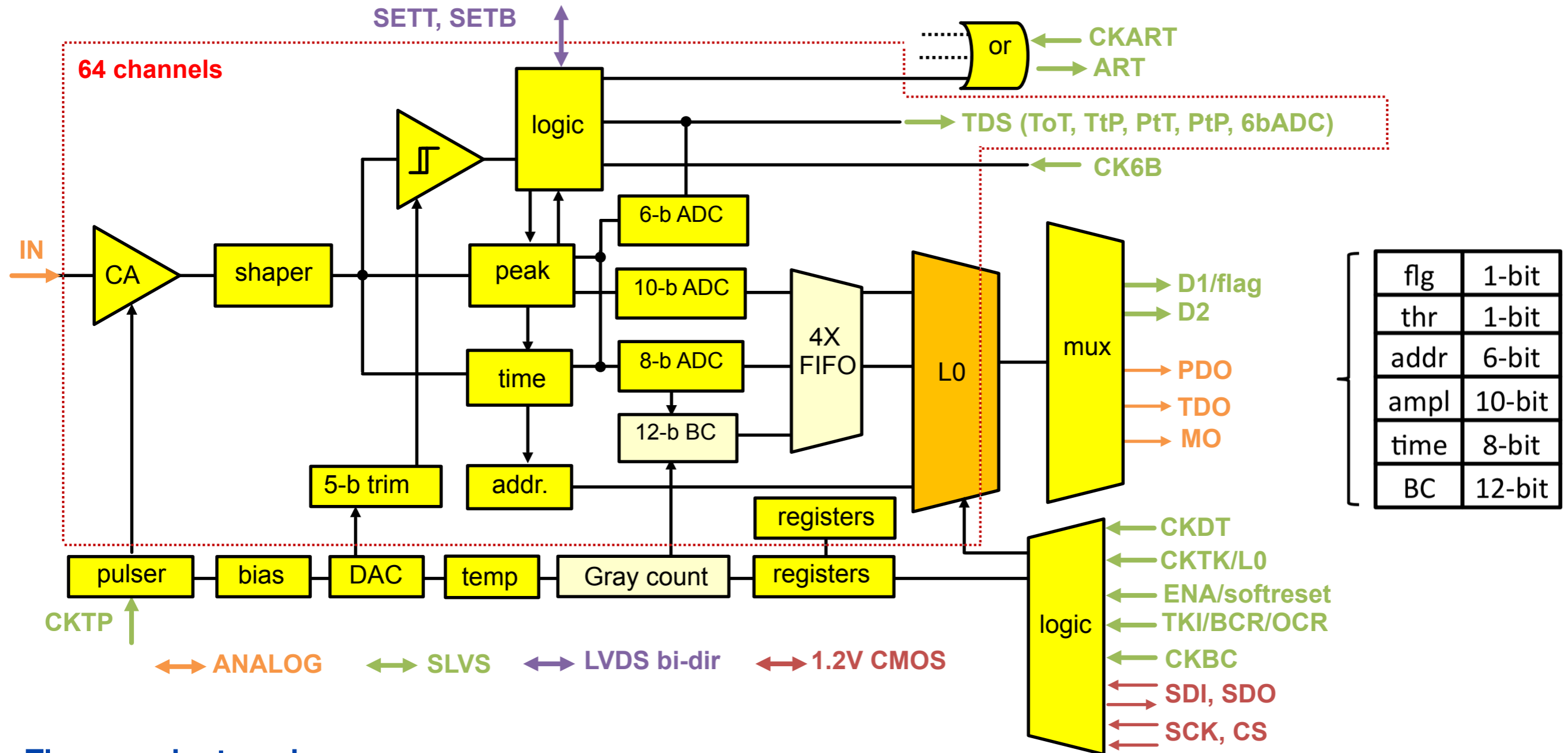
- In order to evaluate the ADC performance, a **full scan** with fine step was performed
- The ADC cannot be driven with a sinusoidal waveform for accurate estimation of its “noise” from the FFT
- In that sense the **DNL** and **INL** is calculated and used to estimate the **ENOB** of the 10-bit ADC
- The non-linearity introduced by the ADC is of the order of **2x10⁻⁵**
- **Equivalent number of bits ~8 (noise free)** for the 10-bit ADC



$$\sigma_c = \sqrt{\frac{1}{12} + \frac{1}{m-2} \sum_{i=1}^{m-2} INL_i^2}$$

Then, ENOB can be calculated as

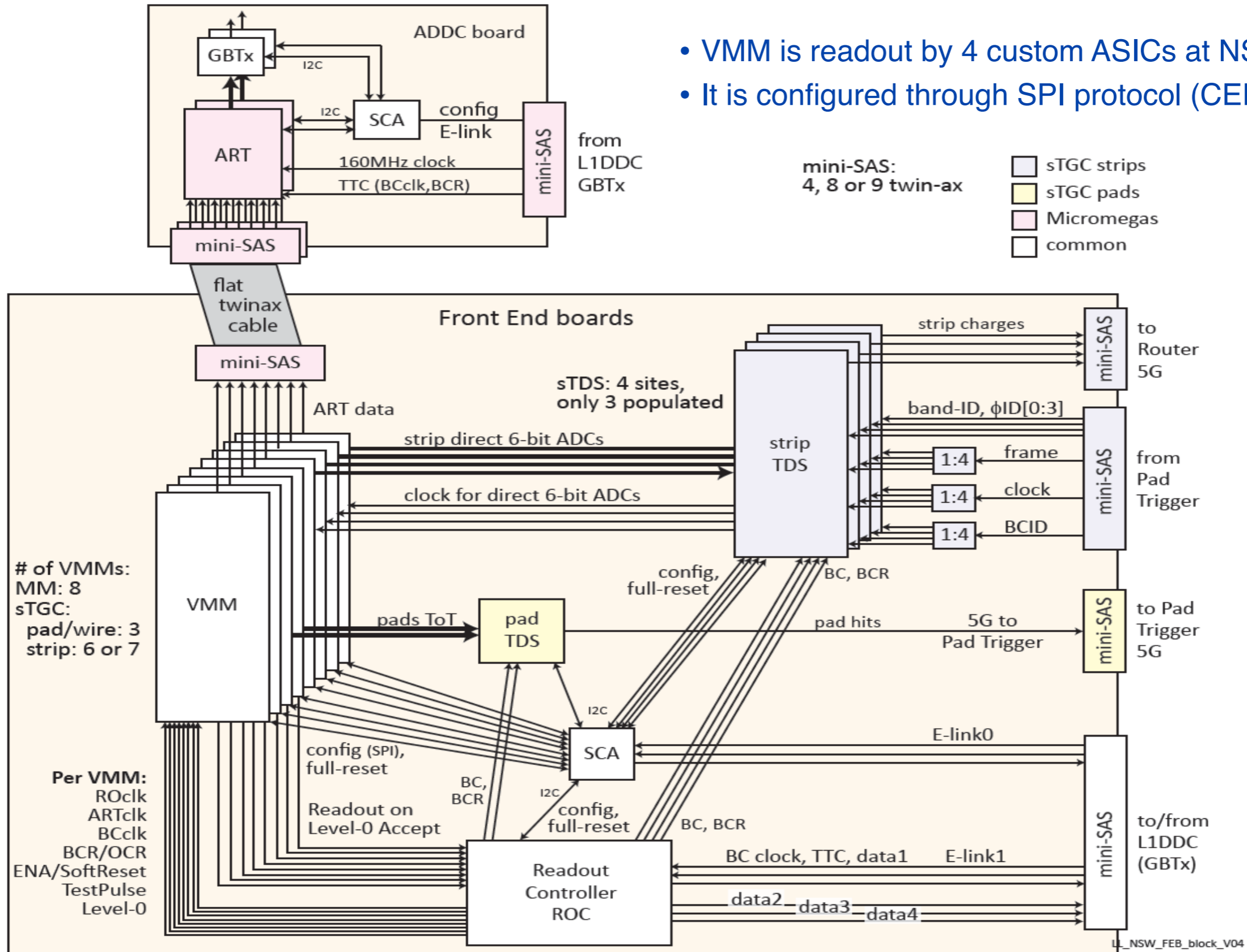
$$ENOB = \log_2 \frac{m}{\sigma_c \sqrt{12}}$$



Three readout modes

- **Mixed** mode with peak & time analog output + address (**external ADCs**)
- **Digital continuous** with **internal ADCs** and 38-bit data at 2 outputs with 200MHz DDR, **trigger-less** or with **external trigger** and auto reset
- **Level-0** processor **external trigger mode** with 64-deep latency FIFO programmable acceptance windows with 8b/10b encoding

- VMM is readout by 4 custom ASICs at NSW
- It is configured through SPI protocol (CERN SCA ASIC)



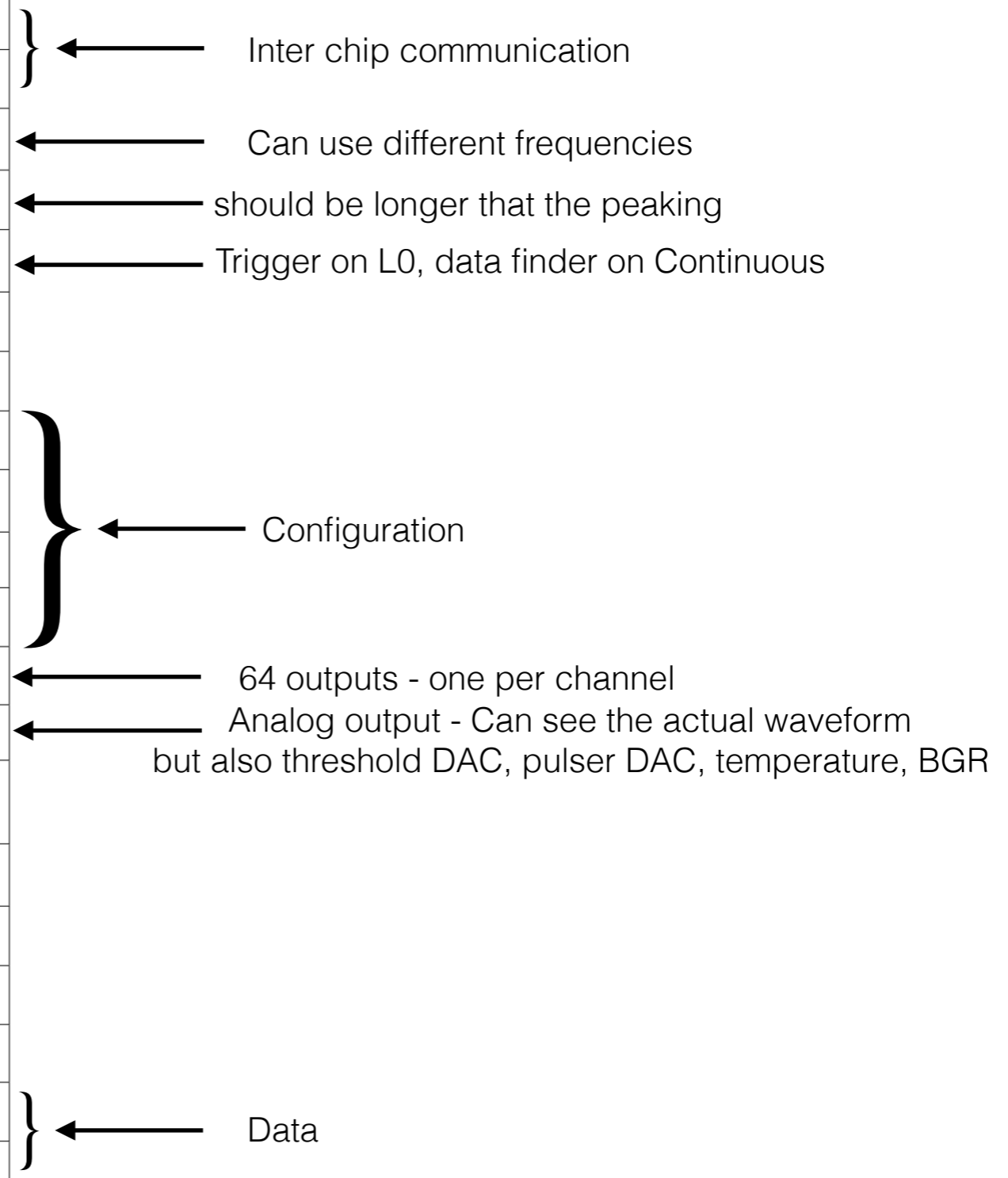
Global bits (defaults are 0)	Description
sp	input charge polarity ([0] negative, [1] positive)
sdp	disable-at-peak
sbnx	routes analog monitor to PDO output
sbft [0 1], sbfp [0 1], sbfm [0 1]	analog output buffers, [1] enable (TDO, PDO, MO)
slg	leakage current disable ([0] enabled)
sm5-sm0, scmx	monitor multiplexing. <ul style="list-style-type: none"> Common monitor: scmx, sm5-sm0 [0 000001 to 000100], pulser DAC (after pulser switch), threshold DAC, band-gap reference, temperature sensor) channel monitor: scmx, sm5-sm0 [1 000000 to 111111], channels 0 to 63
sfa [0 1], sfam [0 1]	ART enable (sfa [1]) and mode (sfam [0] timing at threshold, [1] timing at peak)
st1,st0 [00 01 10 11]	peaktime (200, 100, 50, 25 ns)
sfm [0 1]	enables full-mirror (AC) and high-leakage operation (enables SLH)
sg2,sg1,sg0 [000:111]	gain (0.5, 1, 3, 4.5, 6, 9, 12, 16 mV/fC)
sng	neighbor (channel and chip) triggering enable
stot [0 1]	timing outputs control 1 (s6b must be disabled) <ul style="list-style-type: none"> stpp,stot[00,01,10,11]: TtP,ToT,PtP,PtT TtP: threshold-to-peak ToT: time-over-threshold PtP: pulse-at-peak (10ns) (not available with s10b) PtT: peak-to-threshold (not available with s10b)
sttt [0 1]	enables direct-output logic (both timing and s6b)
ssh [0 1]	enables sub-hysteresis discrimination
stc1,stc0 [00 01 10 11]	TAC slope adjustment (60, 100, 350, 650 ns)
sdt9-sdt0 [0:0 through 1:1]	coarse threshold DAC
sdp9-sdp0 [0:0 through 1:1]	test pulse DAC
sc010b,sc110b	10-bit ADC conv. time (increase subtracts 60 ns)
sc08b,sc18b	8-bit ADC conv. time (increase subtracts 60 ns)
sc06b, sc16b, sc26b	6-bit ADC conversion time
s8b	8-bit ADC conversion mode
s6b	enables 6-bit ADC (requires sttt enabled)
s10b	enables high resolution ADCs (10/8-bit ADC enable)
sdcks	dual clock edge serialized data enable
sdcka	dual clock edge serialized ART enable
sdck6b	dual clock edge serialized 6-bit enable
sdrv	tristates analog outputs with token, used in analog mode
stpp [0 1]	timing outputs control 2
slvs	enables direct output IOs
stcr	enables auto-reset (at the end of the ramp, if no stop occurs)
ssart	enables ART flag synchronization (trail to next trail)

Global bits (defaults are 0)	Description
s32	skips channels 16-47 and makes 15 and 48 neighbors
stlc	enables mild tail cancellation (when enabled, overrides sbip)
srec	enables fast recovery from high charge
sbip	enables bipolar shape
srat	enables timing ramp at threshold
sfrst	enables fast reset at 6-b completion
slvsbc	enable slvs 100 Ω termination on ckbc
slvstp	enable slvs 100 Ω termination on cktp
slvstk	enable slvs 100 Ω termination on cktk
slvsdt	enable slvs 100 Ω termination on ckdt
slvsart	enable slvs 100 Ω termination on ckart
slvstki	enable slvs 100 Ω termination on cktki
slvsena	enable slvs 100 Ω termination on ckena
slvs6b	enable slvs 100 Ω termination on ck6b
sL0enaV	disable mixed signal functions when L0 enabled
reset reset	Hard reset when both high
sL0ena	enable L0 core / reset core & gate clk if 0
l0offset_i0:11	L0 BC offset
offset_i0:11	Channel tagging BC offset
rollover_i0:11	Channel tagging BC rollover
window_i0:2	Size of trigger window
truncate_i0:5	Max hits per L0
nskip_i0:6	Number of L0 triggers to skip on overflow
sL0cktest	enable clocks when L0 core disabled (test)
sL0ckinv	invert BCCLK
sL0dckinv	invert DCK
nskipm_i	magic number on BCID - 0xFE8
slh, slxh	increases bias current at input node from nominal 1nA to 15nA or 300nA respectively
stgc	extreme charge handling compensation

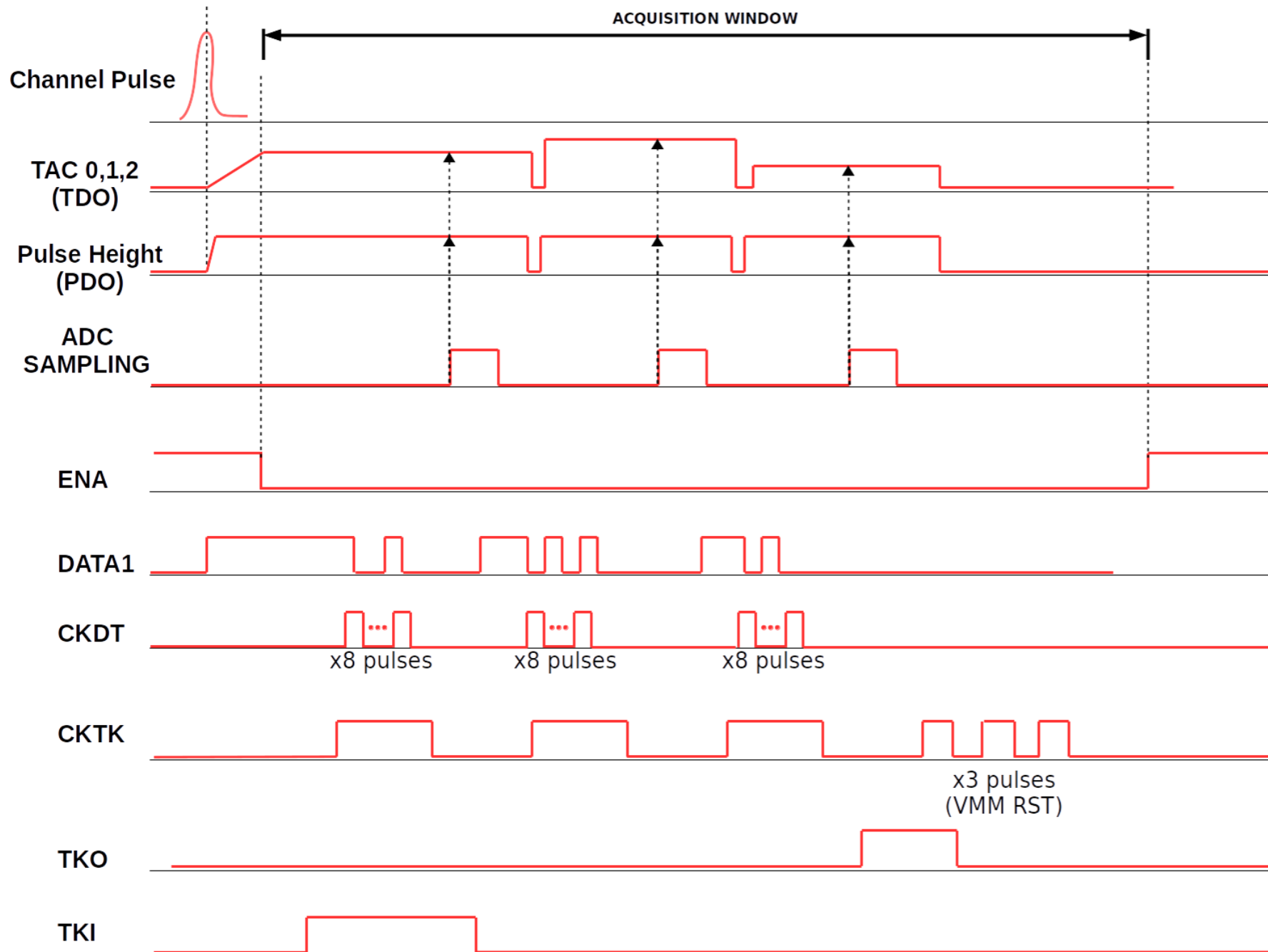


VMM Connectivity on the BGA (NSW)

Name, Position	Con-nection	In, Out or I/O	Type of Signal or Max/Min	Description
sett A19-20	VMM	I/O	Custom LVDS Bi-directional	Channel 0 force-neighbor signal
setb Y19-20	VMM	I/O	Custom LVDS Bi-directional	Channel 63 force-neighbor signal
ckbc (BCclk) C15-16	ROC	In	SLVS	Bunch crossing clock of 40 MHz / External trigger signal
cktp (Test Pulse) B15-16	ROC	In	SLVS	Test pulse clock
cktk (Level-0) D13-14	ROC	In	SLVS	Token clock / L0 (digital NSW mode)
ckdt (ROclk) E15-16	ROC	In	SLVS	Data clock
kart (ARTclk) D19-20	ROC	In	SLVS	ART clock
sdi B17	SCA	In	CMOS	Configuration data input
sdo B18		Out	CMOS	Configuration data output (not used, HiZ state in NSW)
cs B19	SCA	In	CMOS	Chip Select, active low
sck B20	SCA	In	CMOS	Input SPI clock
t0-t63 E17-W20	TDS	Out	SLVS	Direct digital outputs
mo C9	SCA	Out	0-1 V	Analog output for calibration
tki (BCR/OCR) C13-14	ROC	In	SLVS	Token input (an. mode) / (BCR- OCR) / acceptance window in non-L0 cont. mode
tko B13-14		Out	SLVS	Token output (analog mode, not used in NSW)
ena (ENA/Soft Reset) C17-18	ROC	In	SLVS	Acquisition start/stop
ck6b C19-20	TDS	In	SLVS	6-bit ADC Clock
art E13-14	ART2GBT	Out	SLVS	Address in Real Time
data0 D15-16	ROC	Out	SLVS	data line
data1 D17-18	ROC	Out	SLVS	data line first bit, flag in cont.

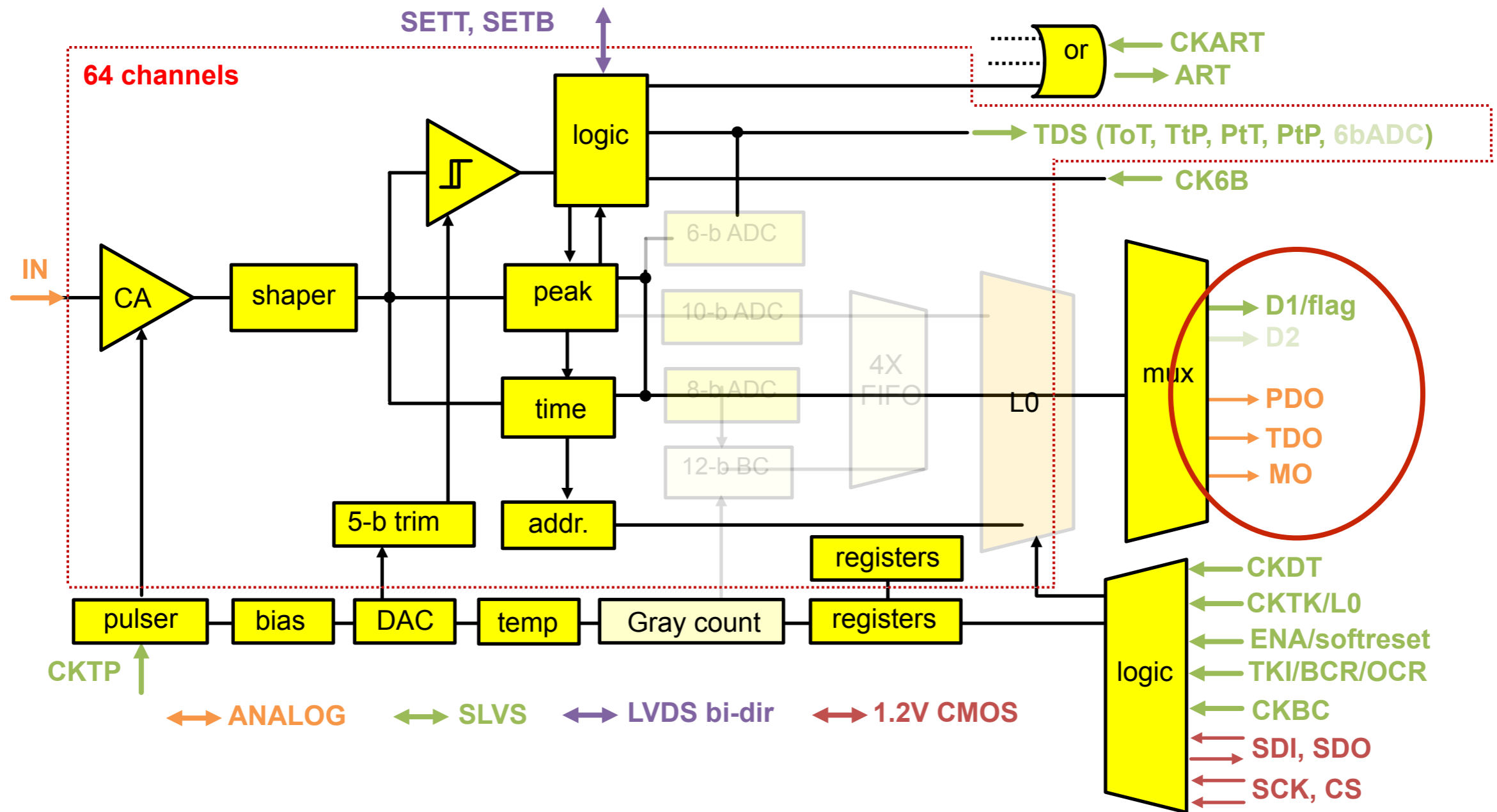


- In two-phase (analog) mode which is the mode originally implemented in the VMM1, the ASIC operates in two separate phases: **acquisition** and **readout** - During the acquisition phase the events are processed and stored in the **analog memories** of the **peak and time detectors**. As soon as a first event is processed, a flag is raised at the digital output.

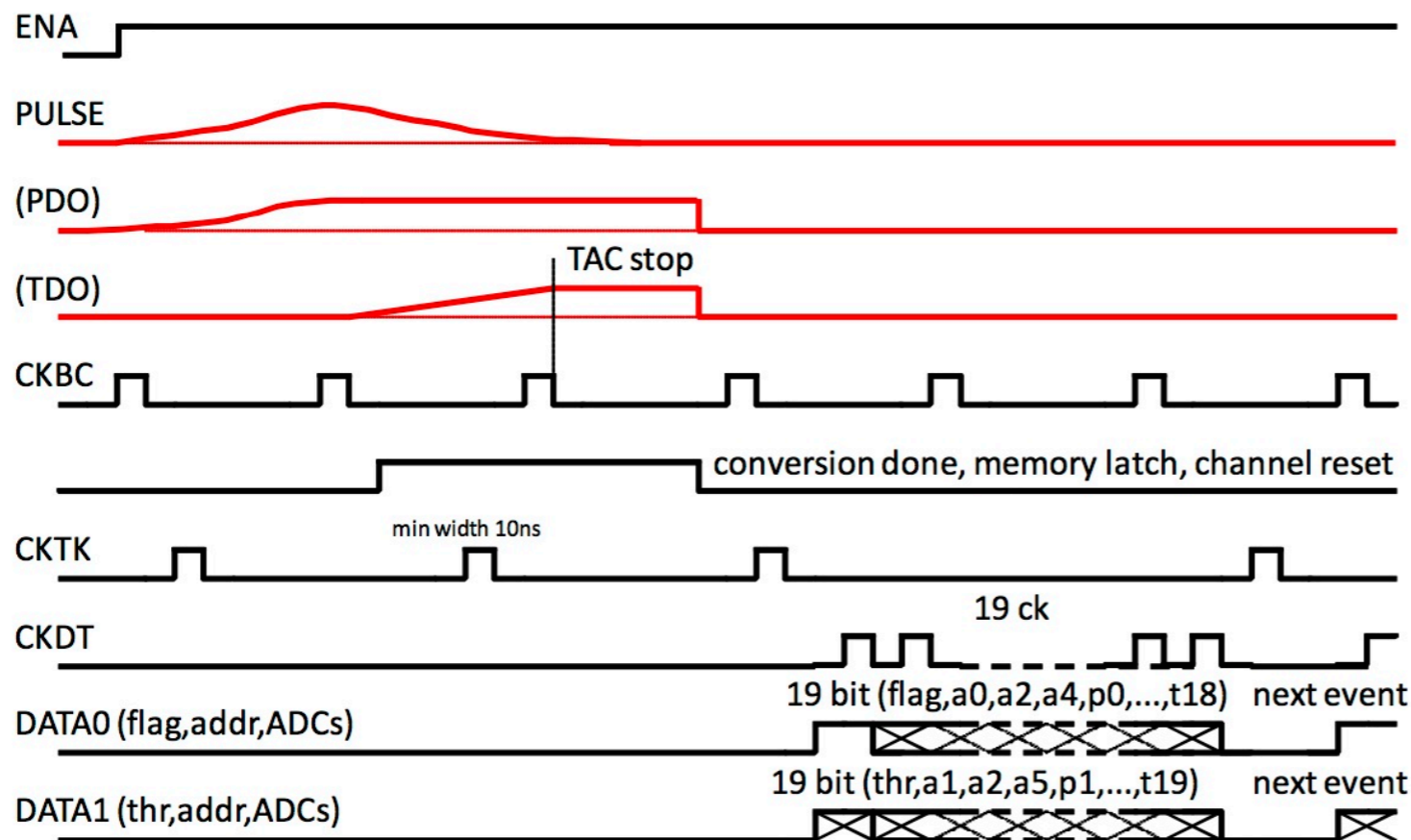


- Once the process is complete the **ASIC can be switched readout phase**. The first set of amplitude and time voltages is made available at the **analog outputs**. The **address of the channel is serialised** and made available at the digital output using **six data clocks**.

- In this mode all analog buffers are multiplexed in the analog outputs
- Lengthy operation since each analog signal needs to be sampled while the address is read out serially

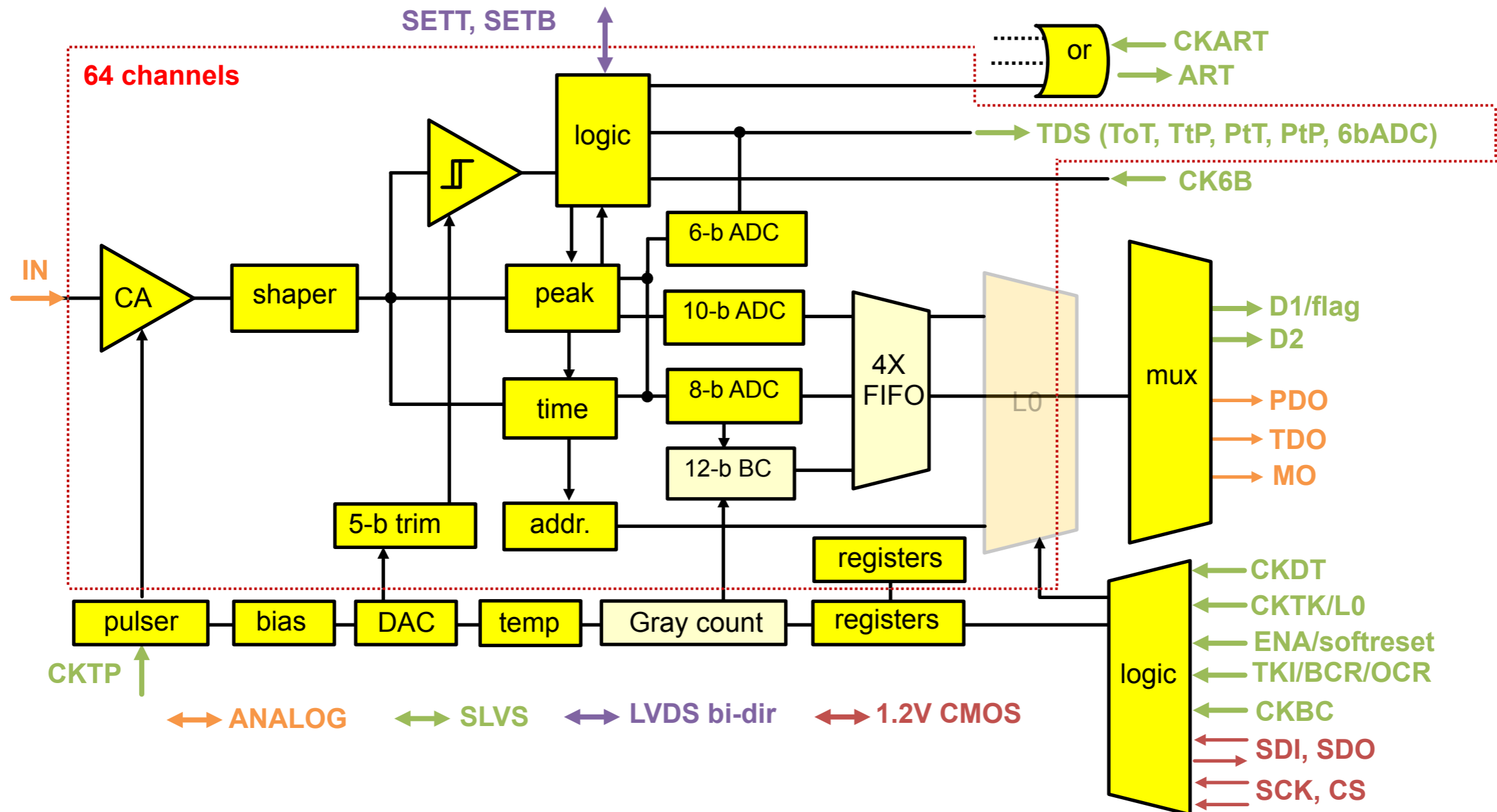


- In this mode the peak and time detectors convert the voltages into currents that are routed to the **6/10-bit ADC and 8-bit ADC respectively**. The 10-bit ADC provides a high resolution A/D conversion of the peak amplitude in a conversion time of about 200 ns. The 8-bit ADC provides the A/D conversion of the timing (measured using the TAC) from the time of the peak or the threshold to a stop signal. The counter value at the TAC stop time is latched into a local 12-bit memory. In the continuous mode the 64 channel digital outputs are available as well providing time-over-threshold (ToT), threshold-to-peak (TtP), peak-to-threshold (PtT), or a 10 ns pulse occurring at peak (PtP) or the 6-bit ADC. The channel self resets at the end of the timing pulse, thus providing continuous and independent operation of all 64-channels.

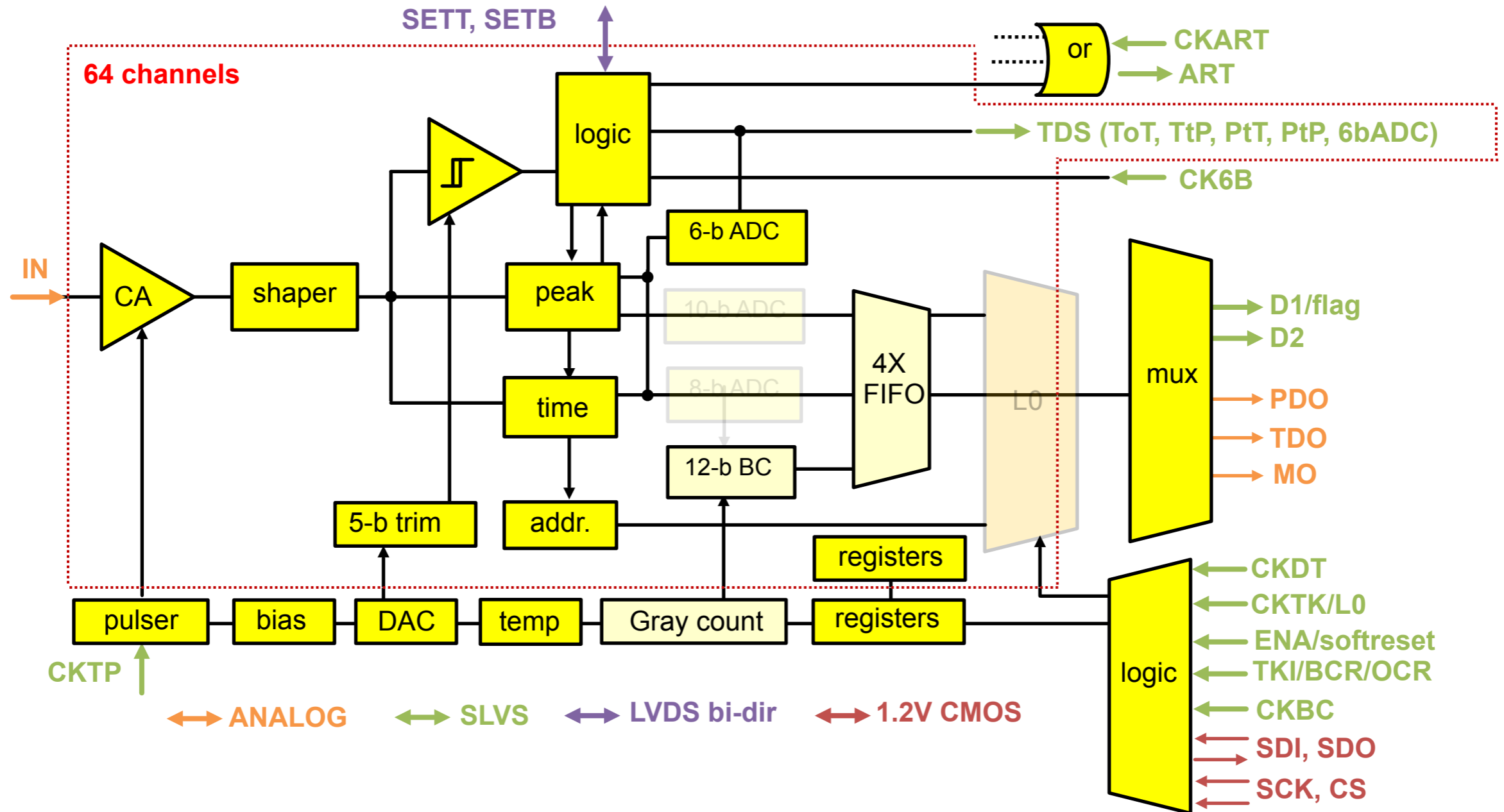


flg	1-bit
thr	1-bit
addr	6-bit
ampl	10-bit
time	8-bit
BC	12-bit

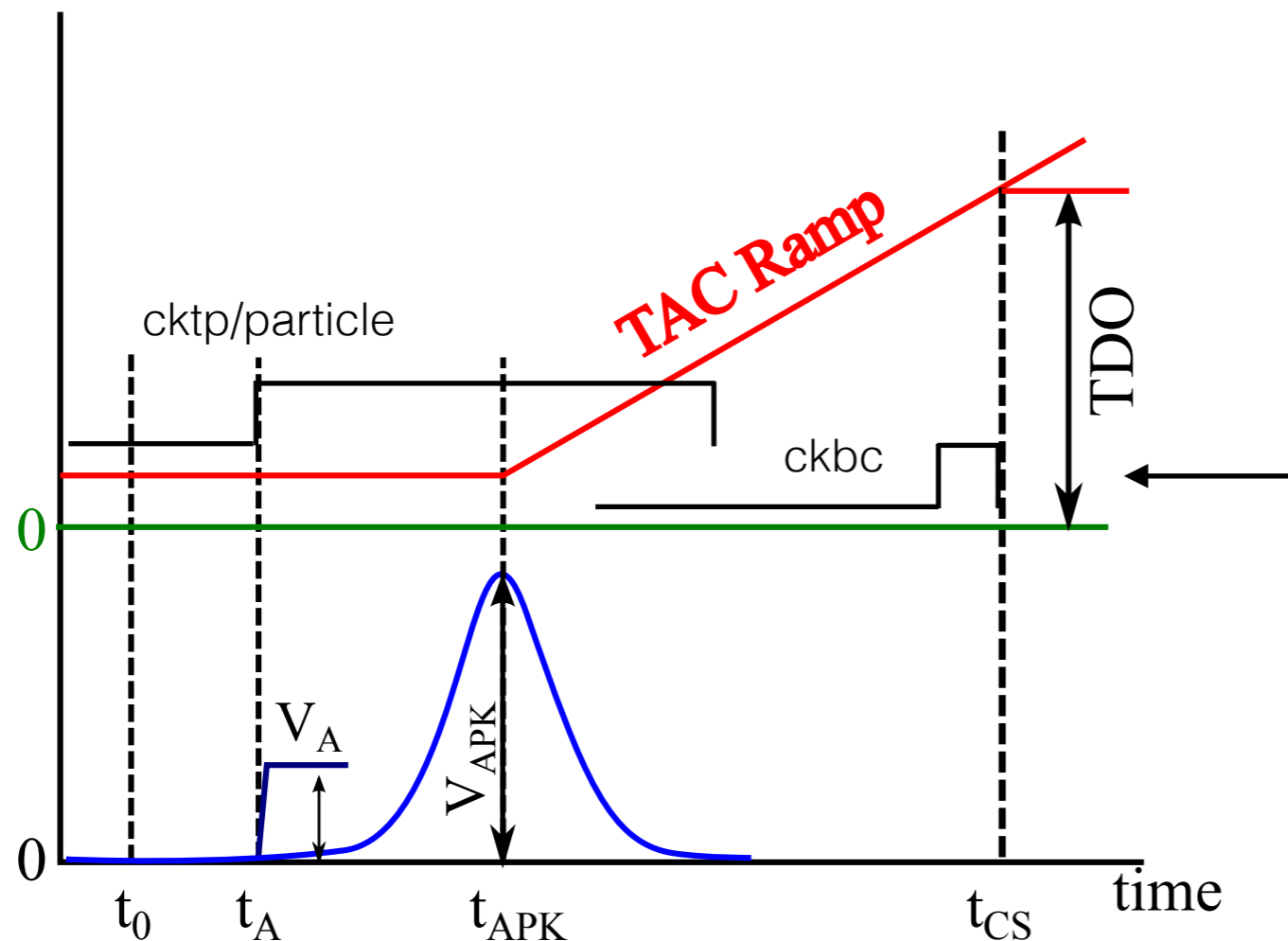
- This mode provides continuous trigger-less readout
- All the outputs and inputs are active
- 6bit ADC conversion within 25ns (configurable)
- 10bit ADC conversion at 200ns (configurable) ← Leading deadtime per channel
- 8bit ADC conversion at 100ns (configurable)



- This mode provides continuous trigger-less readout
- All the outputs and inputs are active
- 6bit ADC conversion within 25ns (configurable) ← Enabling sfrst, channel resets after 6bit conversion - 40ns deadtime
- 10bit ADC conversion at 200ns (configurable) ← ~~Leading deadtime per channel~~
- 8bit ADC conversion at 100ns (configurable)

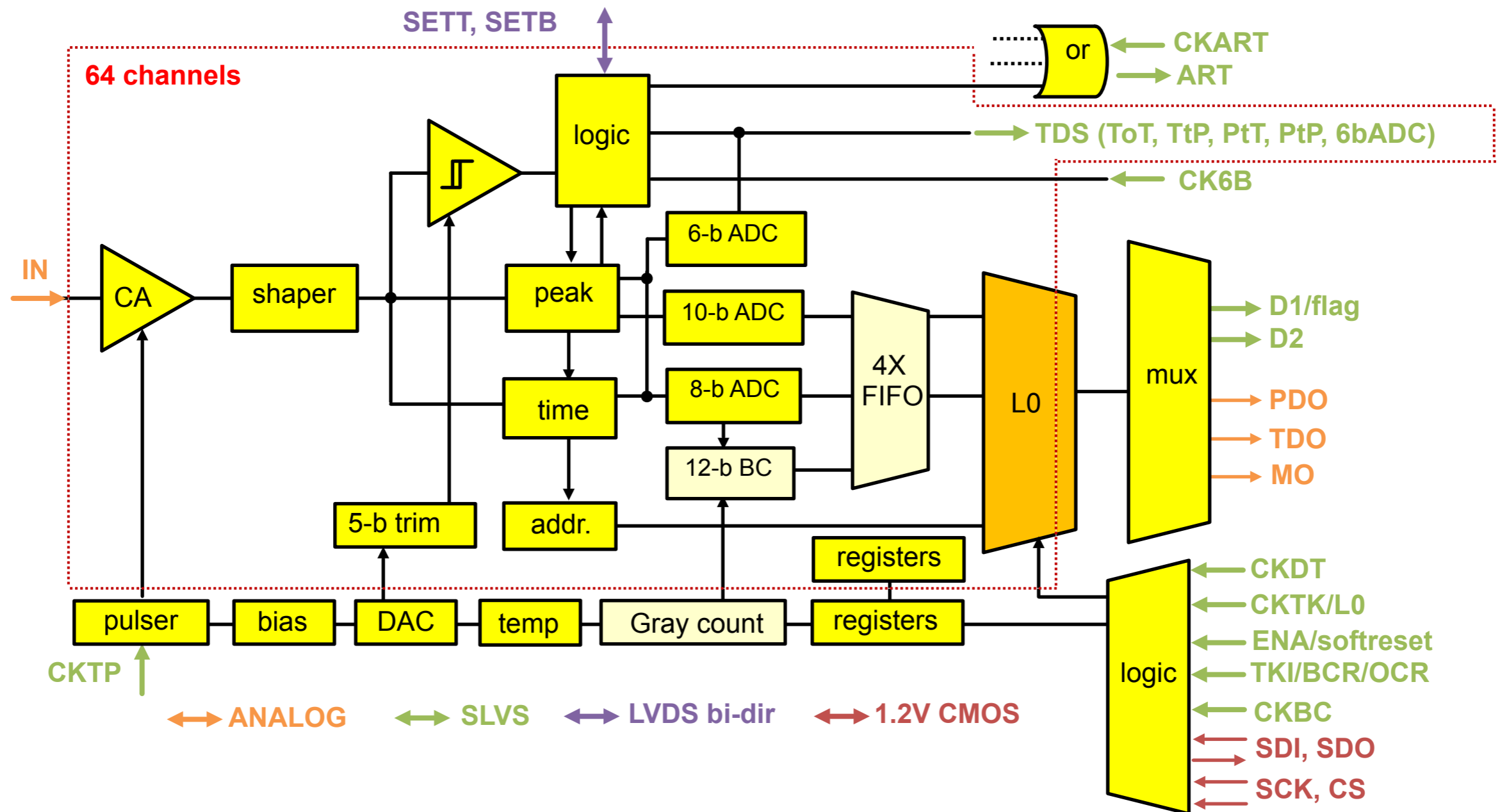


- VMM design targets **synchronous** machines hence can be difficult to use in an environment like a **test beam where asynchronous operation** is needed but **precise timing** is needed to be measured (drift time)
- Most chips designed for synchronous machine **suffer from time jitter in such environment**
- On **VMM a mode was foreseen** to do such measurement where the **ckbc** can be used as a **strobe** and **not like a real clock**
- It can be **send as a trigger signal** with a **fixed latency** achieving precise time measurements

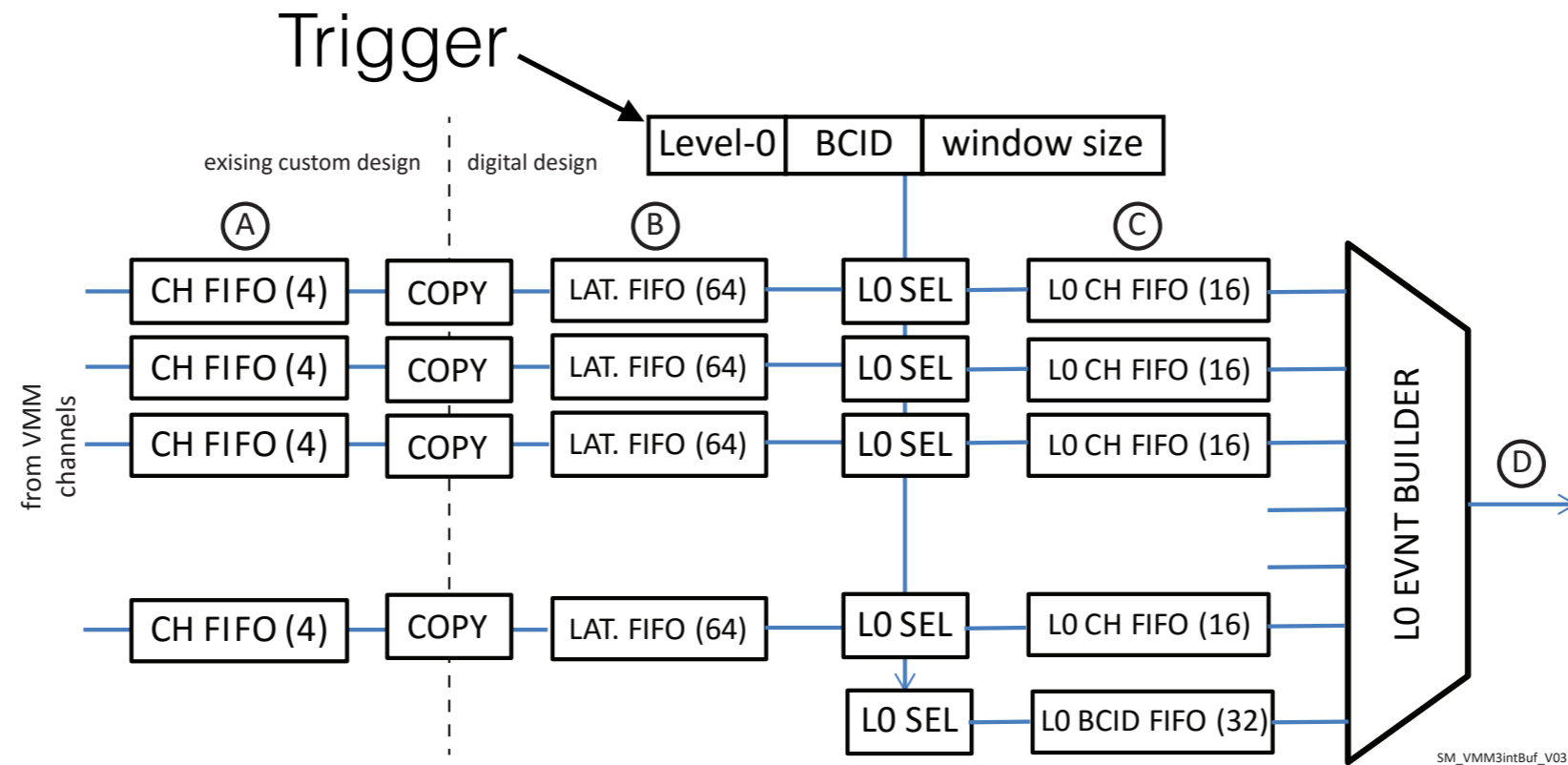


- Trigger signal from **external source**
- Can be combined with **register stcr** where **channel resets** if stop signal not occurs within the **TAC ramp**
- **Implies that trigger is propagated** within the TAC ramp up time (**60ns-650ns**)
- The **longer the TAC** though the **lower the resolution** on 8-bit information from the ADC
- Highly correlated trigger readout and noise subtraction

- The signal processing is done in the same way but the readout is different.
- This is an externally triggered operation for synchronous machines



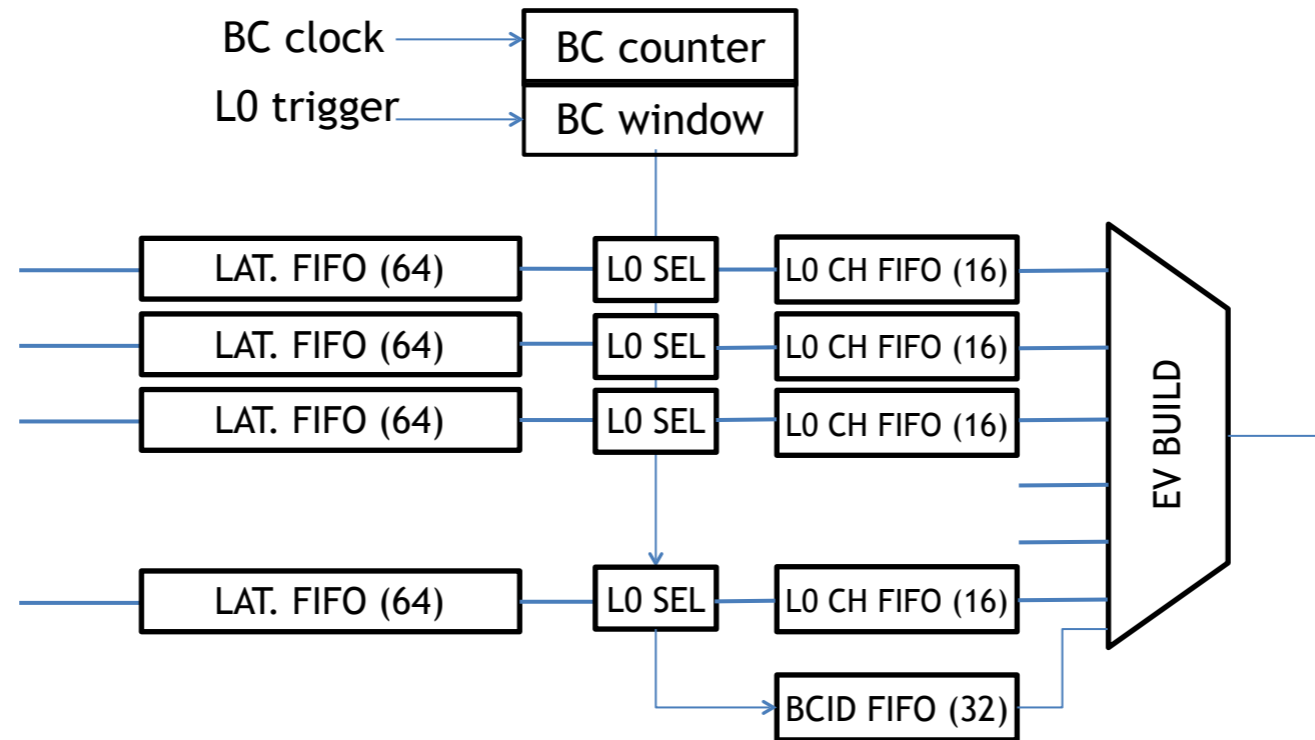
- Each channel has a **Level-0 Selector** circuit which is **connected to the output** of the channel's latency FIFO.
- The **selector finds events within the BCID window** (maximum size of 8 BC clocks) of a Level-0 Accept and **copies them to the L0 Ch FIFO**. The data are available in the output which is running on **IDLE K28.5** in two data lines and can be readout **DDR at a speed of 640Mbps** (160MHz clock tested, effective bandwidth 560Mbps due to 8b/10b encoding).

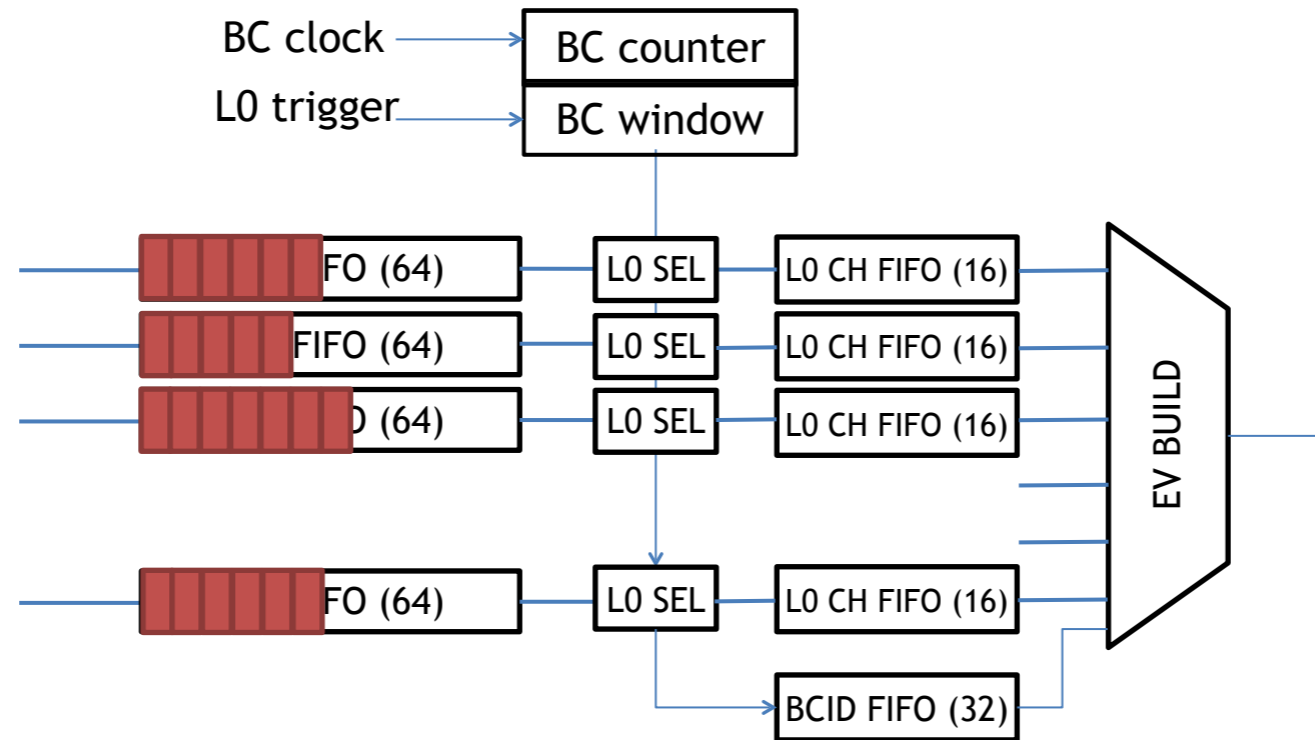


- VMM will build the event with common BCID +relative for each hit
- Header is sent out once no data found

	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26	27	28	29	30	31
header	V	P	orb		BCID (12)												1st word after comma															
hit data	1	P	R	T	Chan# (6)						ADC (10)						TDC (8)				N	rel BCID										

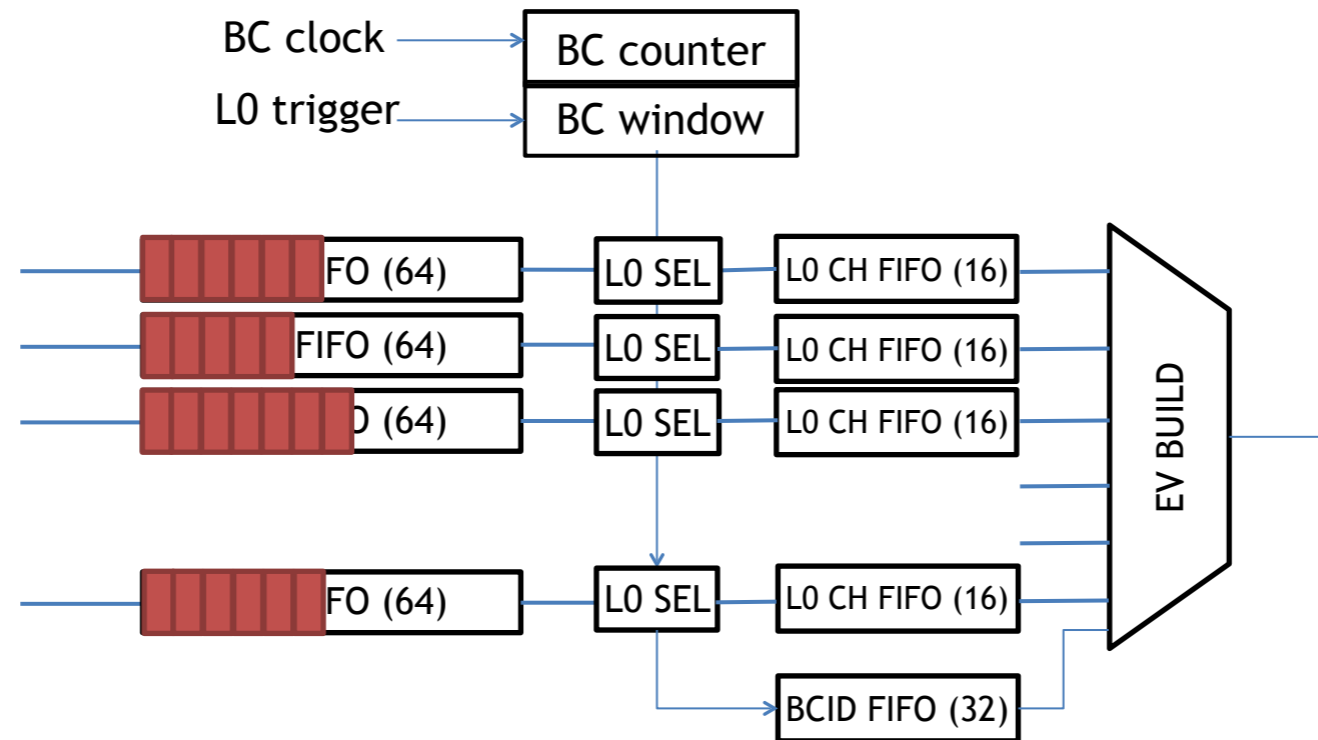
LL_format_VMM3out_V04





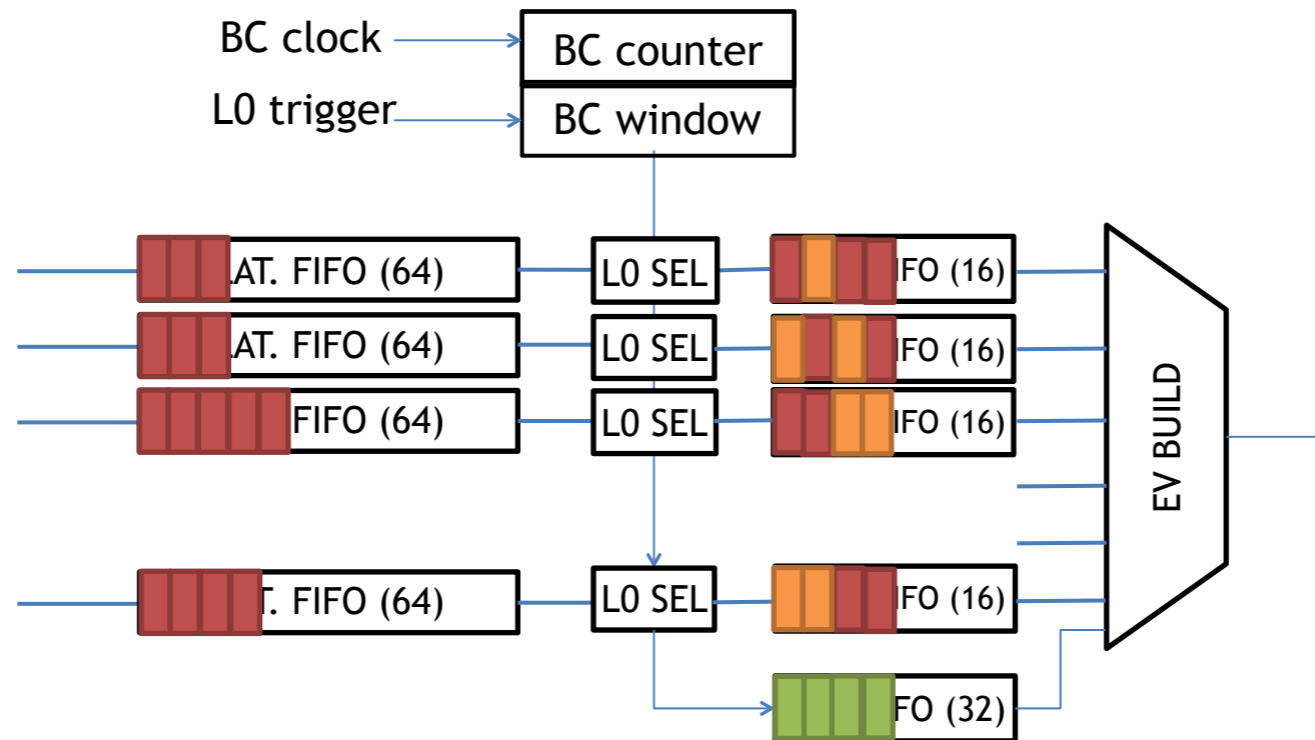
Latency FIFO takes data from the mixed-signal front-end

- FIFO designed to accommodate 4 MHz data in a 10 μ s latency window
- 20-bit data: threshold, amplitude (ADC), timing (ADC)



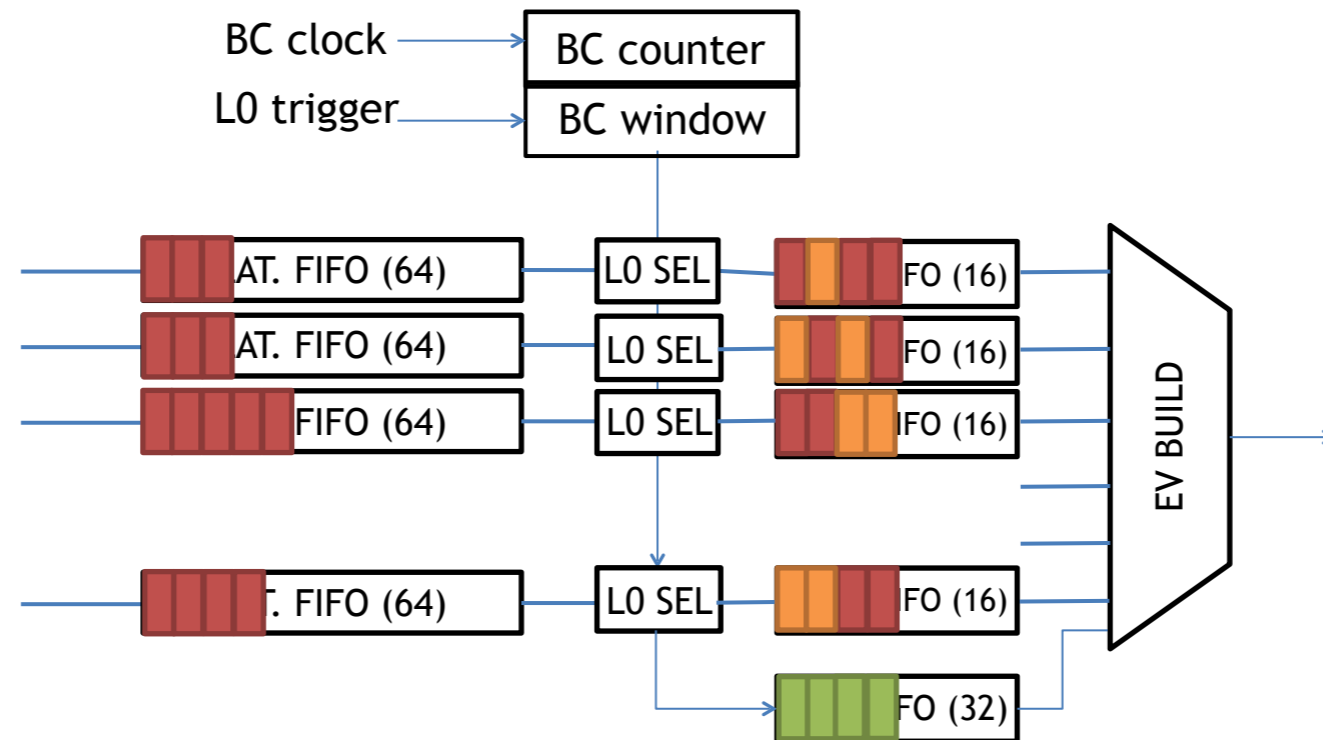
At L0 trigger builds BC trigger window and selects data for the L0 CH FIFO

- flushes old data
- fills non-valid data as needed (for simultaneous overflow)
- builds BCID FIFO



At **L0 trigger** builds BC trigger window and **selects data** for the L0 CH FIFO

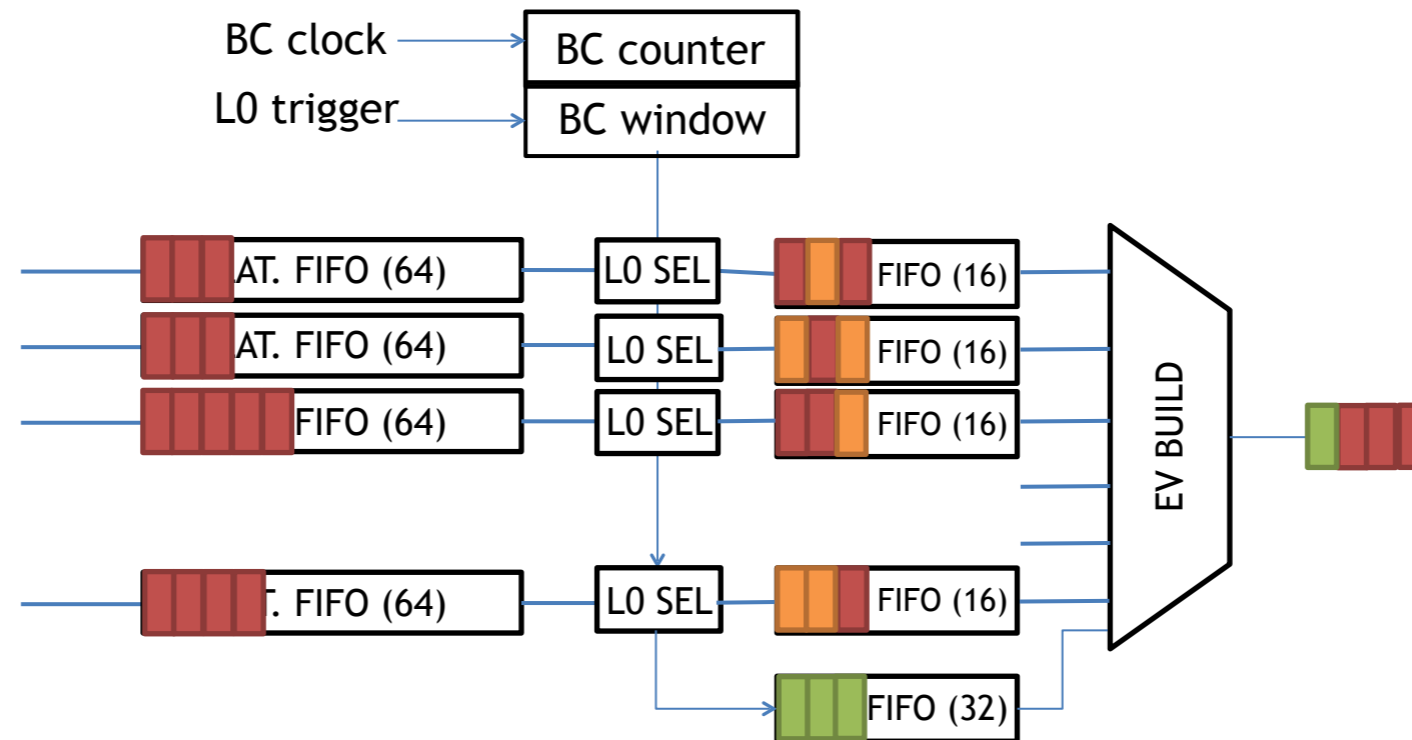
- flushes old data
- fills non-valid data as needed (for simultaneous overflow)
- builds BCID FIFO



Builds event

- BCID followed by valid data with address
- header
- event built in $< 1\mu\text{s}$

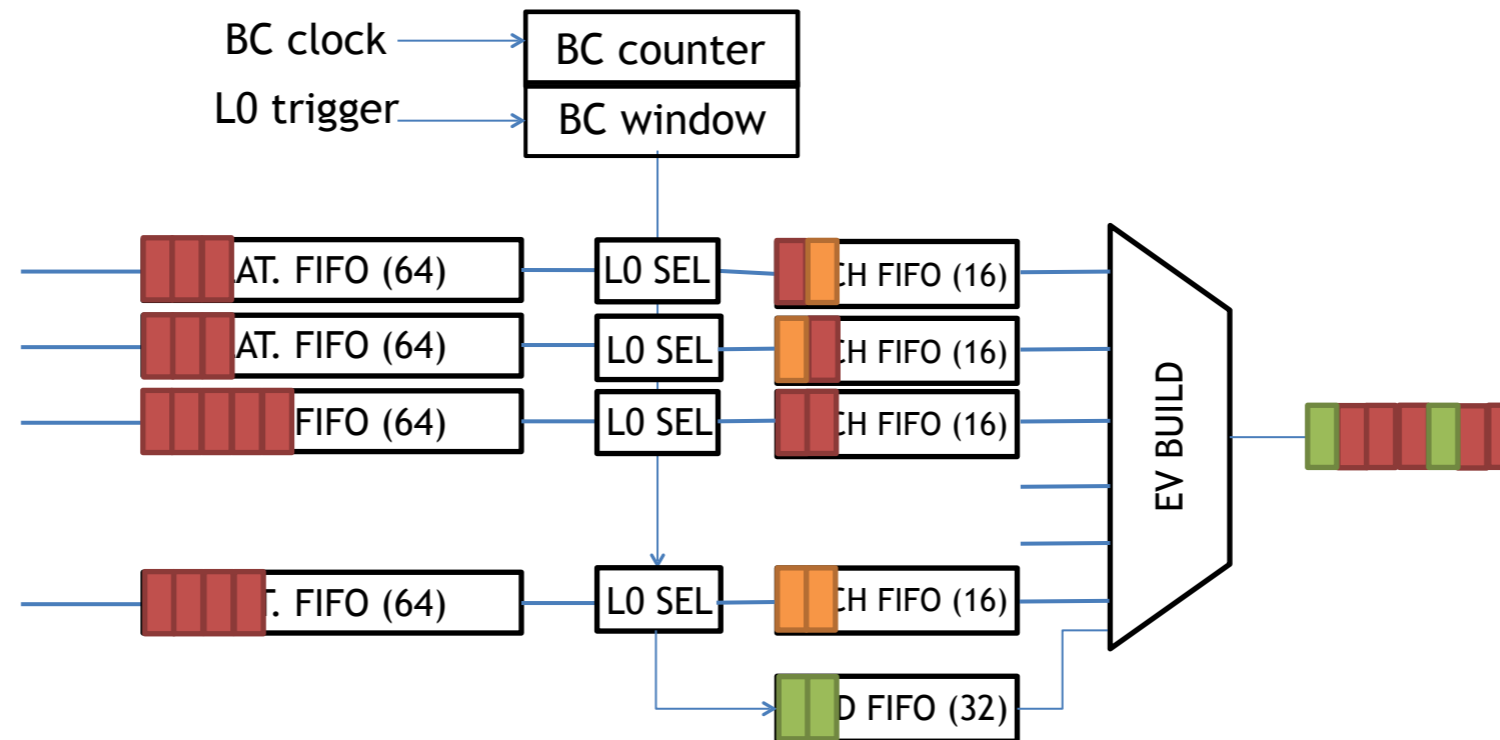
Sends data through two data links (DDR, 640MB/s)



Builds event

- BCID followed by valid data with address
- header
- event built in $< 1\mu\text{s}$

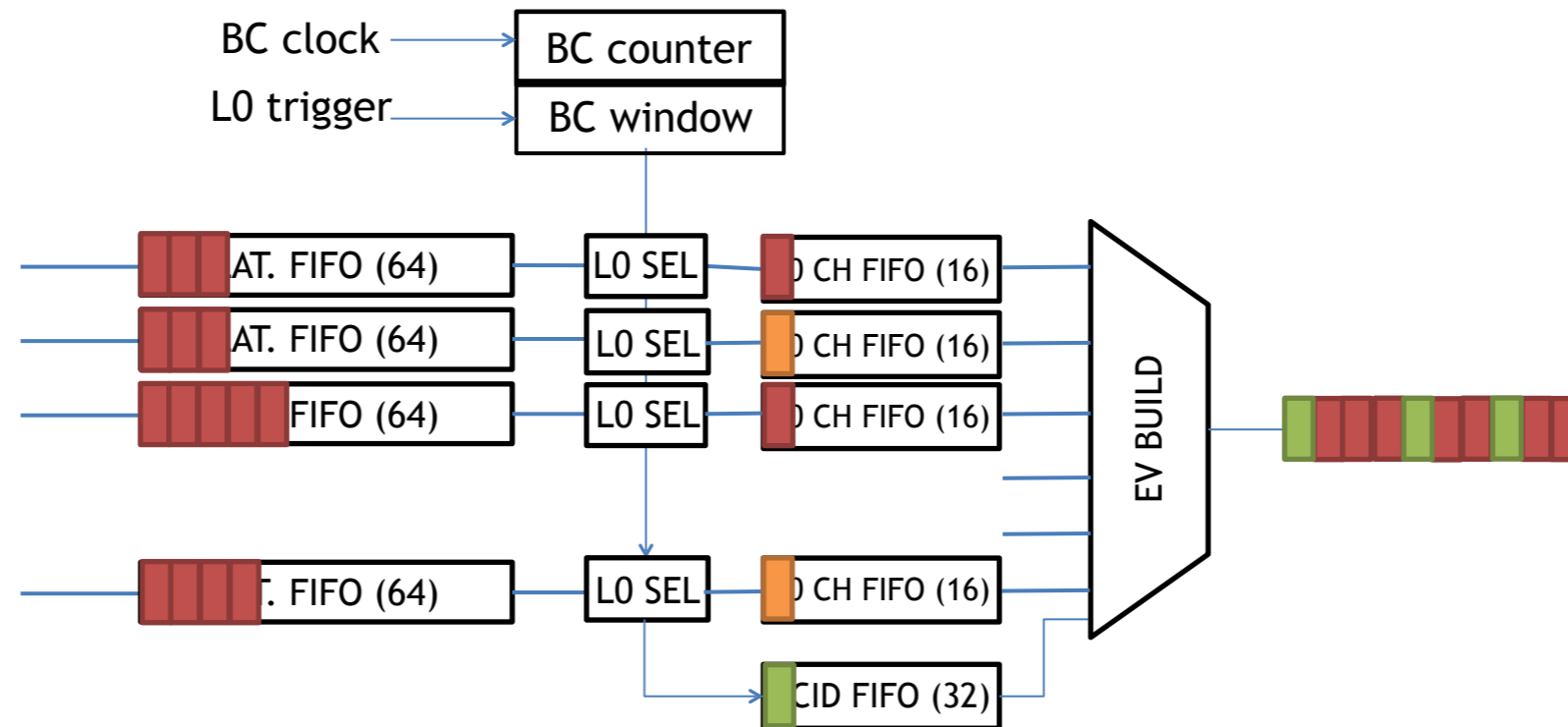
Sends data through two data links (DDR, 640MB/s)



Builds event

- BCID followed by valid data with address
- header
- event built in $< 1\mu\text{s}$

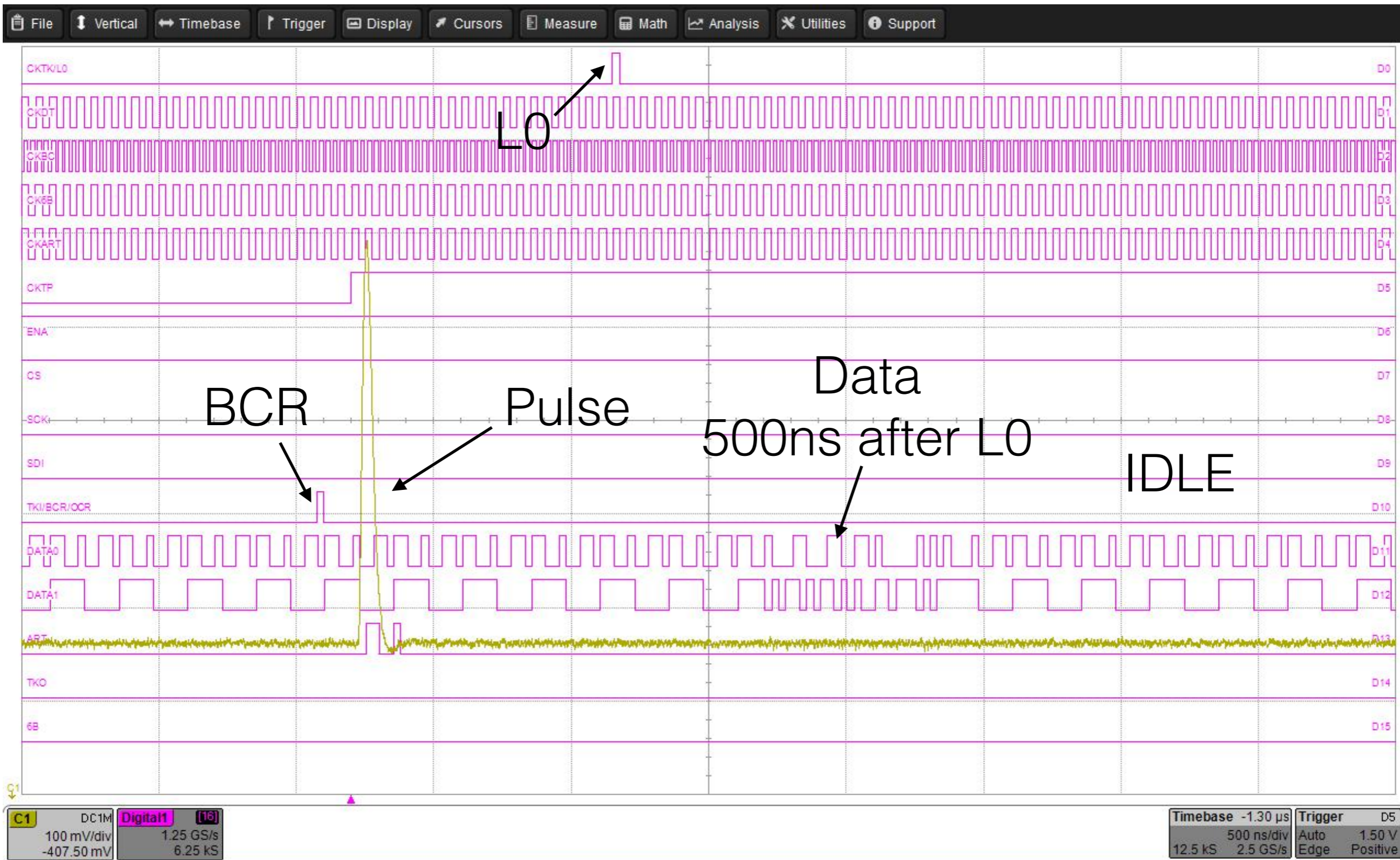
Sends data through two data links (DDR, 640MB/s)



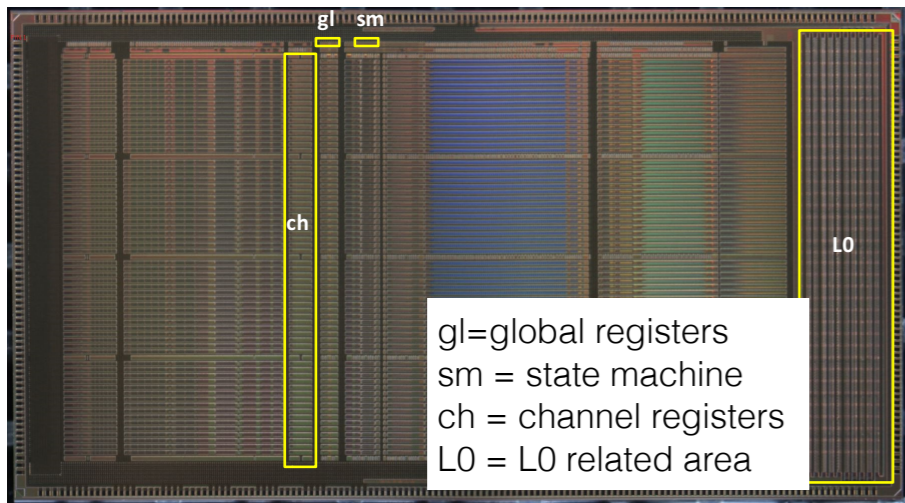
Builds event

- BCID followed by valid data with address
- header
- event built in $< 1\mu\text{s}$

Sends data through two data links (DDR, 640MB/s)



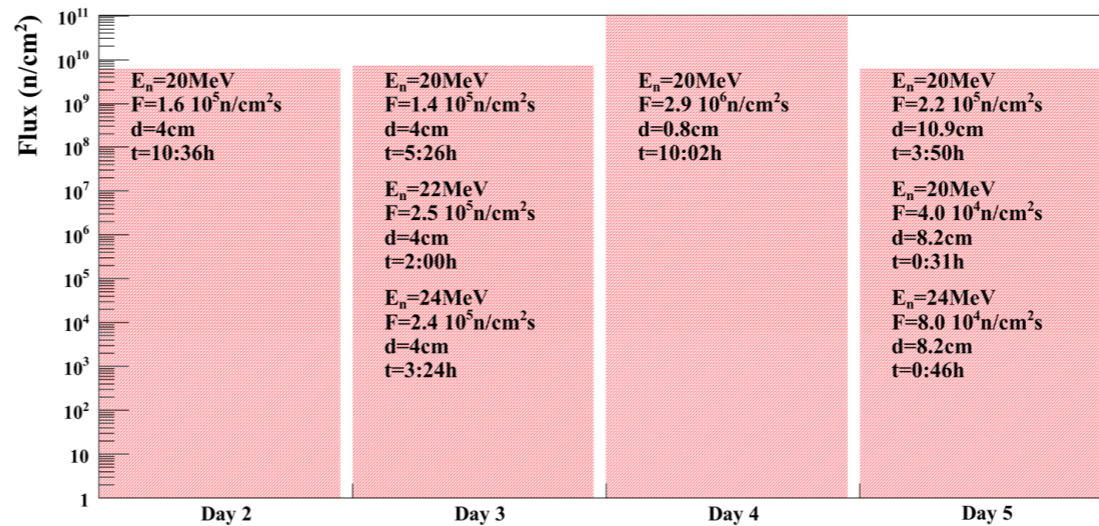
- In the VMM3a there are **three types of storage elements** that require SEU protection, the **configuration registers**, the **state machine** control logic and the **L0 logic**
- To mitigate for SEU two techniques are used:
 - **Dual Interlocked Cells (DICE)** for the protection of the configuration registers
 - **Triple Modular Redundancy (TMR)** for the state machines and the L0 Logic blocks
- L0 Data
 - Single-bit faults on data are flagged by a **parity bit**
 - The parity is registered in the FIFOs and transmitted outside
- In both neutron irradiation and ^{60}Co irradiation **all SEU were corrected by the protection**



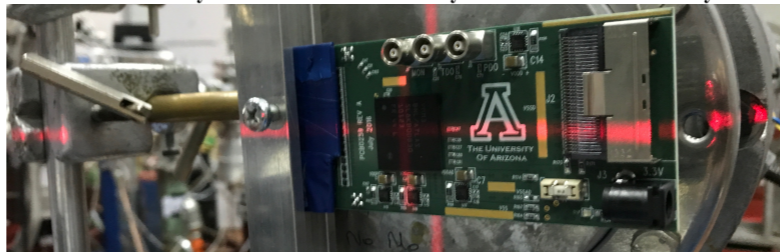
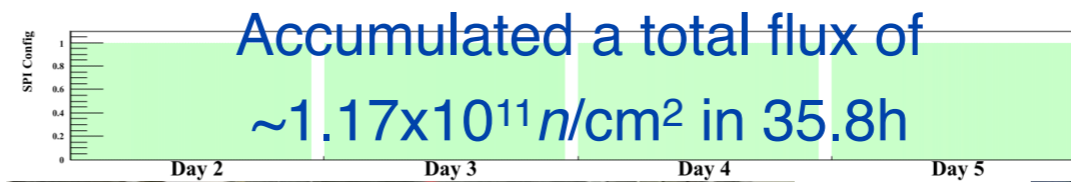
gl=global registers
 sm = state machine
 ch = channel registers
 L0 = L0 related area

L0 block protection

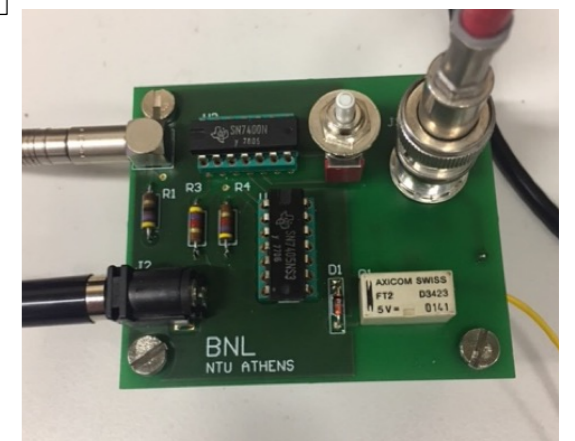
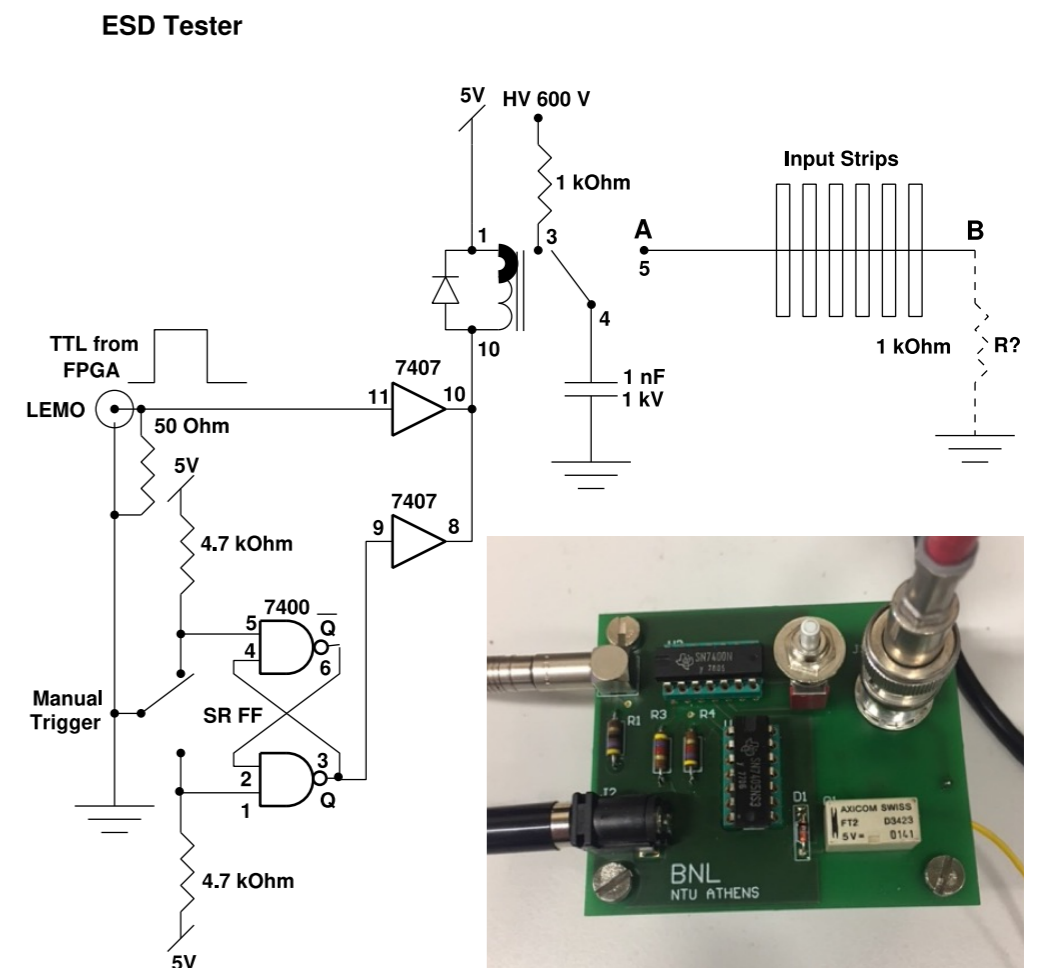
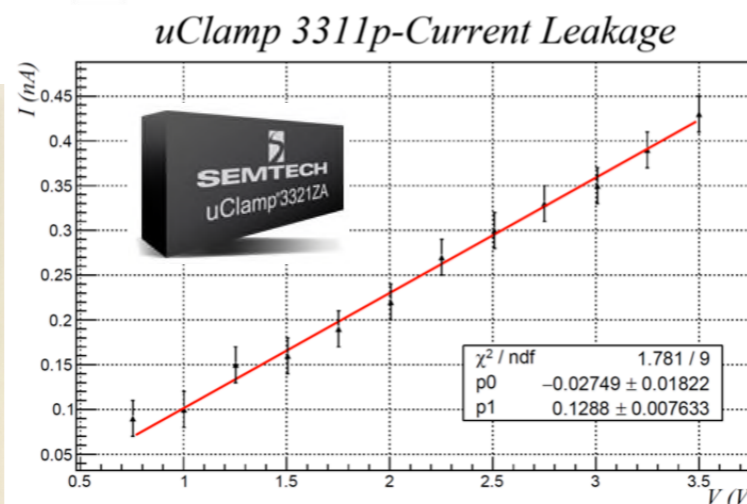
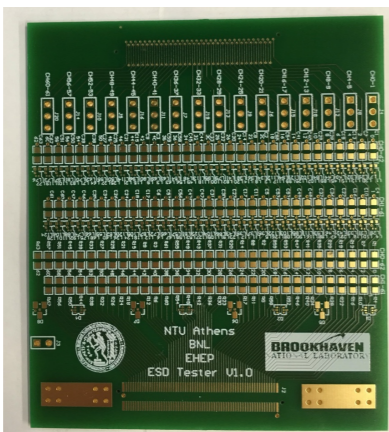
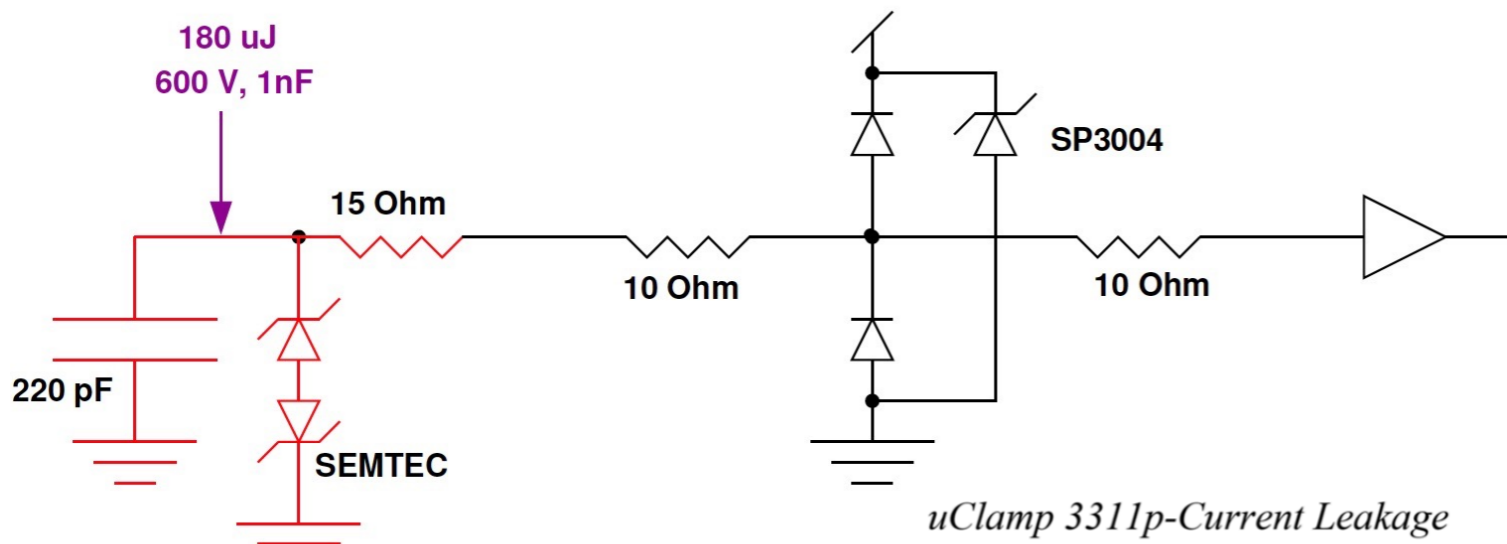
Block	Method
BC counter	TMR
Latency FIFO CTRL	Parity on FIFO pointer, FIFO resets if parity error
L0 FIFOs Control	TMR
Event Builder	TMR
LOA register/Nskip circuit	TMR



4 VMM3a were irradiated at the ^{60}Co source at BNL up to 1MRad



- Since the VMM2, we have experience major channel (initial NUP4114 issue). Moving to 130nm technology made the requirements on input protection higher. Current protection scheme based on the SP3004 seems inadequate to protect the VMM front ends
- A dedicated ESD testing procedure was lunched allowing a systematic test of the VMM input.
- A VMM board (MMFE1) with Panasonic connector and a VMM socket was developed to perform systematic tests. On top a Panasonic based connector daughter-board was built to test different protection schemes and different footprints.
- 220 pF capacitor emulates typical MM strip capacitance, a channel like this survived repeated discharges while without protection is dead after a single discharge. Then survived zapping overnight (>30,000 discharges)



The IBM CMOS8RF Design Manual specifies the operating temperature range to be from -55°C to 125°C . However device life time degrades rapidly at high temperatures. The case temperature should be kept below 50°C and preferably in the range $30\text{--}40$ and should be verified and compared to the junction temperature provided by the VMM ASIC. The VMM includes a temperature sensor which can be read out by appropriately programming the monitor output and digitized by the SCA setting (in configuration mode) $\text{scmx} = 0$, $\text{sm5--sm0} = 000100$ (see Table 6). The die temperature is approximately given by:

$$^{\circ}\text{C} = \frac{725 - V_{\text{sensor}}}{1.85}$$

where V_{sensor} is the temperature sensor reading in mV. The case temperature of a single-chip

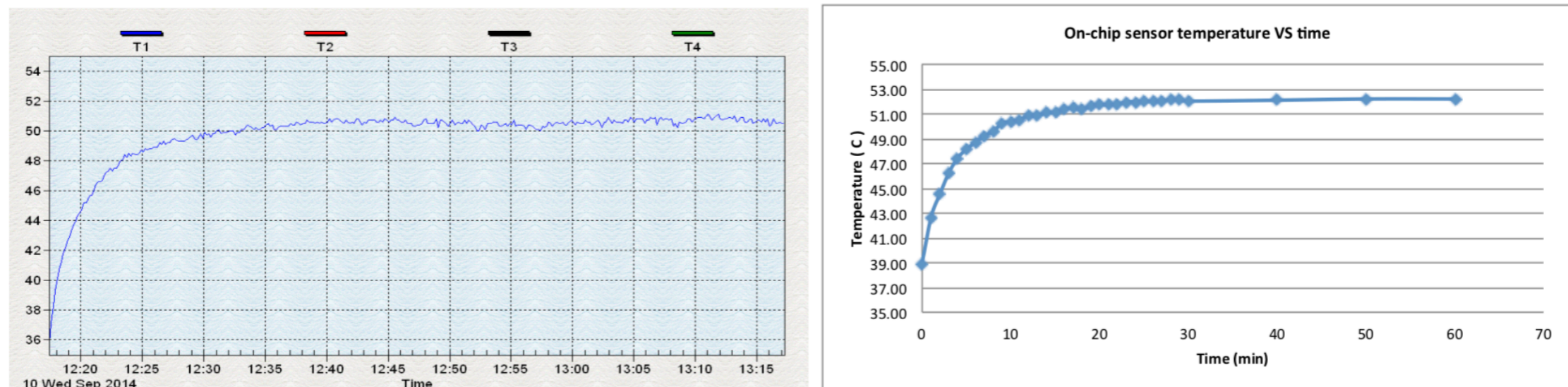
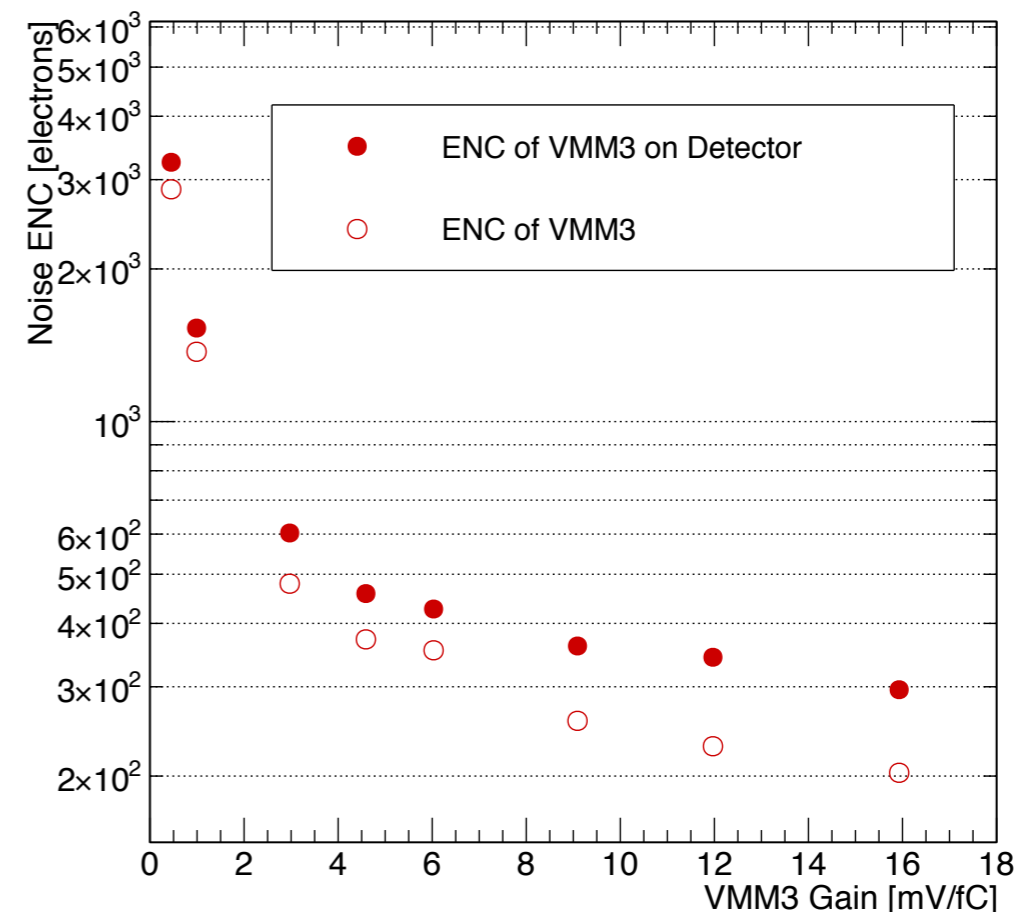
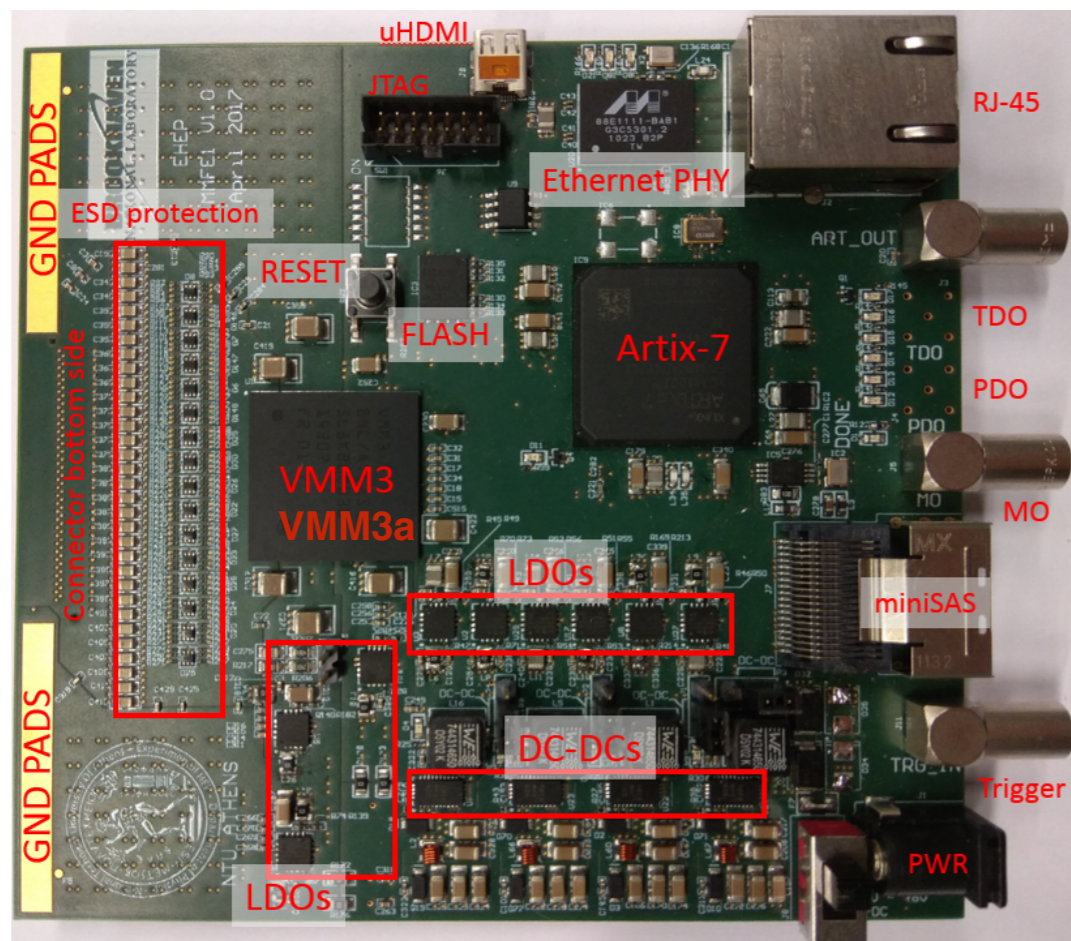


Figure 8: Left plot, the VMM case temperature. Right plot, the junction temperature as a function of time after turn-ON



- **Setup** of of 2x MMFE1s on 2x Resistive Micromegas chambers (Ar+7%CO₂ 400μm pitch, 5mm drift)
- Custom made **firmware** and **software** was developed allowing to **trigger** with scintillator system
 - Mode to control the CKBC externally
- **High data rate** ~20KHz/channel (VMM can reach 4MHz), arrived at the limit of Gbps UDP connection
- **Noise levels** of 300 e⁻ ENC at gain 9mV/fC, 200ns



- For the test-beams, a **DAQ + control software** was developed allowing operations like configuration and calibration. Highly configurable, multi-threaded and reliable (VMM electronics)
- The applications is developed in Qt and C++ based on a UDP handshake protocol

Run Control →

Communication →

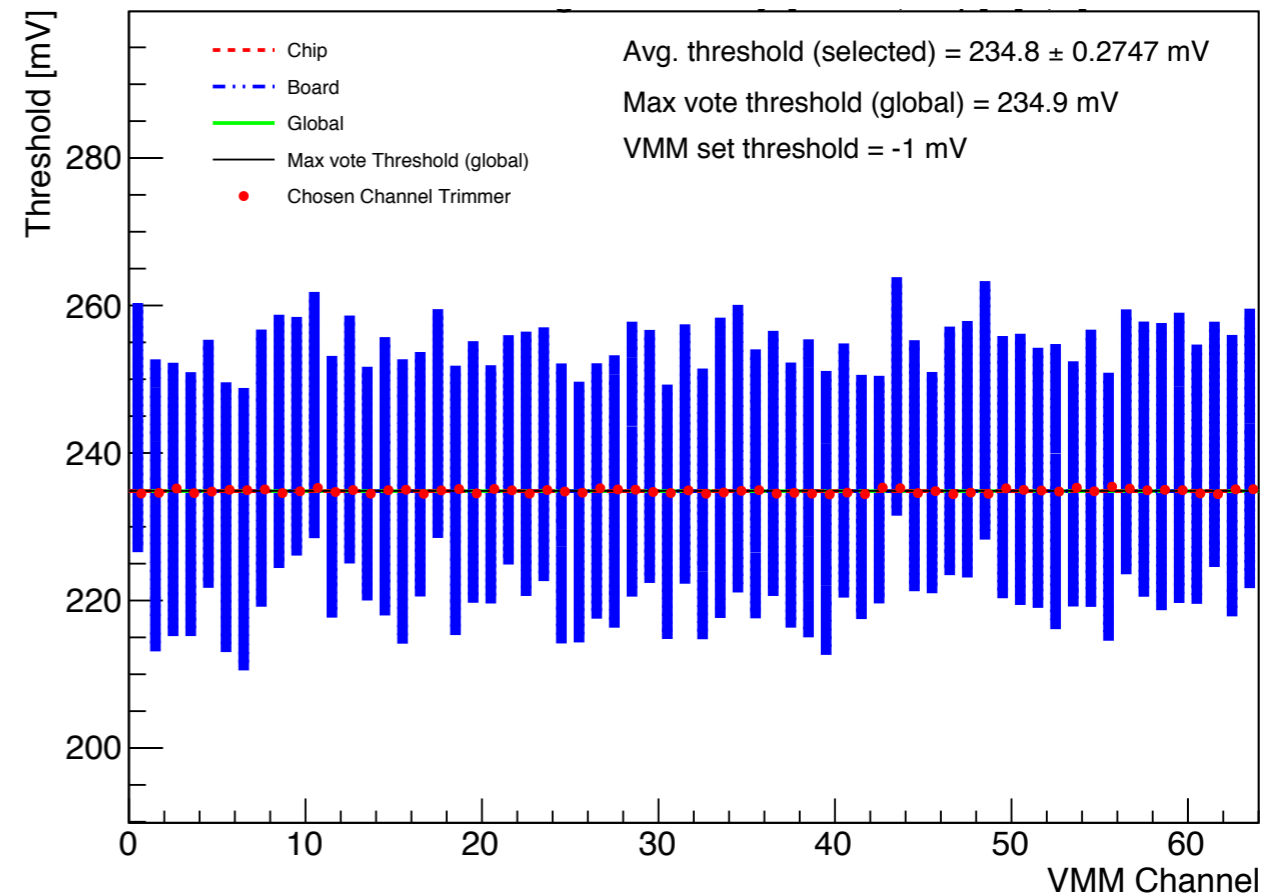
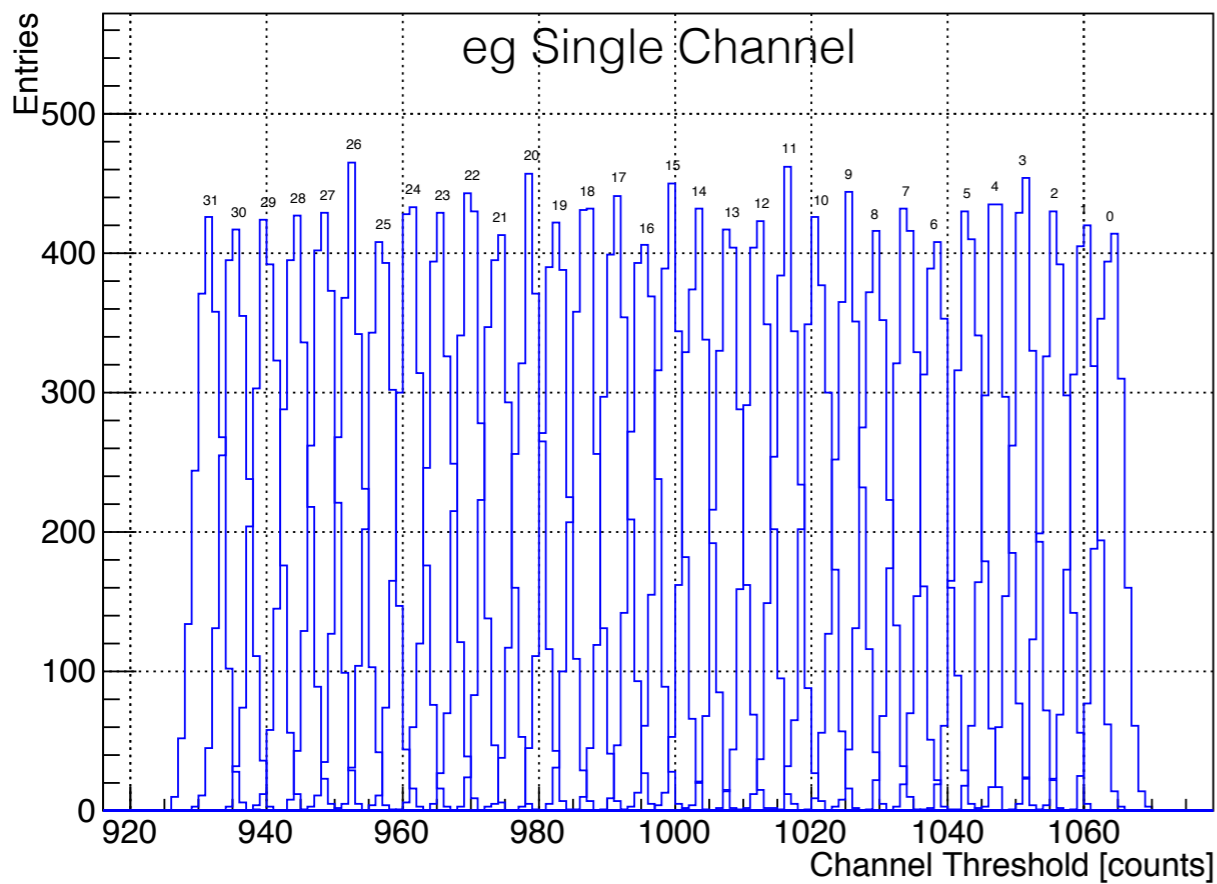
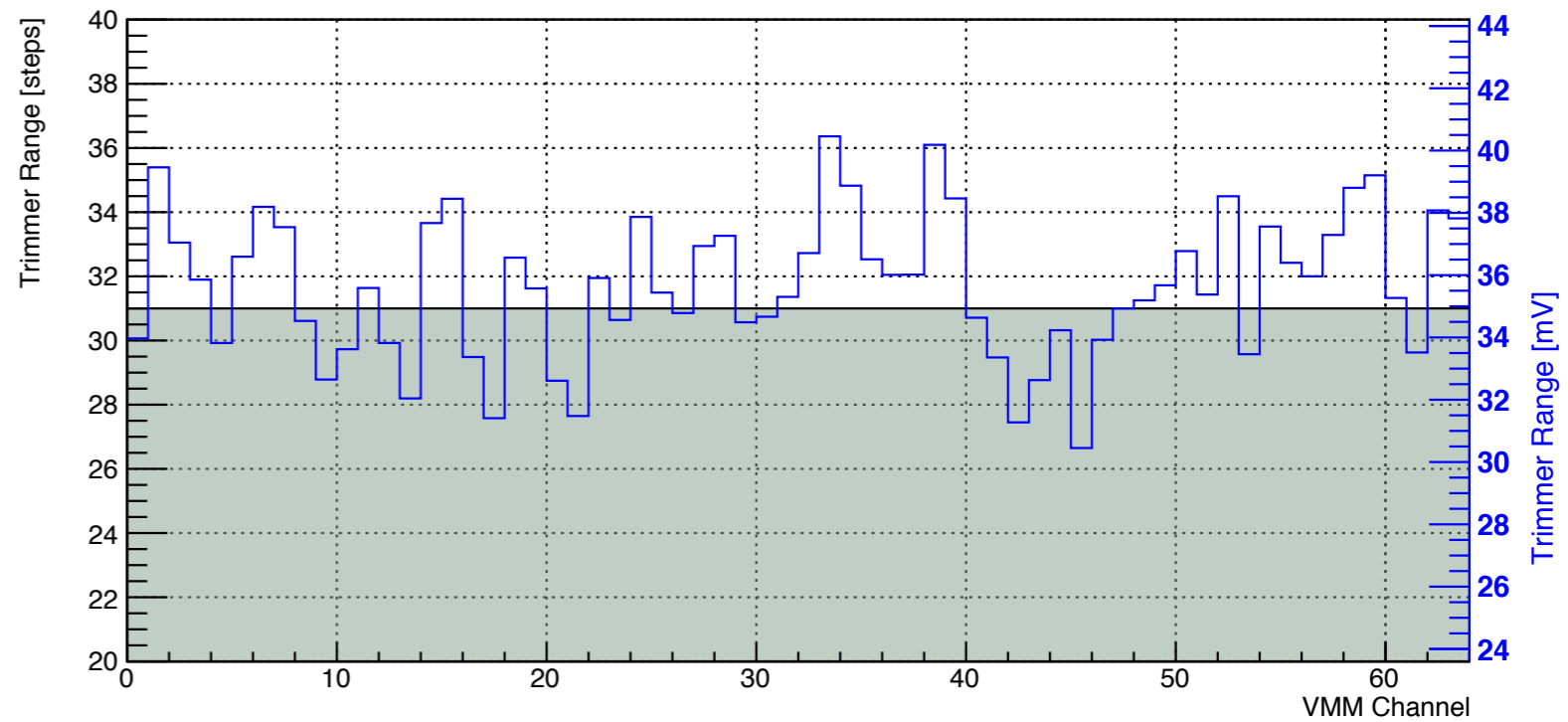
Operation parameters (FPGA) →

Data flow monitor →

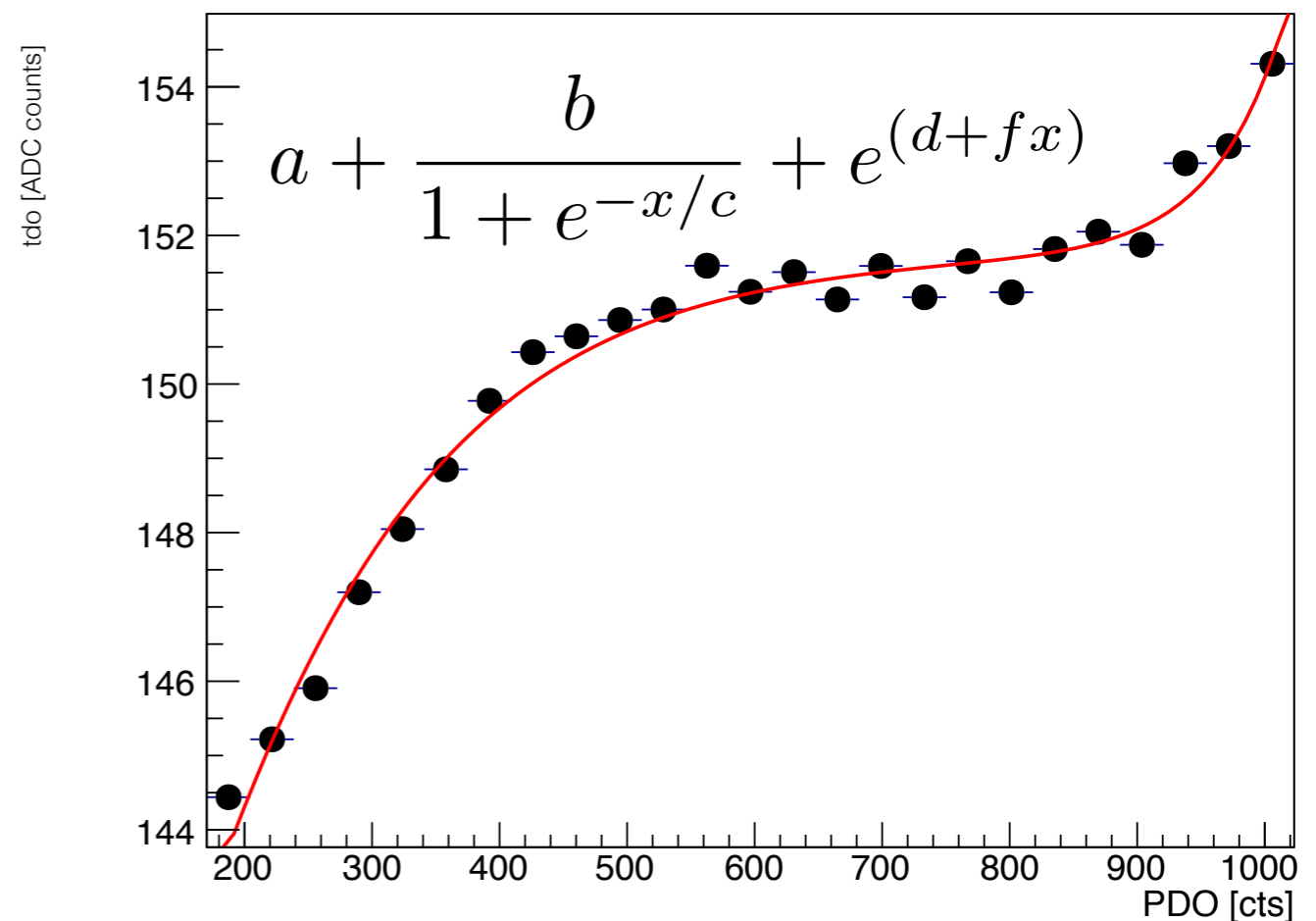
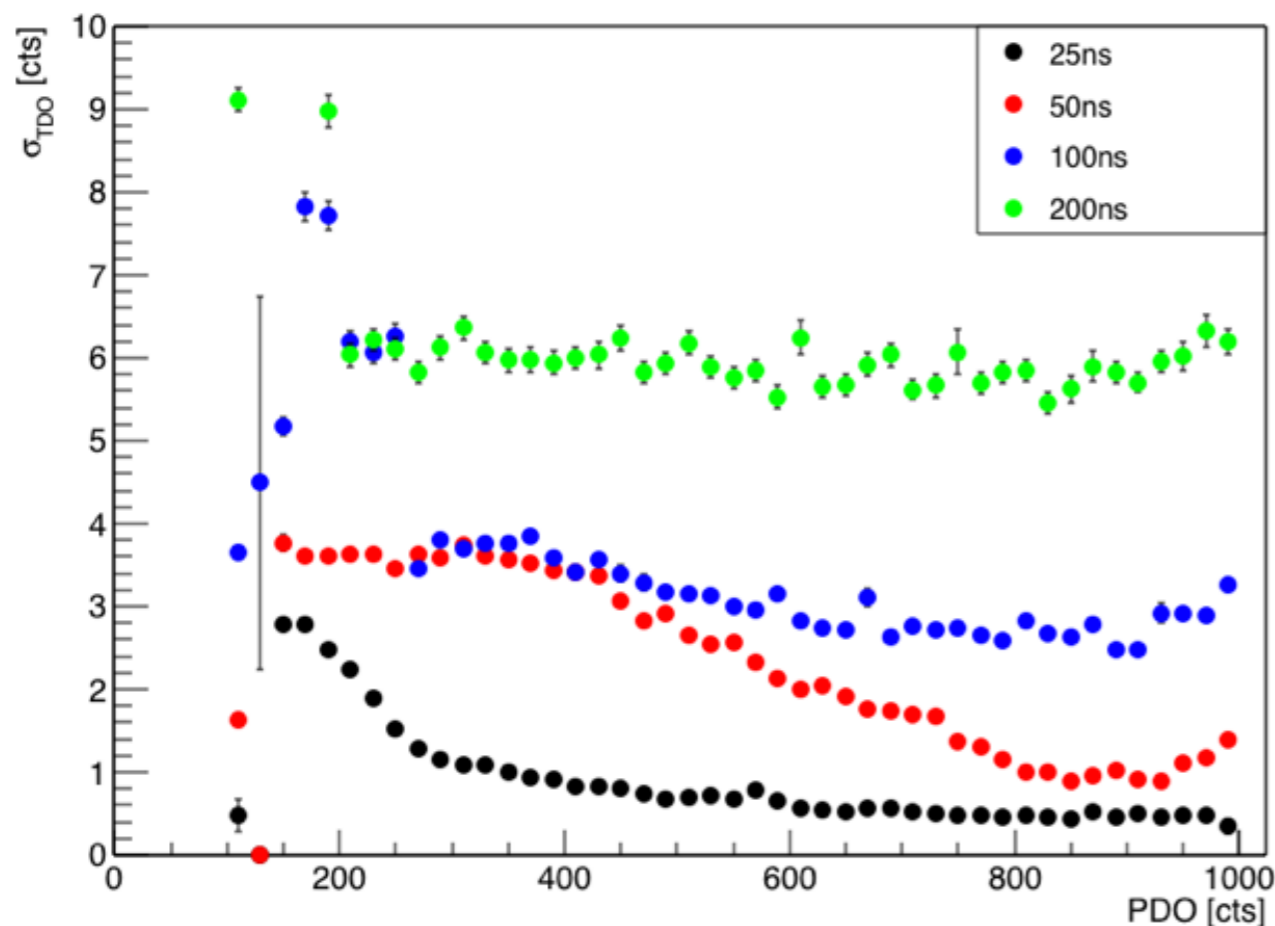
VMM Configuration Calibration Tab Debug monitor ←

The screenshot shows the VERSO - v4.2.1 application window. The top section contains 'Run Status' with buttons for 'Start Run', 'Stop Run', 'Calibration', 'Monitor', and 'DataFlow'. Below this are 'Counters' for Triggers and Hits, and a 'Communication' section showing 'All Boards Alive' and IP addresses. The 'Configure' section includes 'FEB Select', 'VMM Select', 'Mode' (Pulsed, External, Fixed Window), and 'Acquisition' (ACQ On/Off) settings. The 'CKTK & CKBC' section has 'CKTK Max' and 'CKBC Freq.' settings. The 'DataFlow' window at the bottom shows a graph of 'Rate [Hz]' vs 'Elapsed Time of Run' with four data series: Input hit rate, Record hit rate, Record trigger rate, and Empty packet rate.

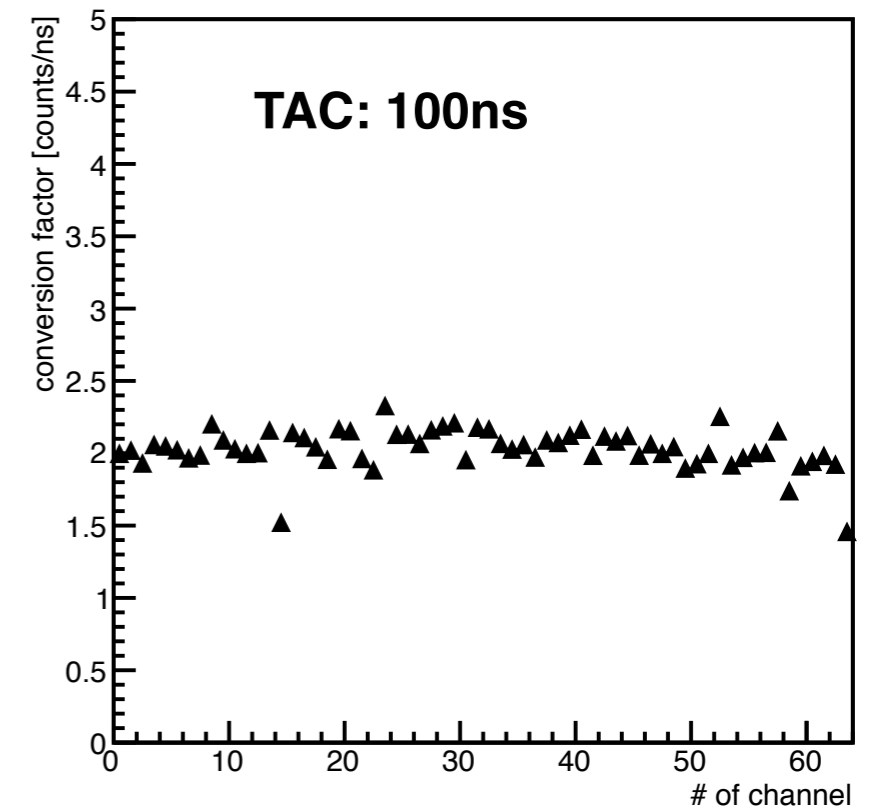
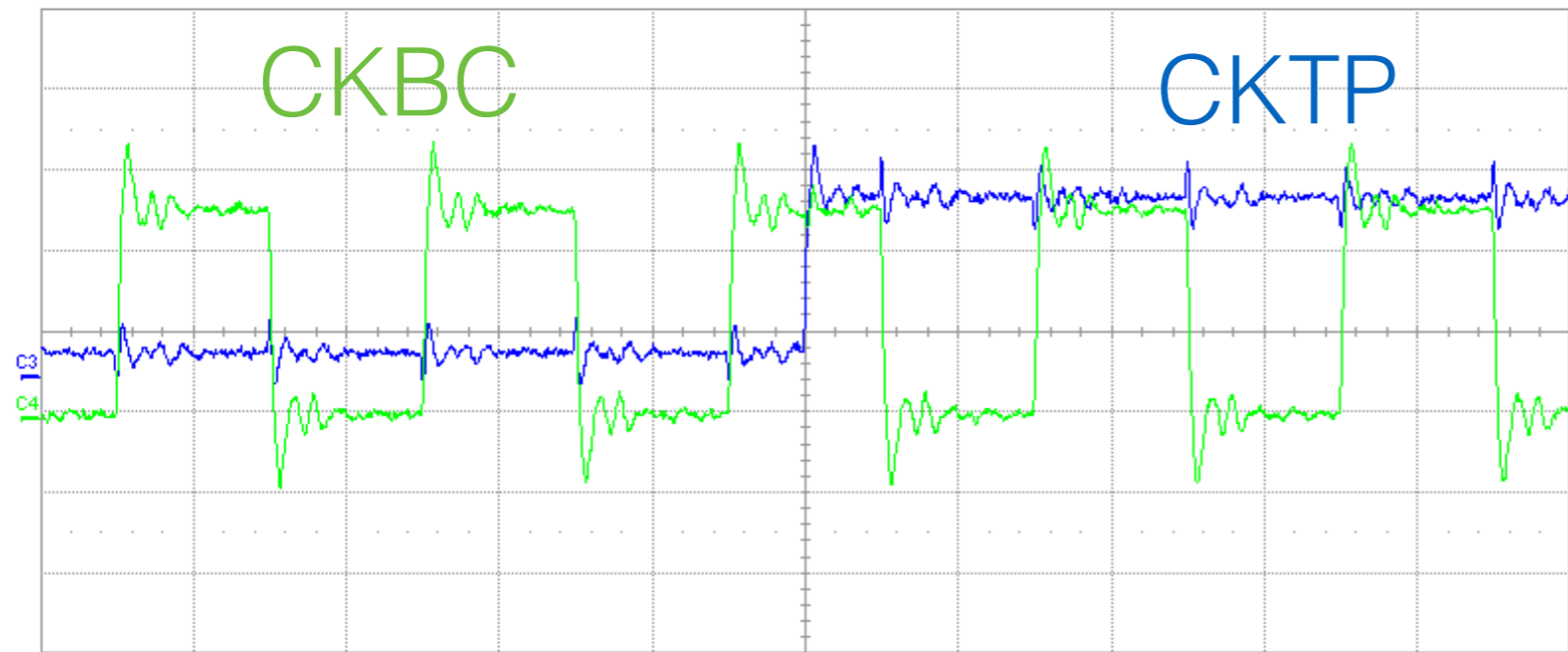
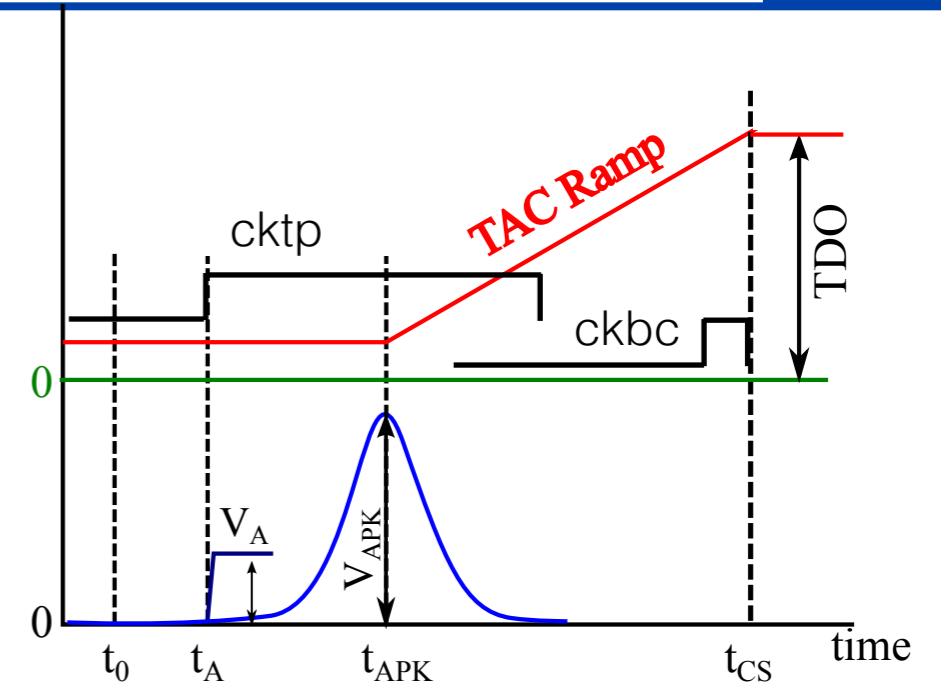
- Full scan performed on the channel trimmers across full threshold range
- Found that the full range of 31 counts shows normal behaviour
- Minimum range of 30mV across all channels, good uniformity.
- Equalisation can be performed easily



- Calibration of the following:
 - Timing resolution along amplitude: This is taking into account on the fitting of the event as an error. Other errors like the longitudinal diffusion is negligible with respect this.
 - Time walk: There is a dependance of the time finding (peak or threshold) from the signal amplitude. This is a correction applied on the timing reconstruction. Fitting the full distribution will improve more the results. To be done.

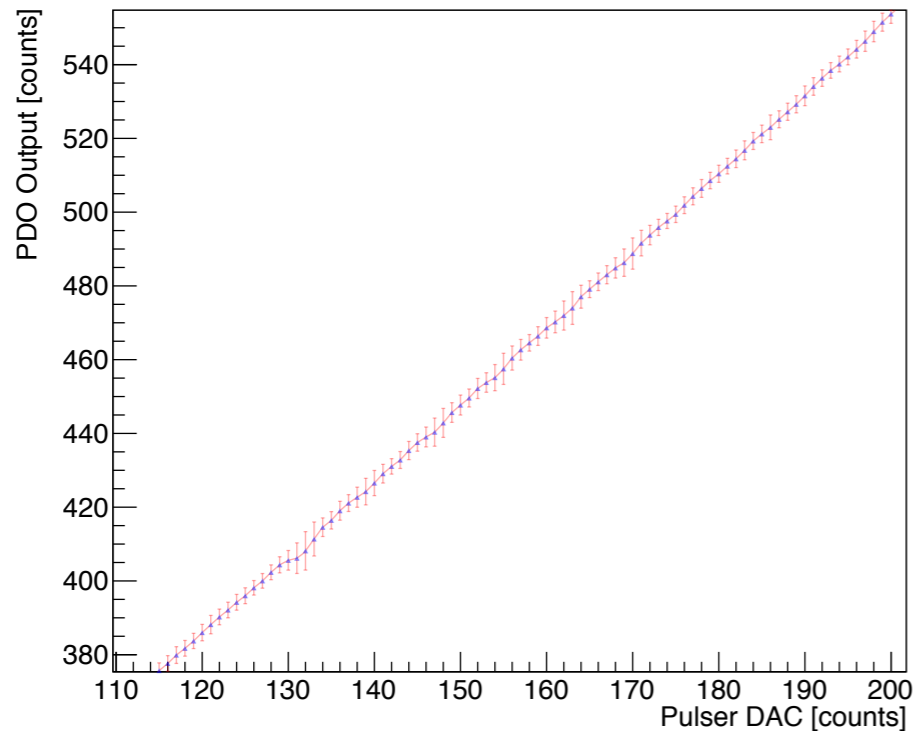


- Calibration of the TAC for different ramps was automatically performed. Skewing clocks method (NSW mode, VMM3/3a) and latency method (Not NSW VMM3) were used to extract the ramping rate and pedestal.
- Uniformity is good, the extracted constants were used in analysis to convert from ADC counts to ns

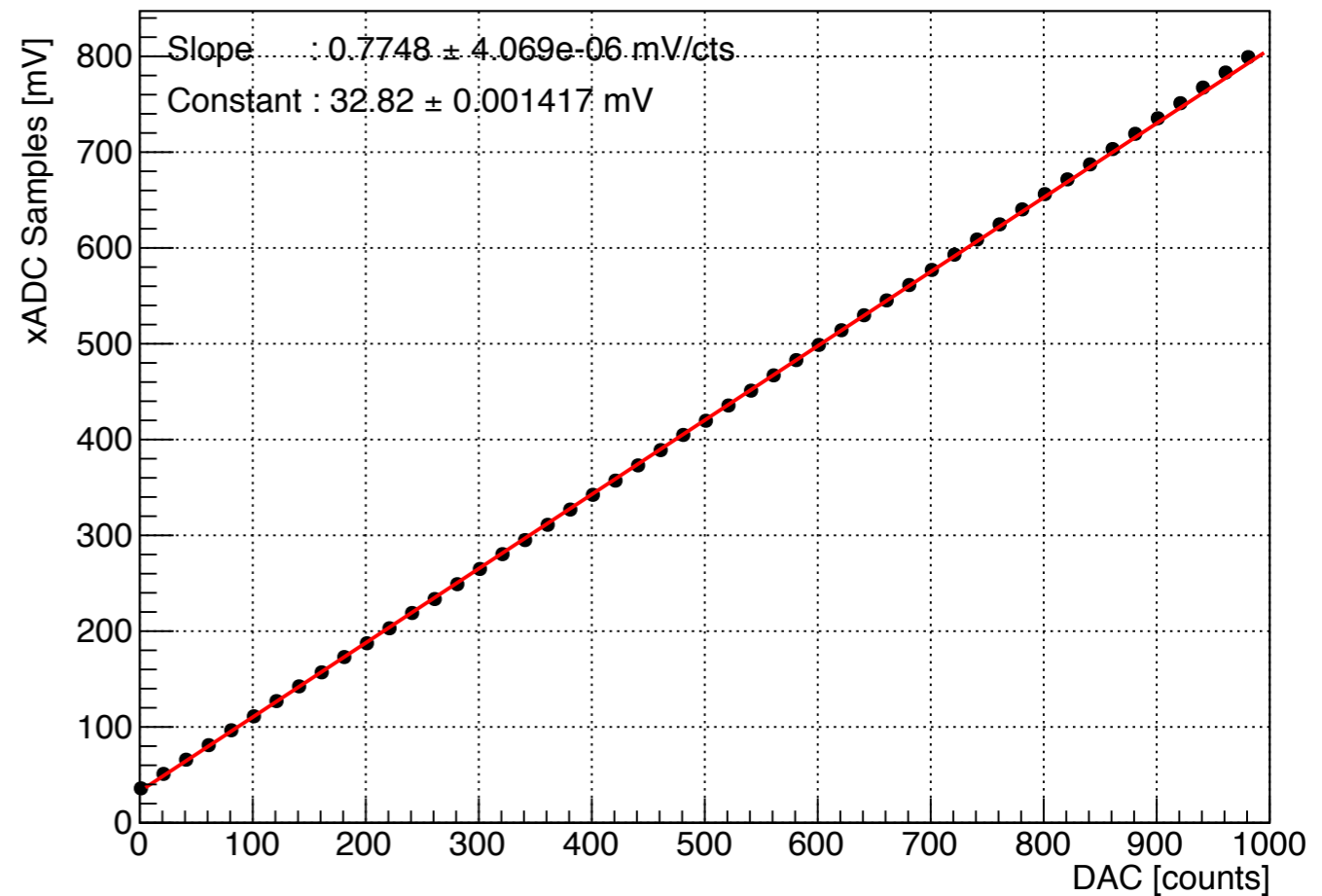


- VMM features as well an internal pulser which can cover the full range on all the gain settings
- Varying the input and measuring the PDO, a charge/gain calibration can be done for each channel

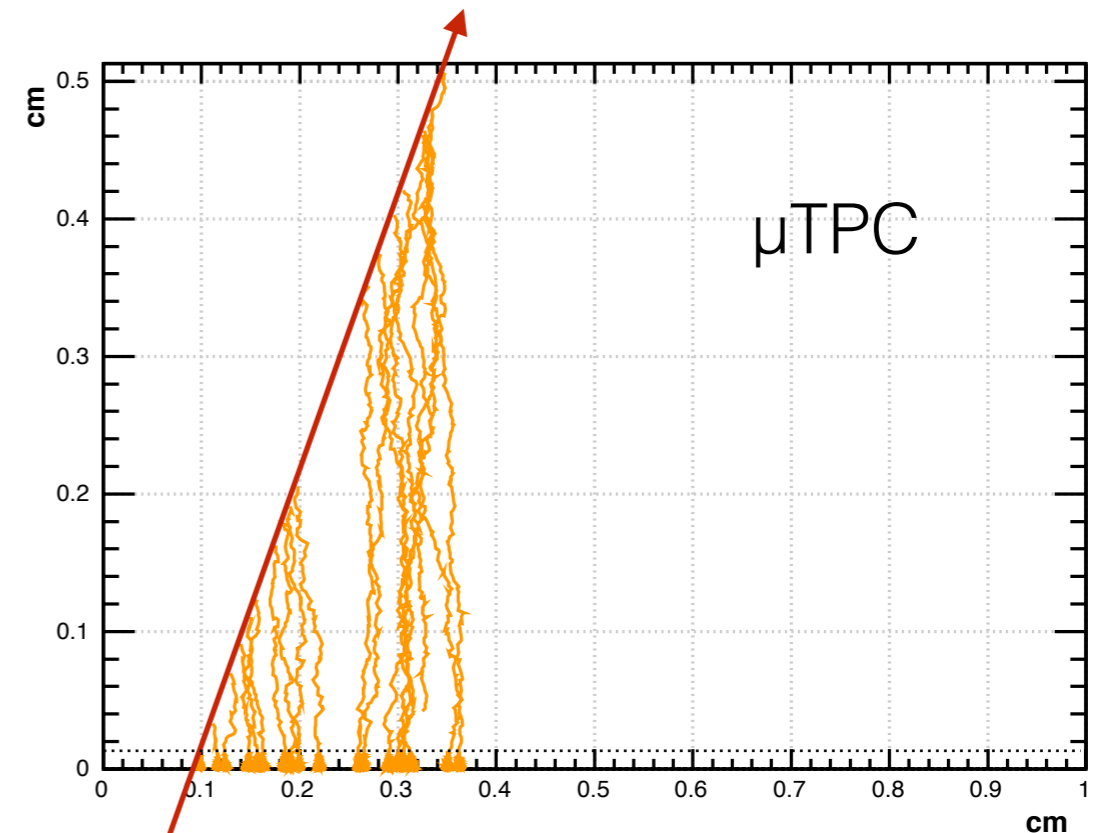
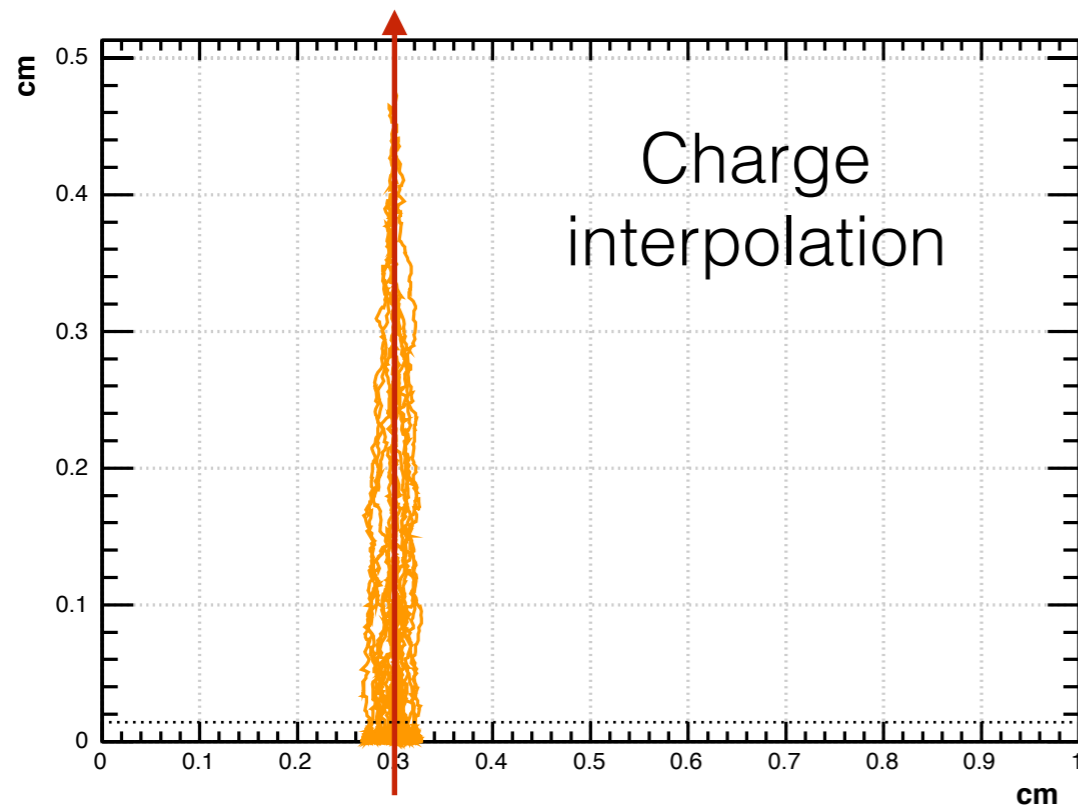
Channel bits (defaults are 0)	Description
sc [0 1]	large sensor capacitance mode ([0] <~200 pF , [1] >~200 pF)
sl [0 1]	leakage current disable [0=enabled]
st [0 1]	300 fF test capacitor [1=enabled]
sth [0 1]	multiplies test capacitor by 10
sm [0 1]	mask enable [1=enabled]
sd0-sd4 [0:0 through 1:1]	trim threshold DAC, 1 mV step ([0:0] trim 0 V , [1:1] trim -29 mV)
smx [0 1]	channel monitor mode ([0] analog output, [1] trimmed threshold))
sz010b, sz110b, sz210b, sz310b, sz410b	10-bit ADC zero
sz08b, sz18b, sz28b, sz38b	8-bit ADC zero
sz06b, sz16b, sz26b	6-bit ADC zero



Pulser DAC xADC Calibration [board,chip]=[2,0]



- In the NSW the **track range** is 5° - 28° for the tracks originated from the IP
- It is mandatory to be able to **reconstruct the position** of a particle that transversed the detector under an angle **with high resolution**
- The **charge centroid method** is proven **not to be able to provide** good resolution for tracks over 5° .
- Instead a **new method** was implemented in the Micromegas called the **μ TPC**
- This method implies on **measuring with high accuracy ($<3\text{ns}$) the arrival time** of the primary ionisation above a strip.

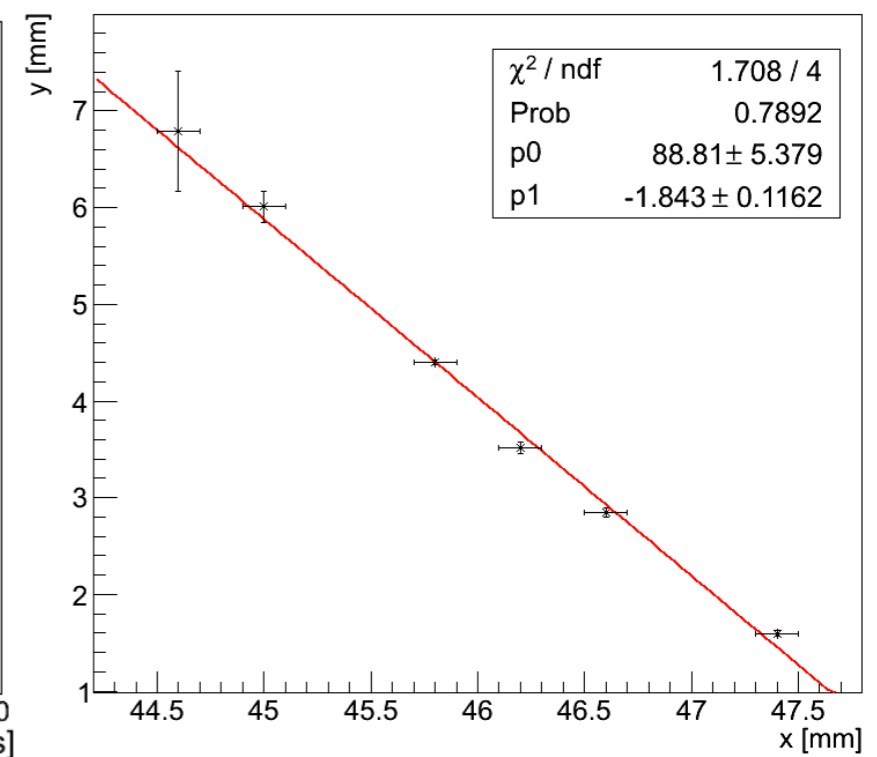
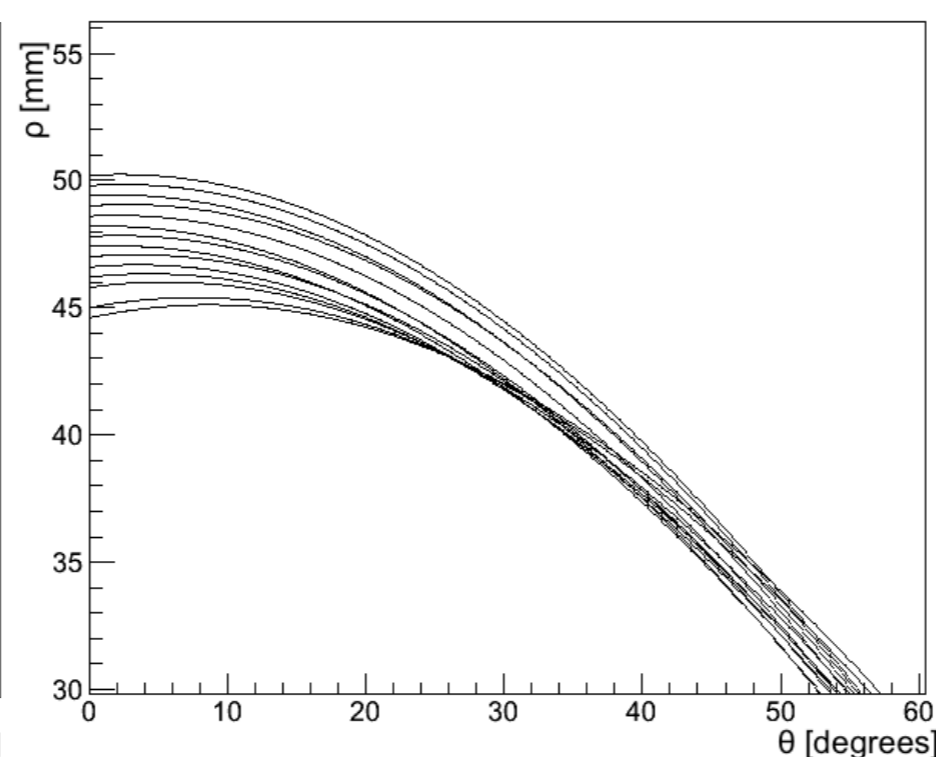
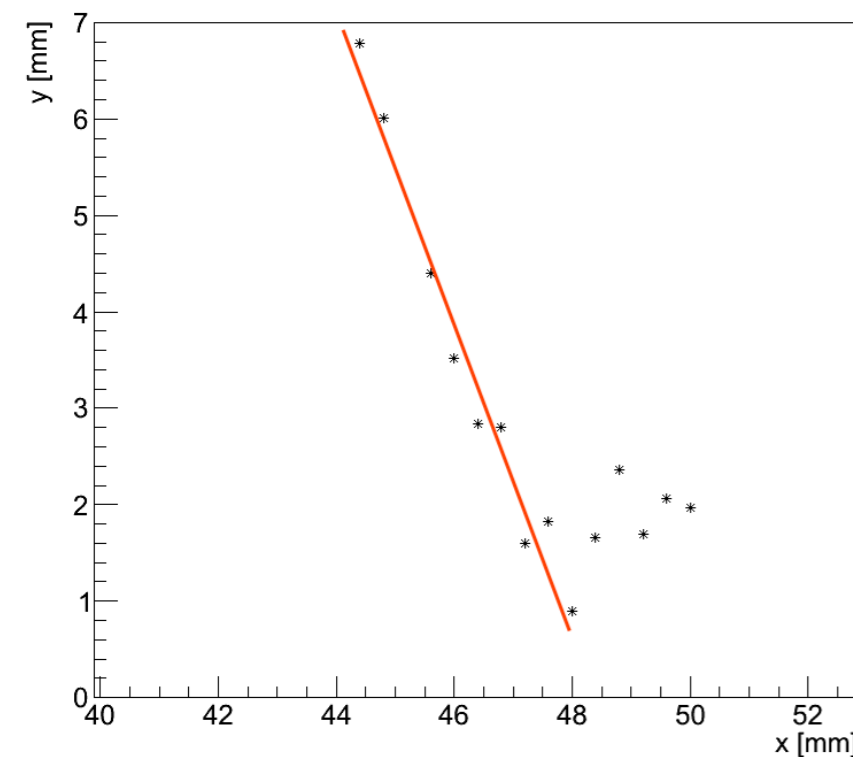


- **Clustering** strips for **inclined tracks** is a challenging task due to:
 - Ionisation statistics, there are **fluctuations** on **primary** cluster generation that can give "holes" in between a cluster of strips **depending on the incident track angle**
 - Generation of **delta electrons**
 - **Multiple** track events
 - **Noise** in the system
- For this reason, a **pattern recognition technique** including the **Hough Transform** is used as a **filter** efficiently removing noise, delta electrons, separating double track events

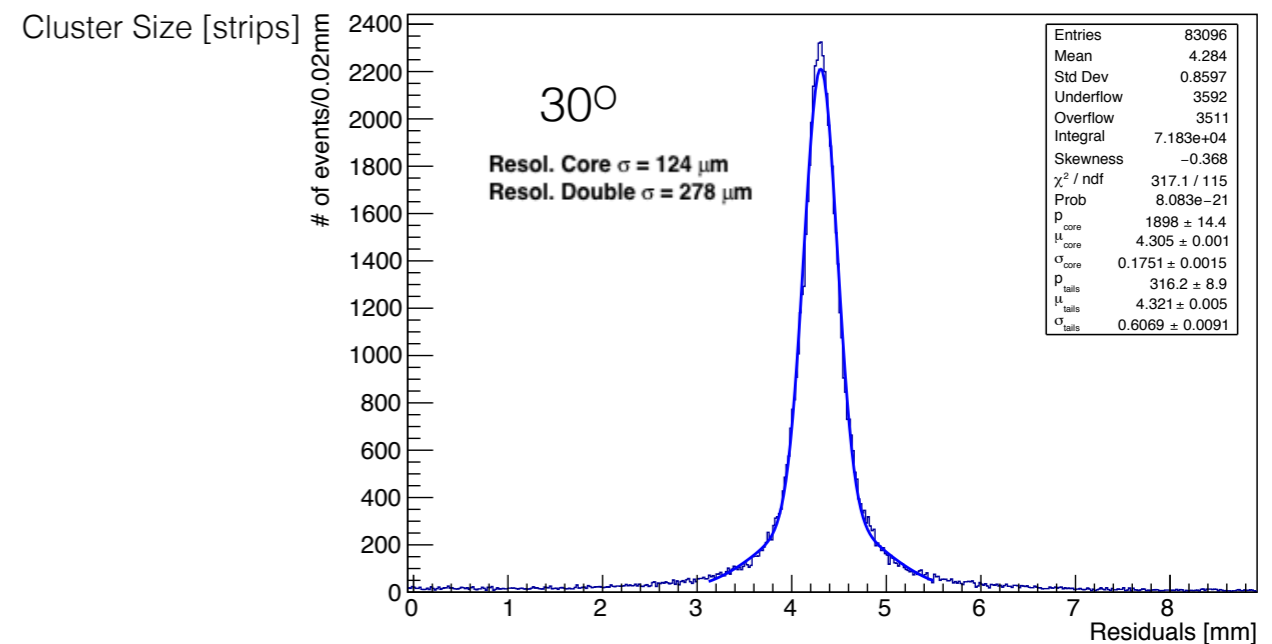
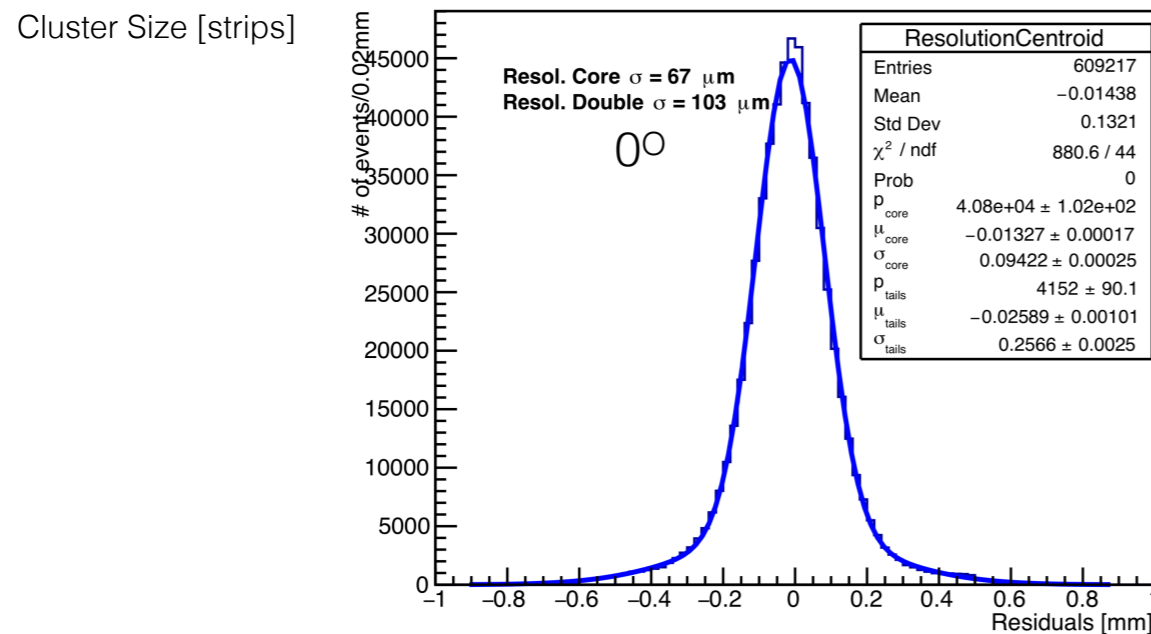
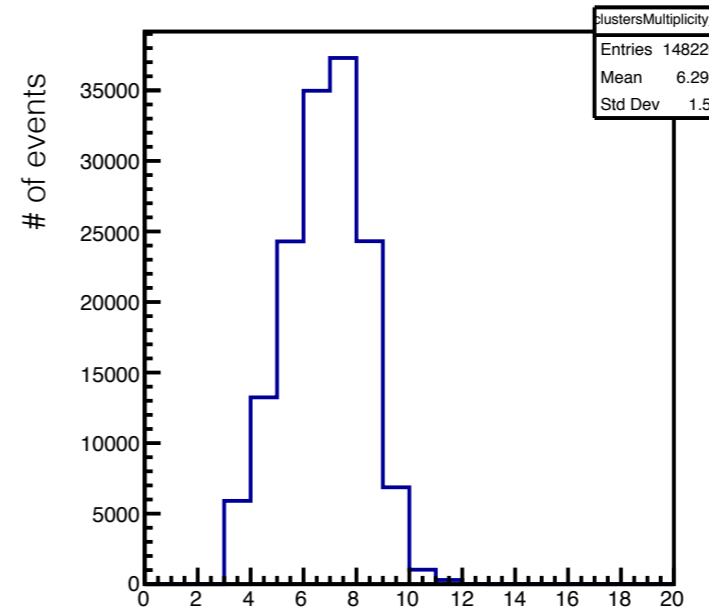
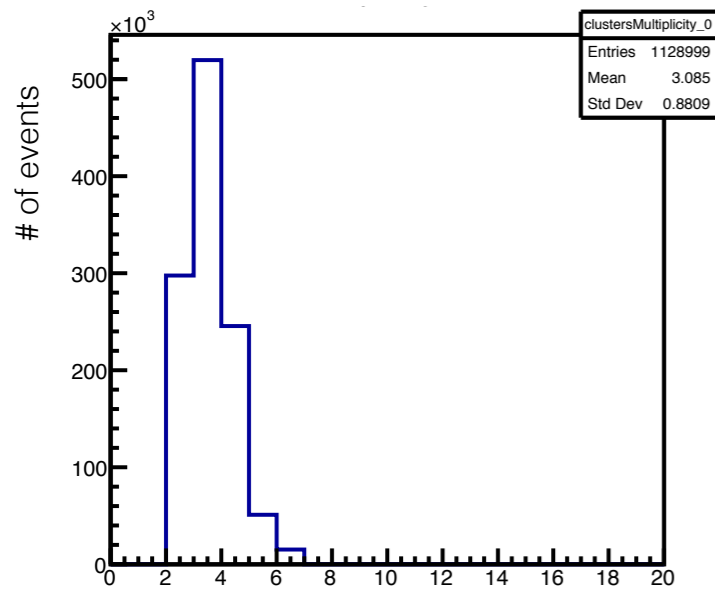
Hough line

Hough Space

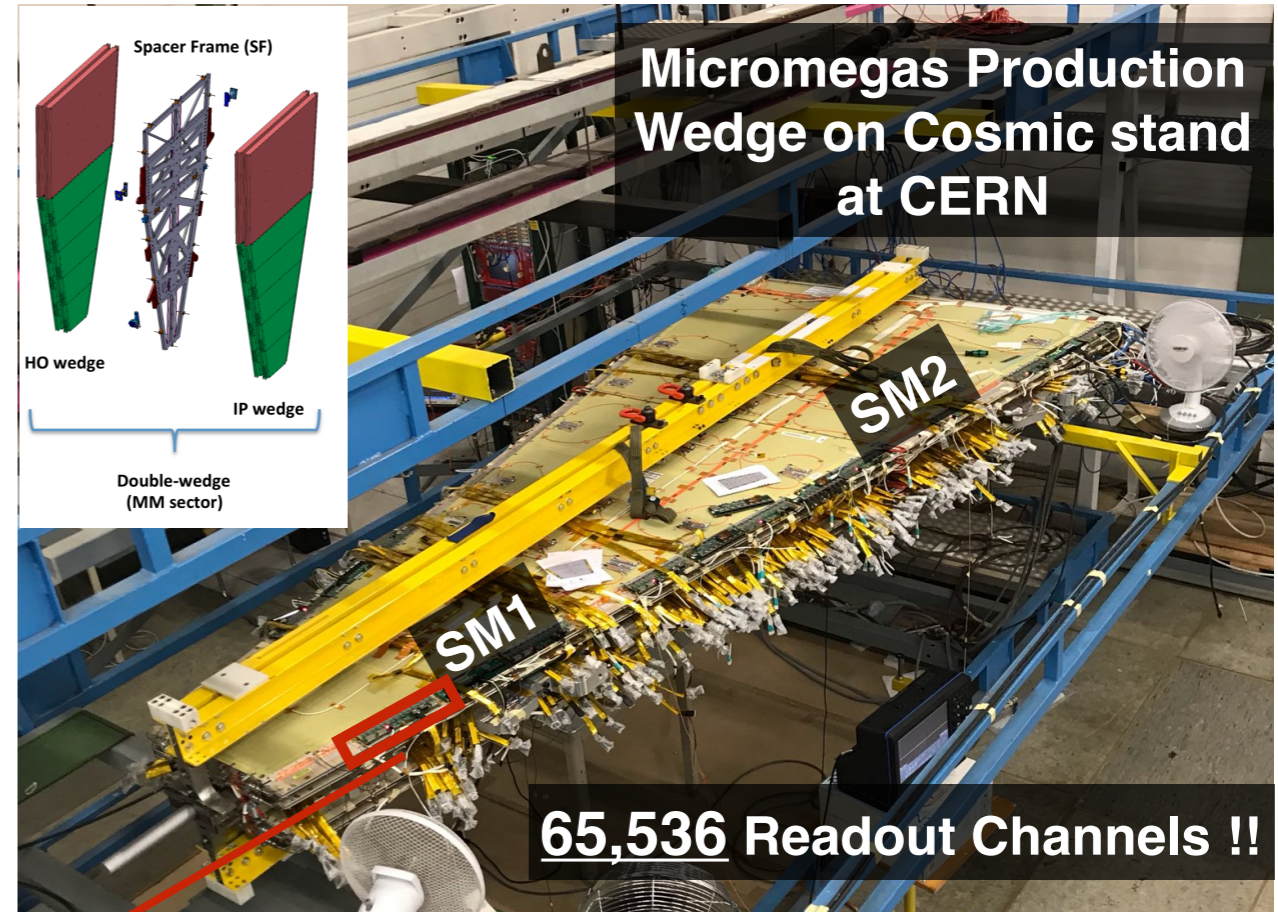
Linear fit after filter



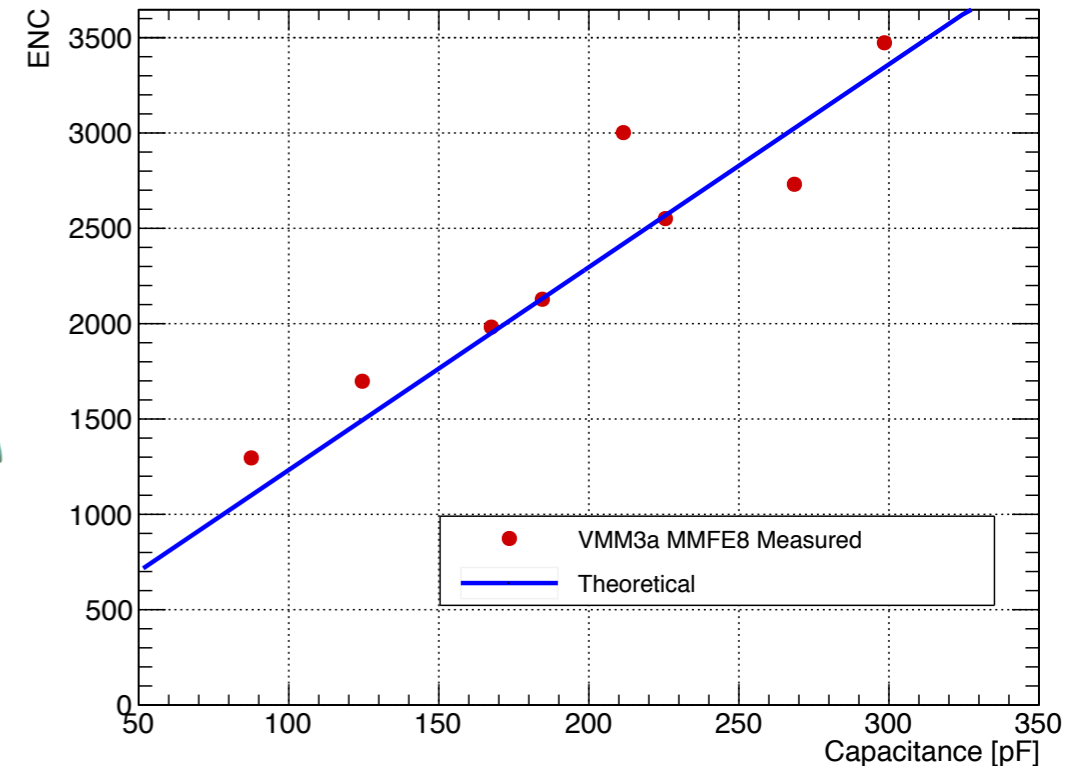
- To form **clusters** of strips per particle hit **charge centroid** was used for perpendicular tracks and **μ TPC** for tracks under an angle with a pattern recognition filter
- After **forming the clusters** the position **resolution** is measured by **subtracting** the space point reconstructed in different layers (identical layers)
- The spatial performance of the detectors is **satisfactory** for the NSW application



- **Micromegas** Production modules are arriving at CERN BB5 **integration** facility
- They are **assembled** on either side of a stiffening panel
- On either side of the wedge, **32x MMFE8** frontend boards are installed (128x total/sect)
- **Alignment** platforms are glued on the surface
- All the FE boards are readout through custom made 4.8Gbit serialiser boards with fibres
- **Highly complex** services routing (electronics readout, fibres, LV, HV, cooling, gas)

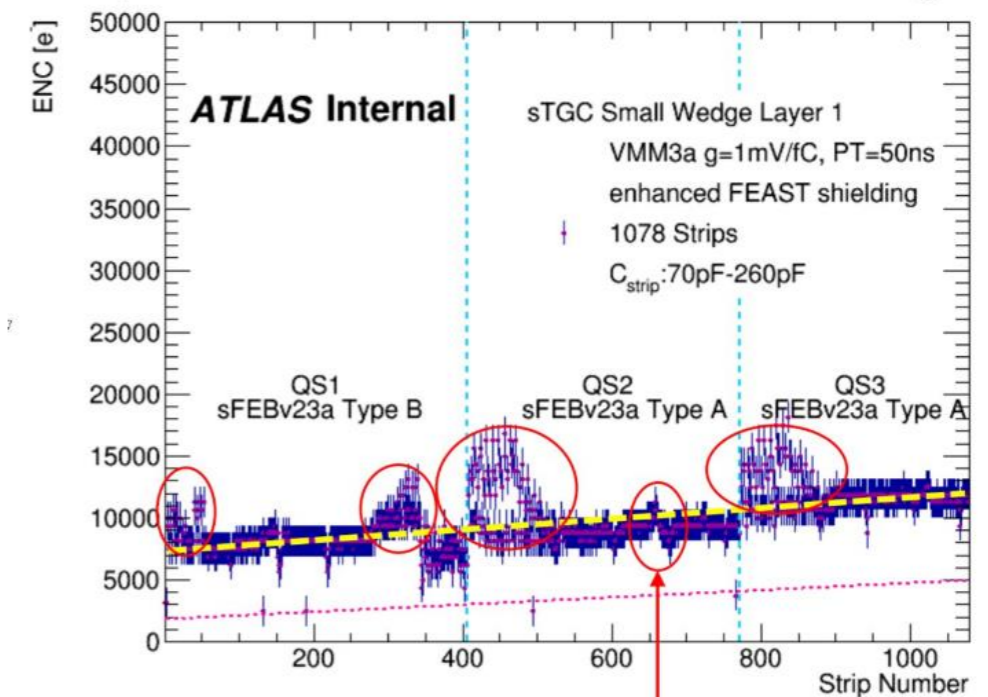


400-pin
21 x 21 mm²
BGA
1mm pitch

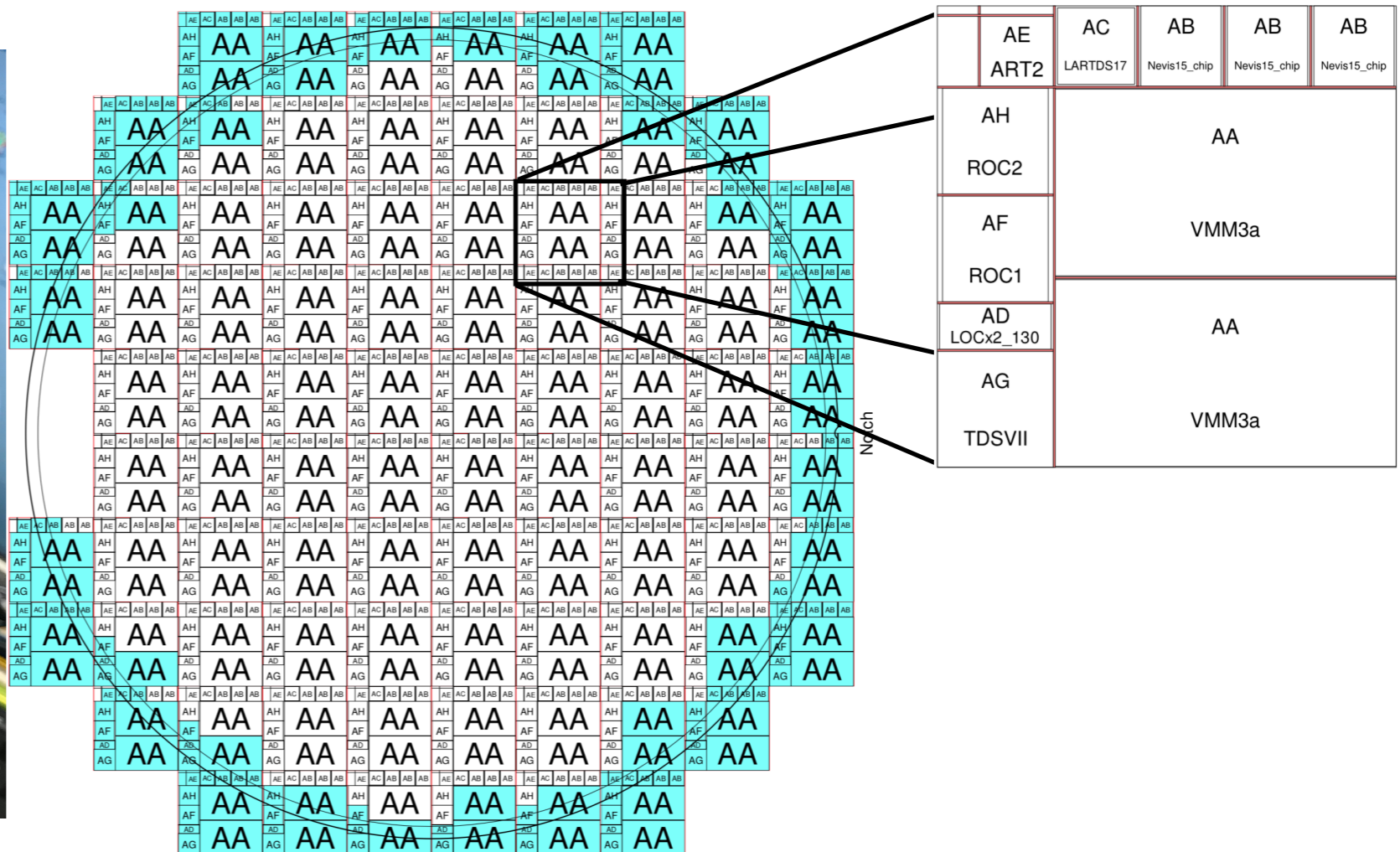
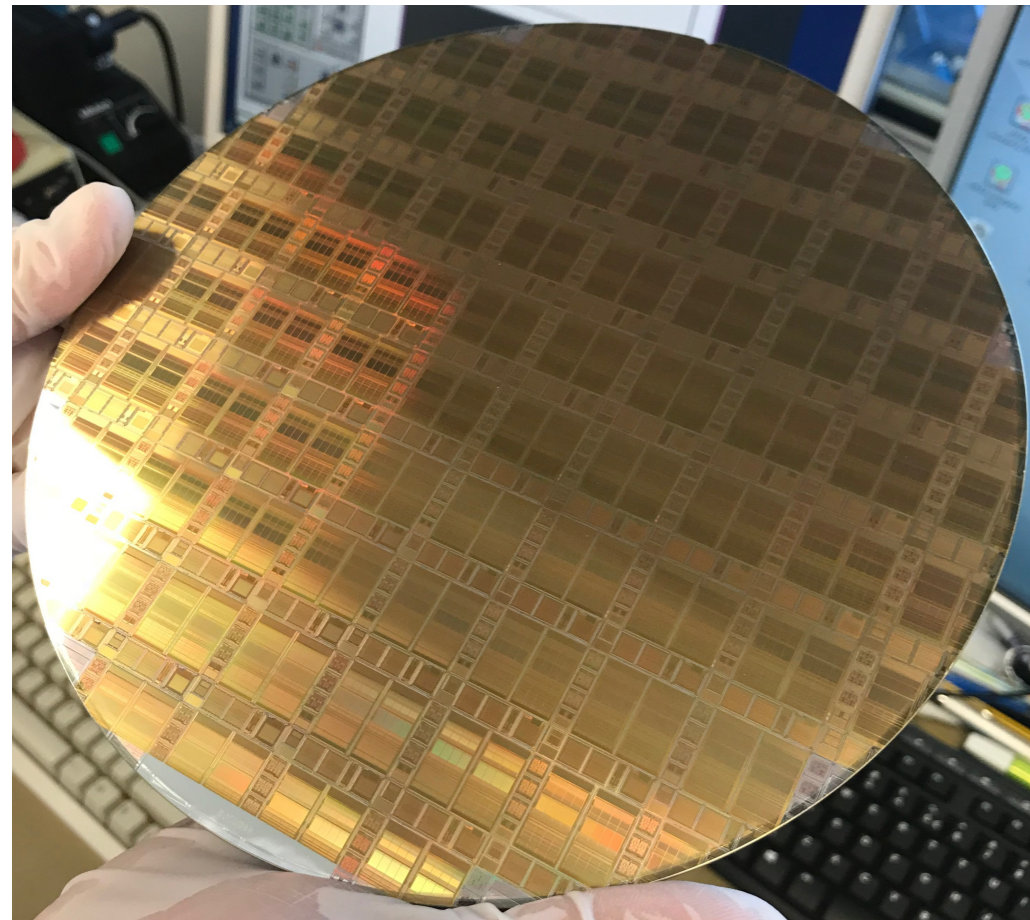


- **Noise** measurements with the production electronics and chambers show good performance and **matches the theoretical expectations**

- **sTGC** Production modules are arriving at CERN 180 **integration** facility
- They are **assembled** on either side of micromegas
- On either side of the wedge, **24 sFEB + 24 pFEB** frontend boards are installed
- All the FE boards are readout through custom made 4.8Gbit serialiser boards with fibres



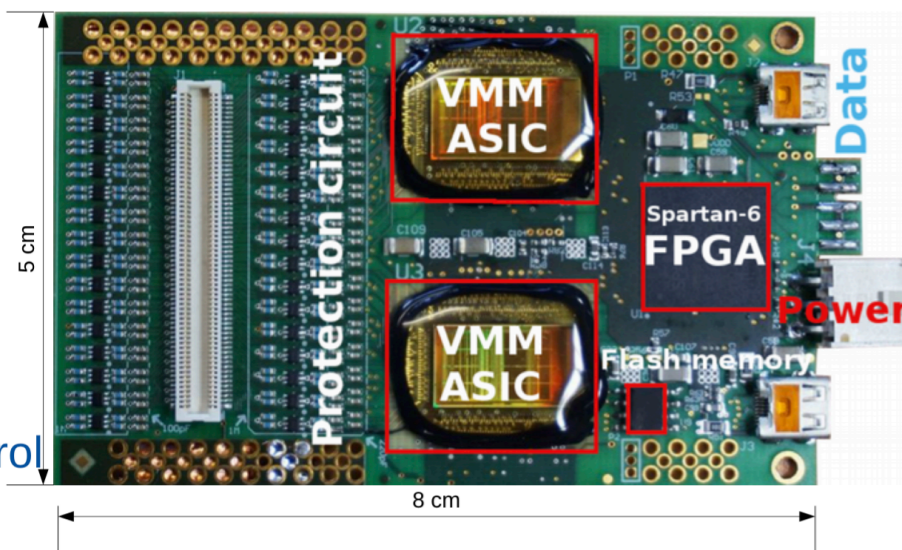
- The **VMM** is produced in a 8" wafer with 2 copies of the chip in a reticle, **total 113 chips per wafer**. In the same floor-plan other ATLAS ASICs are included
- ATLAS has submitted a **production order of 70,000 Chips** - delivery starts on mid-May 2019
- Many iterations with experts from Global Foundries to improve the yield (currently ~72% due to damage on the Baseline stabiliser circuit). Already got indications on issues in their processes
- No export restrictions, courtesy of BNL which grants access for manufacturing under a license



VMM has been as well **of interest** and in some cases **accepted already** in the following other than NSW applications:

- Focal Plane Detector for NUMEN
- interest from n_TOF at CERN
- Mu2e at Fermilab
- DUNE Near Detector at Fermilab
- CERN RD51 SRS system (replace APV hybrids) which is a hub for **many other applications**

SRS-Mini2



GPVMM



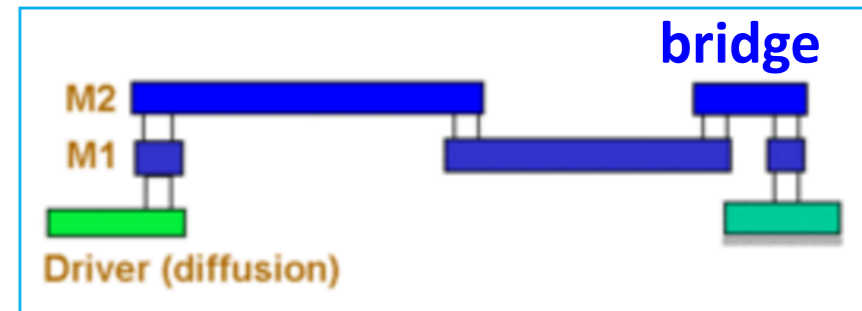
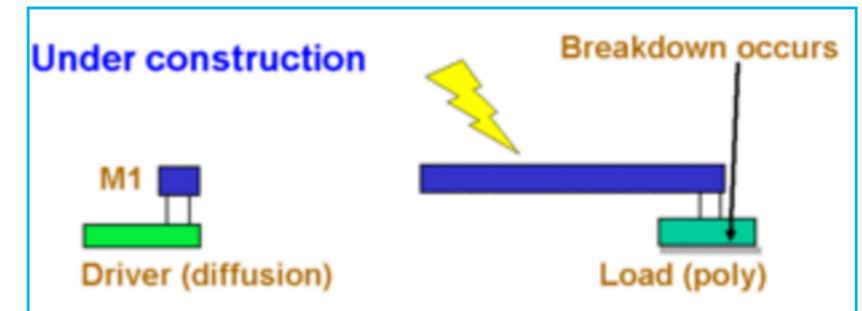
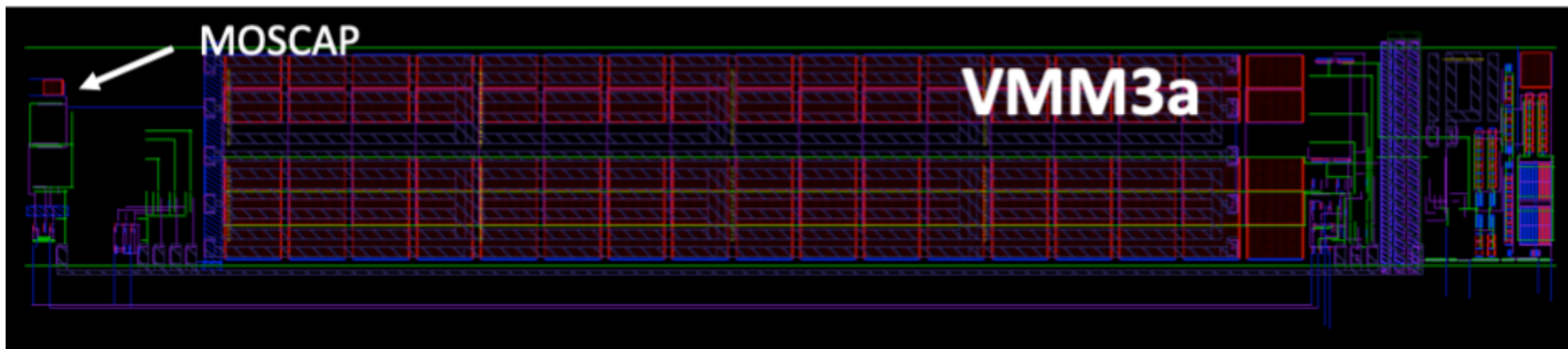
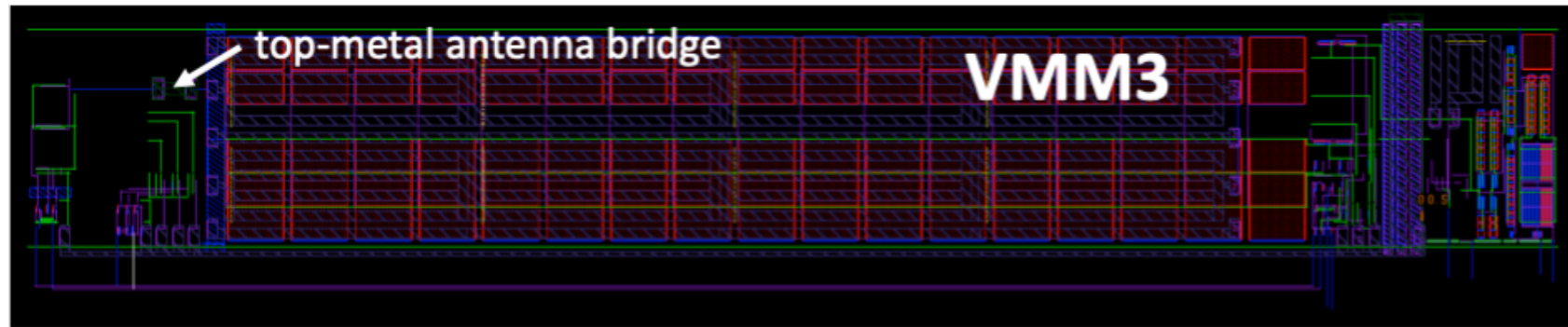
mu2e

Group	Application	VMM hybrids	Contact
ESS Lund / BrightnESS	NMX instrument @ ESS	164	Dorothea PFEIFFER
University of Science and Technology of China	RICH R&D for future colliders in China (CEPC and STCF)	156	LUI Jianbei
Bonn University	BASTARD neutron detector	71	Jochen KAMINSKI Markus KÖHLI
Mainz University	MAGIX experiment @ MESA*	211	Stefano CAIAZZA
Budker Institute of Nuclear Physics, Novosibirsk	μ Well MPGD R&D	22	Lev I. SHEKHTMAN
INFN Tieste	Generic R&D	10	Silvia DALLA TORRE
Tsukuba University	ALICE FoCal, Si Pads	50	CHUJO Tatsuya
GDD group CERN	Generic R&D	16	Eraldo OLIVERI
Peking University	CMS GEM upgrade	52	Dayong WANG
LMU Munich	Ion Tomography with Micromegas	16	Felix KLITZNER
LMU Munich	Medical physics with MPGDs, Si	48	Jona BORTFELDT
ETH Zurich	GBAR experiment @ CERN	\approx 40	Gianluca JANKA
CERN	BGV(Beam Gas Vertex) beam monitor*	200	Robert KIEFFER
University of Virginia, Charlottesville	EIC tracker @ RHIC*	Not known yet	Kondo GNANVO

- **VMM developments** for the last 7 years **concluded** with a **successful production** version for ATLAS, **VMM3a** (fourth iterations from the beginning)
- **VMM3a** was **tested** thoroughly on bench, prototype and production detectors achieving the needed performance
- **Already a success** being **proposed** for multiple **systems** and **experiments**
- **New Small Wheels** are the first and the **biggest MPGD upgrade** in high energy physics and **VMM** was **validated** after multiple reviews to **fulfil the requirements**
 - ATLAS entered the **production of 70,000 chips**.
 - CERN RD51 joined the production with 2,800 chips

Thank you for your attention

- On VMM3a, under the suspicion of antenna damage, the bridge of the top layer was replaced by a MOSCAP
- Just to note that both designs are satisfying the DRC for antenna damage



- The issue cannot be explained from simulation: Points to damage of the gate, or degradation of it

